

MATHEMATICAL MODEL OF CONTROL SYSTEMS

Introduction:-

System: when a no. of elements or Components are connected in a sequence to perform a specific function is known as a system.

Control System: In a System when the output quantity is controlled by varying the input quantity, then the System is called a Control System.

Types of Control Systems:

1) open loop System

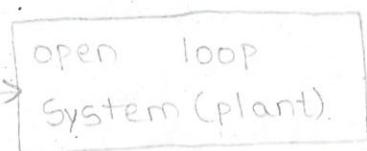
2) closed loop System

Open loop System:-

Command Signal

Excitation

$\theta(t)$



Response.

Any physical System which does not automatically correct the variations in its output is called open loop System.

control system in which the output quantity has no effect upon the input quantity are called open loop Control System.

In open loop System the output can be vary by varying the input but due to external disturbances the System output may

in output input to correct the output. In open loop system, the changes in output are corrected by changing the input manually. Therefore, these are called manual control systems.

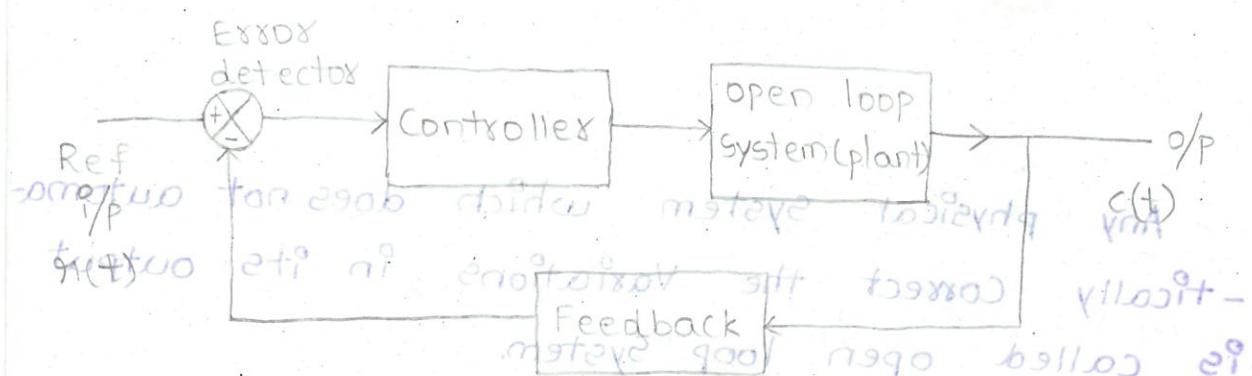
Advantages:-

- the open loop systems are simple and economically.
- the open loop systems are easier to construct.
- Generally, the open loop systems are stable.

Disadvantages:-

- these are inaccurate and unreliable.
- the changes in output due to external disturbances are not corrected automatically.

Closed Loop System:-



Control Systems in which the output has an effect on output upon the input quantity in order to maintain the desired output value are called closed loop systems. The open loop system can be modified as closed loop system by providing feedback. The feedback automatically corrects the changes in o/p

System: It is a system which takes input from environment and gives output to environment.

Advantages:

- These are more accurate.
- The sensitivity of the system may be made small to make the system more stable.

→ The closed loop systems are less effected by noise.

Disadvantages:

→ These are complex and costly.

→ The feedback in closed loop system may lead to oscillatory response.

→ The feedback reduces the overall gain of the system.

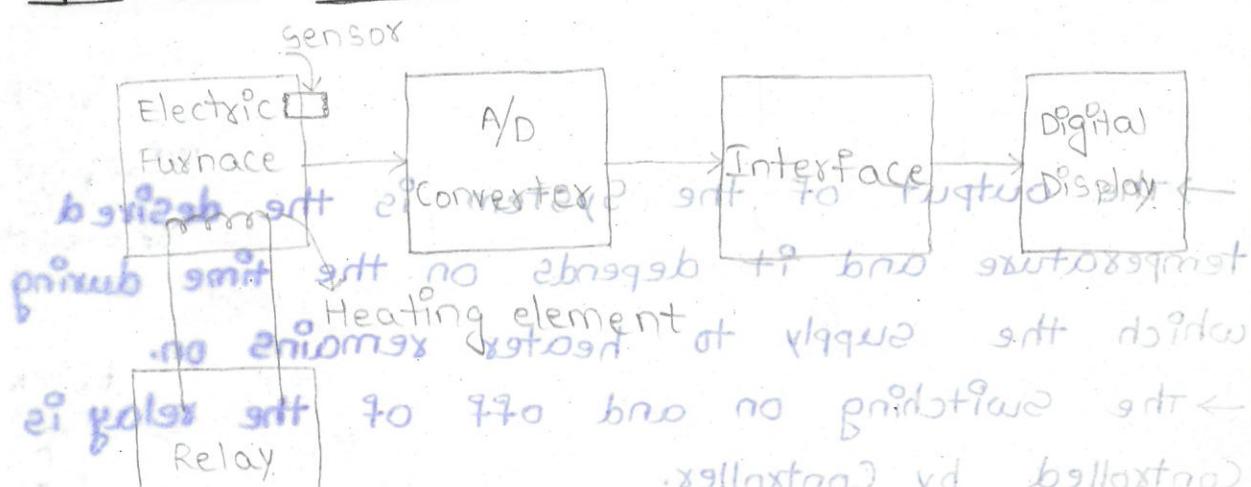
→ Stability is a major problem in a closed loop system hence, more care is needed to design a stable closed loop system.

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Examples of Control Systems:

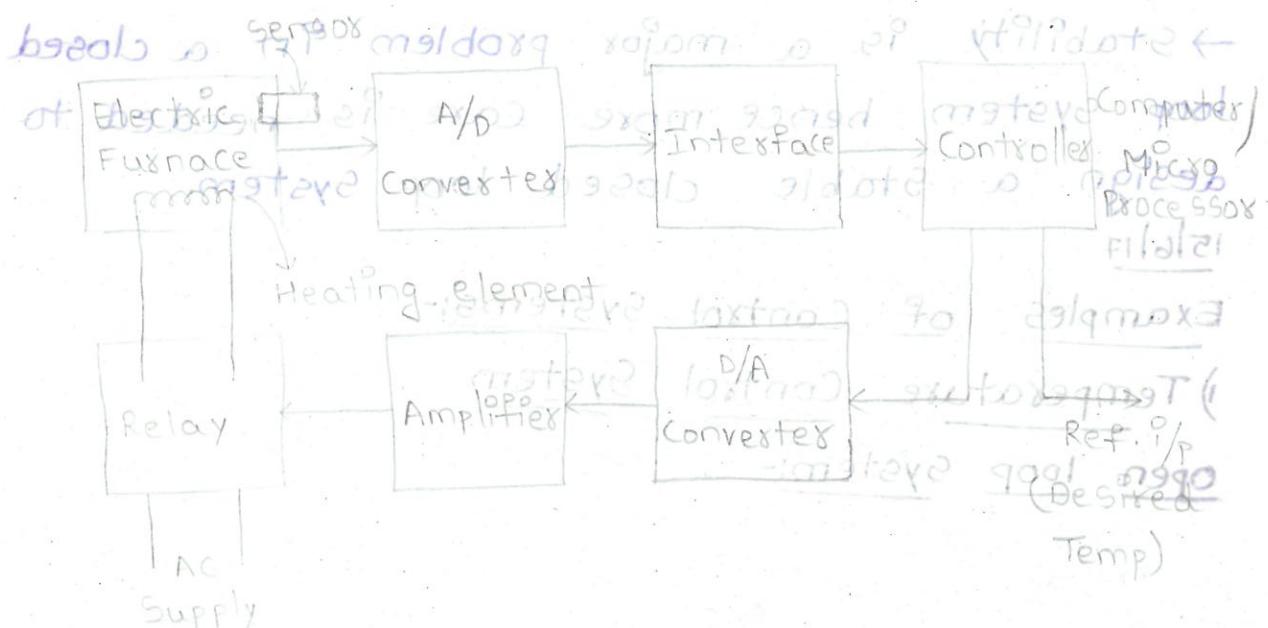
1) Temperature Control System

open loop System:



- The output in the system is the desired temperature and it depends on the time during which the supply to heater remains on.
- the on and off of the supply is controlled by time setting of the relay.
- the temperature is measured by sensor and converted to digital signal by A/D converter.
- The digital signal is given to digital display to display the temperature.
- In this system, if there is any change in output temperature then the time setting of the relay is not altered automatically.

Closed loop System:-



- The output of the system is the desired temperature and it depends on the time during which the supply to heater remains on.
- The switching on and off of the relay is controlled by controller.

and Converted to Digital Signals by A/D Converter.
→ The Controller Compares the actual temperature with desired temperature.
→ If it finds any difference then, it sends signal to switch on (or) off.
→ the relay through D/A Converter and Amplifier.
thus, the system automatically corrects any changes in output.

2) Traffic Control System:

open loop System:

→ Traffic Control by traffic signals are operated on time basis constitutes an open loop system.
→ The sequence of control signals are based on time slot given for each signal. The time slot are decided based on traffic.

→ The system will not measure the density of traffic before giving the signals. Since, the time slot does not change according to traffic density, the system is open loop system.

Closed loop System:

→ Traffic Control System can be made as a closed loop system if the time slots of the signals are decided based on density of traffic.

→ In closed loop traffic control system the density of traffic is measured on all the

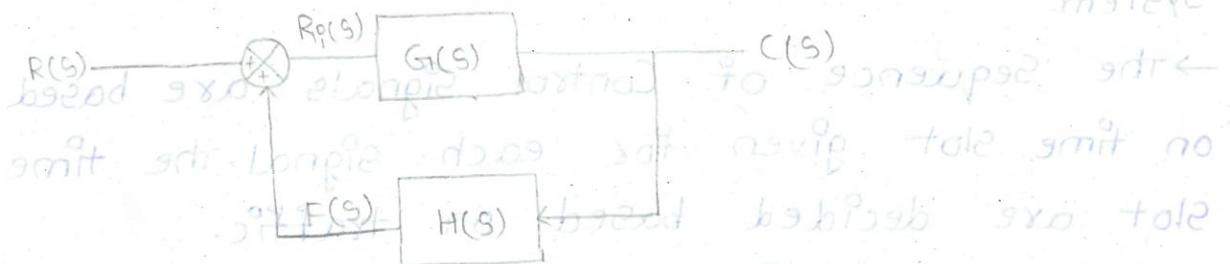
→ then, the computer decides the timings of the control signals based on density of traffic.
 → since, the closed loop system dynamically changes the timings the flows of vehicles will be better than open loop system.

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Feedback: - the sample portion of output signal is fed to input signal is known as feedback. there are two different types of feedback.

1) Regenerative feedback

2) Degenerative feedback

Regenerative Feedback:



If a feedback signal is a positive going signal w.r.t input signal is known as Regenerative feedback.

The transfer function of closed loop system

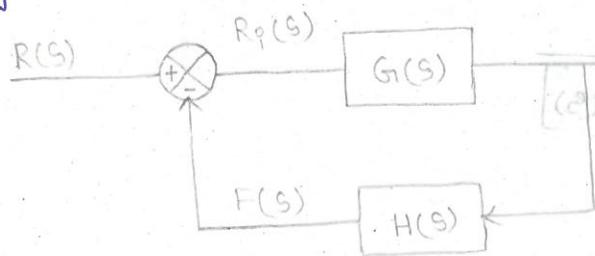
$$\text{is } T(s) = \frac{C(s)}{R(s)} = \frac{G(s)}{1 - G(s)H(s)}$$

where $G(s)$ is transfer function of open loop system and $H(s)$ is transfer function of feedback network.

Degenerative Feedback:

If a feedback signal is negative going

degenerative feedback.



$$\begin{aligned} C(s) &= R_f(s) \\ R_f(s) &= R(s) - F(s) \\ R_f(s) &= R(s) - H(s)C(s) \\ R_f(s) &= \frac{R(s)}{1 + H(s)G(s)} \\ T(s) &= \frac{C(s)}{R(s)} = \frac{G(s)}{1 + H(s)G(s)} \end{aligned}$$

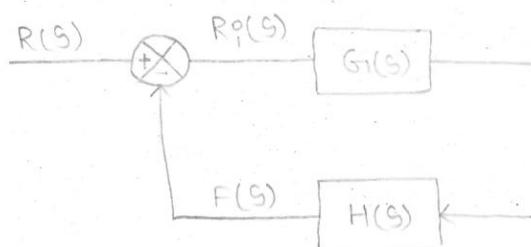
The transfer function of open loop system

is

$$T(s) = \frac{C(s)}{R(s)} = \frac{G(s)}{1 + H(s)G(s)}$$

Effect of feedback on gain:-

Let us consider,



$$\begin{aligned} C(s) &= R_f(s) \\ R_f(s) &= R(s) - F(s) \\ R_f(s) &= R(s) - H(s)C(s) \\ R_f(s) &= \frac{R(s)}{1 + H(s)G(s)} \\ T(s) &= \frac{C(s)}{R(s)} = \frac{G(s)}{1 + H(s)G(s)} \end{aligned}$$

Transfer function of open loop system is

$$G(s) = \frac{C(s)}{R_o(s)} \quad \text{--- ①}$$

Transfer function of feedback system (or)

network is

$$H(s) = \frac{F(s)}{C(s)} \quad \text{--- ②}$$

From the circuit, we have without reference to

$$R_o(s) = R(s) - F(s)$$

$$\Rightarrow R(s) = R_o(s) + F(s)$$

$$\Rightarrow R(s) = R_o(s) + H(s)C(s) \quad [\text{From ②}]$$

$$\Rightarrow R(s) = R_o(s) + H(s)G(s)R_o(s) \quad [\text{From ①}]$$

$$\Rightarrow R(s) = R_o(s)[1 + H(s)G(s)] \quad \text{--- ③}$$

∴ The transfer function of closed loop system

From ③

$$T(s) = \frac{C(s)}{R_p(s) [1 + H(s) G(s)]}$$

$$T(s) = \frac{G(s)}{1 + G(s) H(s)}$$

∴ the transfer function (gain) decreases with feedback.

Effect of feedback on stability:-

the transfer function of closed loop system is

$$T(s) = \frac{G(s)}{1 + G(s) H(s)}$$

AS $G(s) H(s) \gg 1$

$$\therefore T(s) = \frac{G(s)}{G(s) H(s)}$$

$$\therefore T(s) = \frac{1}{H(s)}$$

∴ $T(s)$ depends on $H(s)$ but doesn't depend on $G(s)$, thus, due to feedback stability increases.

Effect of feedback on sensitivity:-

Sensitivity is defined as percentage change in transfer function with feedback to percentage change in transfer function without feedback.

$$\therefore \text{sensitivity} = \frac{\% \text{ change in } T(s)}{[\% \text{ change in } H(s)]}$$

$$\Rightarrow \text{sensitivity} = \frac{\delta T(s)/T(s)}{\delta H(s)/[H(s) + 1]}$$

$$\therefore \text{sensitivity} = \frac{\delta T(s)}{\delta H(s)} \times \frac{G(s)}{G(s) + 1}$$

Partial differentiating $T(S)$ wrt $G_1(S)$

$$\frac{\partial}{\partial G_1(S)} [T(S)] = \frac{\partial}{\partial G_1(S)} \left[\frac{G_1(S)}{1 + G_1(S)H(S)} \right]$$

$$\frac{\partial T(S)}{\partial G_1(S)} = \frac{[1 + G_1(S)H(S)](1) + G_1(S)H(S)}{[1 + G_1(S)H(S)]^2}$$

$$\Rightarrow \frac{\partial T(S)}{\partial G_1(S)} = \frac{1}{[1 + G_1(S)H(S)]^2}$$

$$\therefore \text{Sensitivity} = \frac{\partial T(S)}{\partial G_1(S)} \times \frac{G_1(S)}{T(S)}$$

$$= \frac{1}{[1 + G_1(S)H(S)]^2} \times \frac{G_1(S)}{T(S)}$$

$$= \frac{1}{[1 + G_1(S)H(S)]^2} \times \frac{\cancel{G_1(S)}}{\cancel{G_1(S)}}$$

$$\therefore \text{Sensitivity} = \frac{1}{1 + G_1(S)H(S)}$$

Controlled variable and Manipulated Variable-

→ A Control System is an interconnection of components that gives the desired response.

→ The primary objective of any control system is to maintain the output of system to

desired value.

→ The output variable to be regulated is called as Controlled Variable.

→ The desired value of the Controlled variable as is called as reference variable.

→ The system has input variables which can be manipulated to modify the controlled

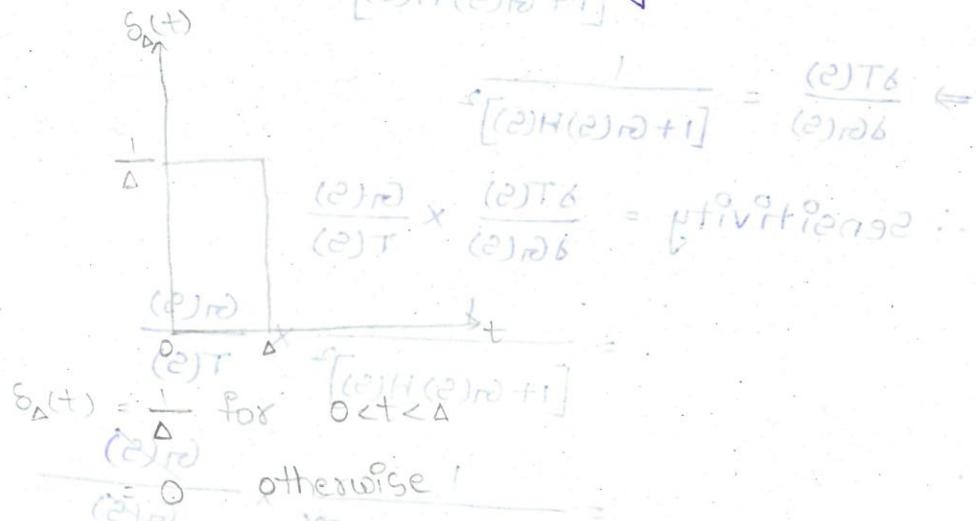
as Manipulated Variables.

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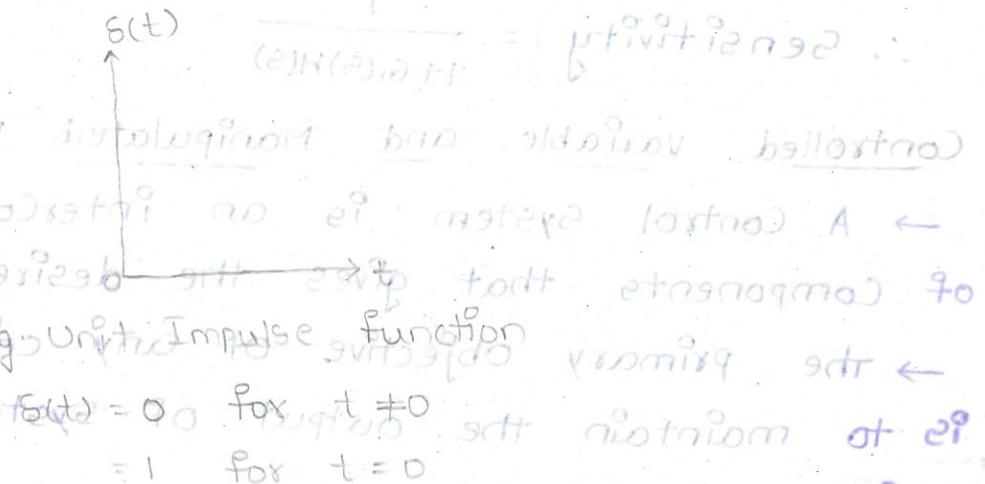
Mathematical methods of Control Systems

i) Impulse Response Method:

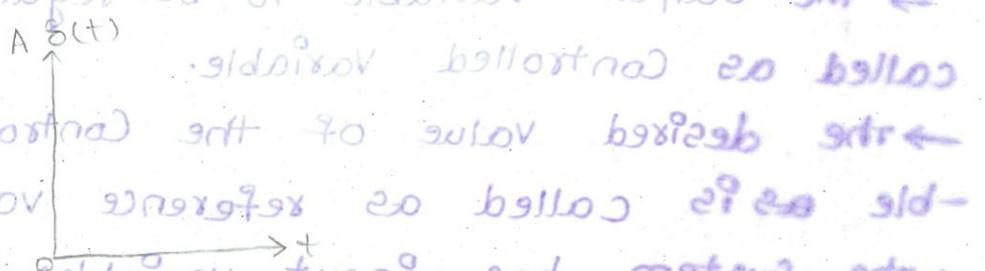
Let us Consider the pulse signal



If $\Delta \rightarrow 0$ then pulse signal is said to be unit impulse $\delta(t)$.



Impulse function with amplitude A is given as



The shifted impulse signal given as

$$\int_{t_0}^{\infty} s(t-t_0) \delta(t-t_0) dt = (2) V$$

if the time between t_0 & t is t , then

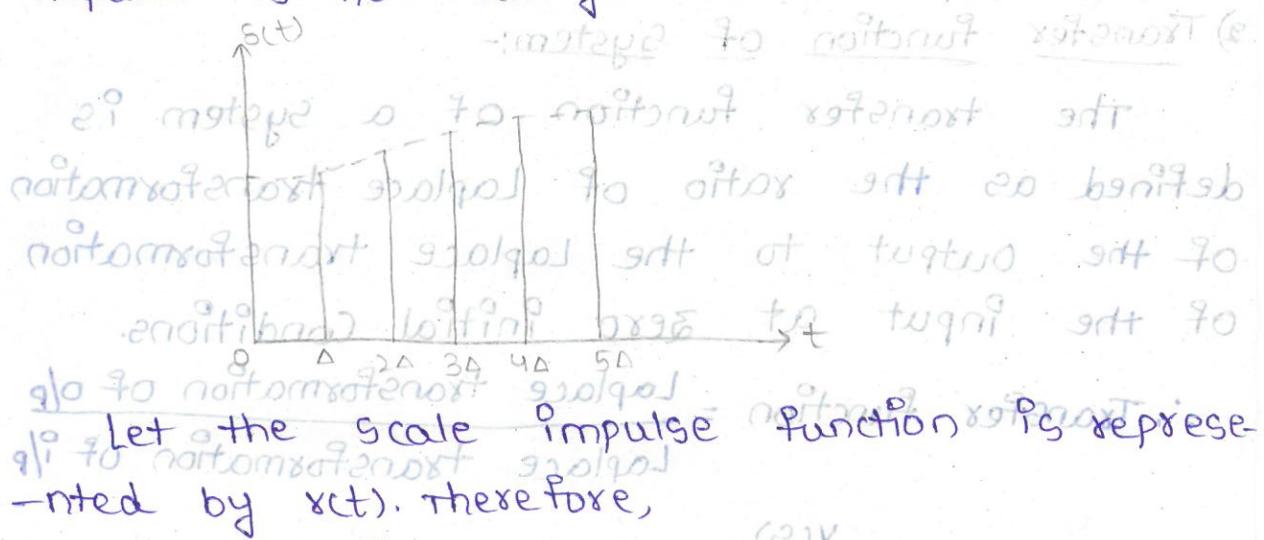
$$\int_{t_0}^{\infty} s(t-t_0) \delta(t-t_0) dt = (2) V$$

$$s(t-t_0) = 0 \text{ for } t \neq t_0$$

$$\int_{t_0}^{\infty} s(t-t_0) \delta(t-t_0) dt = (2) V$$

Scale Impulse Function

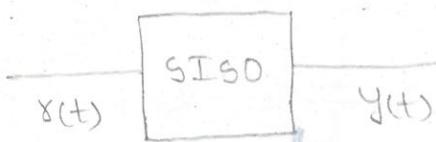
A Impulse function is said to be Scale impulse function when the magnitude of the impulse is not unity.



Let the scale impulse function is represented by $s(t)$. therefore,

$$x(t) = \int_0^{\infty} s(t') s(t-t') dt' \quad (2) V \quad (2) n$$

consider an SISO system no feedback



The output $y(t)$ is the transfer function of input signal $x(t)$.

$$\therefore y(t) = T[x(t)] \quad (2) V + (t) V + (t) V = (t) V$$

Applying Laplace transformation, we get the eqn

$$Y(s) = \int_0^{\infty} \left[\int_0^{\infty} x(\tau) s(t-\tau) d\tau \right] e^{-st} dt$$

For simplicity, s is represented with g

$$\Rightarrow Y(s) = \int_0^{\infty} \left[\int_0^{\infty} x(\tau) g(t-\tau) d\tau \right] e^{-gt} dt$$

$$\Rightarrow Y(s) = \int_0^{\infty} \left[\int_0^{\infty} g(t-\tau) e^{-s(t-\tau)} dt \right] e^{-sr} x(\tau) d\tau$$

$$\Rightarrow Y(s) = \int_0^{\infty} g(t-\tau) e^{-s(t-\tau)} dt \int_0^{\infty} x(\tau) e^{-sr} d\tau$$

$$\Rightarrow Y(s) = G(s) X(s)$$

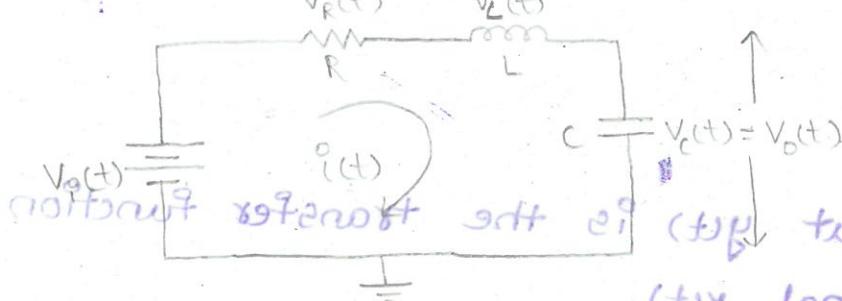
2) Transfer function of System:-

The transfer function of a system is defined as the ratio of Laplace transformation of the output to the Laplace transformation of the input at zero initial conditions.

Transfer function = $\frac{\text{Laplace transformation of o/p}}{\text{Laplace transformation of i/p}}$

$$\Rightarrow G(s) = \frac{Y(s)}{X(s)}$$

Consider an RLC circuit, i.e. no capacitor



Apply KVL to the above circuit,

$$V_o(t) = V_c(t) + V_i(t) + V_R(t) \quad \therefore$$

Applying Laplace transformation, we get

$$\Rightarrow V_o(s) = RI(s) + LS\left(\frac{V_o(s)}{R+LS}\right) + \frac{1}{CS}I(s) + (t)^0 V = (t)^0 V$$

$$\Rightarrow V_o(s) = I(s) \left[R + LS + \frac{1}{CS} \right] + (t)^0 V = (t)^0 V$$

The output of the circuit is calculated across the capacitor

$$\therefore V_o(t) = V_c(t) = \frac{1}{C} \int i(t) dt$$

Applying Laplace transformation, we get

$$\Rightarrow V_o(s) = \frac{1}{CS} I(s)$$

Therefore, the transfer function

$$G(s) = \frac{V_o(s)}{V_p(s)}$$

$$\frac{1}{CS} I(s) \quad (t)^0 \frac{1}{S} = (t)^0 V \frac{1}{FB}$$

$$G(s) = \frac{\frac{1}{CS} I(s)}{I(s) \left[R + LS + \frac{1}{CS} \right]}$$

$$G(s) = \frac{\frac{1}{CS}}{R + LS + \frac{1}{CS}} = \frac{(t)^0 V}{(t)^0}$$

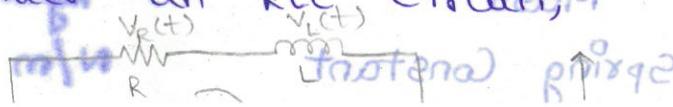
$$G(s) = \frac{1}{RCS + LCS^2 + 1}$$

3) State Variable method:-

It says that the n^{th} order differential equation can be divided into n 1st order differential equations.

The number of 1st order differential equations division will be equal to number of storage elements available in the circuit.

Consider an RLC Circuit,



Apply KVL to the above circuit for voltage

$$V_o(t) = V_R(t) + V_L(t) \Rightarrow V_o(t) = RI + L\frac{di}{dt} = (2)_o V \Leftarrow$$

$$V_o(t) = R_i(t) + \frac{1}{L} \frac{d}{dt} [i(t)] + \frac{1}{C} \int i(t) dt = (2)_o V \Leftarrow$$

$$\Rightarrow \frac{V_o(t) - R_i(t) - \frac{1}{C} \int i(t) dt}{L} = \frac{di(t)}{dt} \text{ (for two sides)}$$

$$\Rightarrow \frac{d}{dt} i(t) = \frac{V_o(t)}{L} - \frac{R_i(t)}{L} - \frac{1}{LC} \int i(t) dt \quad \text{--- (1)}$$

$$\Rightarrow \frac{d}{dt} i(t) = \frac{V_o(t)}{L} - \frac{R_i(t)}{L} - \frac{1}{L} V_c(t) \quad \text{--- (2)}$$

From the circuit, we have $(2)_o I \frac{1}{C} = (2)_o V \Leftarrow$

$$V_o(t) = V_c(t) + \frac{1}{C} \int i(t) dt \text{ (for two sides)}$$

Applying differentiation on b.s $\frac{d}{dt} V_o(t) = (2)_o V \Leftarrow$

$$\frac{d}{dt} V_o(t) = \frac{1}{C} i(t) \quad \text{--- (2)}$$

Eq - (1) & (2) can be solved by using matrix method

$$\begin{bmatrix} V_o(t) \\ i(t) \end{bmatrix} = \begin{bmatrix} 0 & \frac{1}{C} \\ \frac{1}{L} & -\frac{R}{L} \end{bmatrix} \begin{bmatrix} V_o(t) \\ i(t) \end{bmatrix} + \begin{bmatrix} 0 \\ \frac{1}{L} \end{bmatrix} V_c(t)$$

Mathematical models of Control Systems:

I) Mechanical Translation System:-

Parameters Definitions Units

1) x Displacement meters

2) v Velocity meters/sec.

$v = \frac{dx}{dt}$ Acceleration meters/sec²

3) $a = \frac{dv}{dt} = \frac{d^2x}{dt^2}$ Spring constant N/m

4) M Mass kg

5) K Spring Constant N/m

7) f_M	Opposing force due to mass & position	Newton
8) f_B	Opposing force due to friction	Newton
9) f_K	Opposing force due to spring	Newton
10) F	Force	Newton

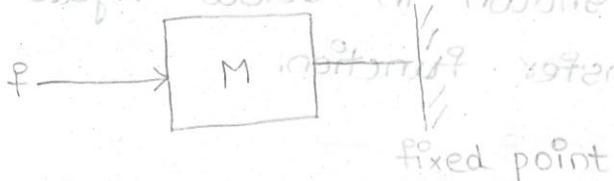
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(Q1 of 2013) principle

A mechanical translation system consists of three elements i.e. a mass, dashpot & spring.

- 1) Mass
- 2) Dashpot
- 3) Spring

Mass: weight of mechanical translation system is represented by mass element. When a force is applied to given mass element, then opposite force f_M develops and it is directly proportional to acceleration.



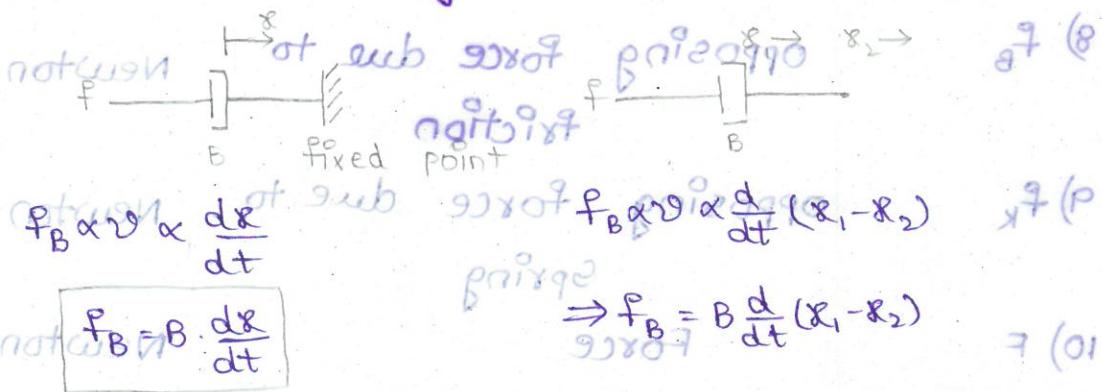
$$f_M \propto a$$

$$\Rightarrow f_M \propto \frac{dv}{dt} \propto \frac{d^2x}{dt^2}$$

$$\Rightarrow f_M = M \frac{d^2x}{dt^2}$$

Dashpot: Friction existing on mechanical translati-

then opposite force f_k develops and it is directly proportional to velocity.



Spring: (Elastic) Velocity

To elastic deformation of the body is represented by spring. When a force is applied to given spring element then opposite force f_k develops which is directly proportional to displacement.

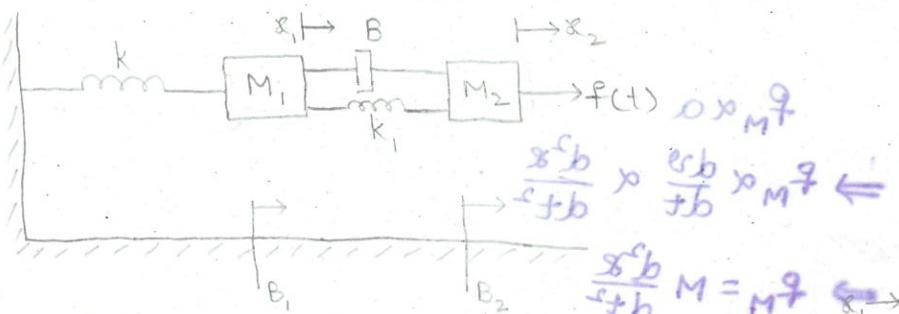
note: $f_k \propto x$ i.e. force is proportional to displacement

$\Rightarrow f_k = kx$ i.e. force is proportional to displacement

Principle of superposition

Problems solved:

- Write the differential equations governing mechanical system shown in below figure and determine its transfer function.



Sol: Free body diagram of M_1

Free body diagram of M_2



From Newton's second law, we have $\ddot{x} + 2\dot{x} + f_2(M) = 0$

$$f_{M_1} + f_k + f_B + f_B + f_k = 0 \quad (2)$$

$$M_1 \frac{d^2 x_1}{dt^2} + k x_1 + B_1 \frac{dx_1}{dt} + B \frac{d}{dt}[x_1 - x_2] + k_1 [x_1 - x_2] = 0$$

Applying Laplace transformation

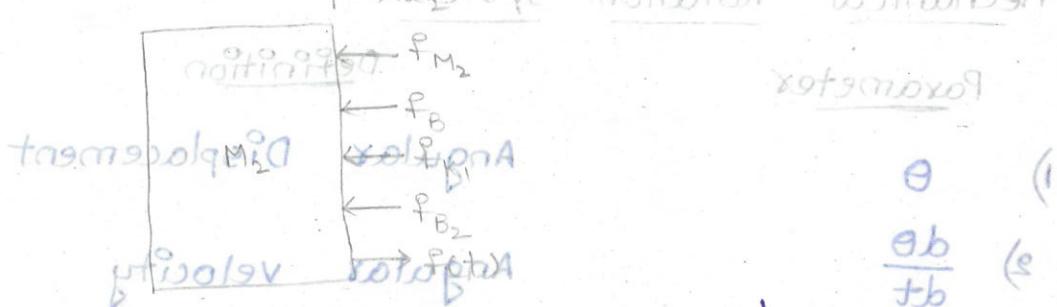
$$M_1 s^2 x_1(s) + k x_1(s) + B_1 s x_1(s) + B s [x_1(s) - x_2(s)]$$

$$[e_{1,1} + 2s] - [(1+s) + (s+k_1)] [x_1(s) - x_2(s)] = 0 \quad (2)$$

$$\Rightarrow x_1(s) [M_1 s^2 + k + B_1 s + B s + k_1] - x_2(s) [B s + k_1] = 0$$

$$\Rightarrow x_1(s) = \frac{[B s + k_1] x_2(s) + (s+2s) s + f_2(M)}{(e_{1,1} + 2s) M_1 s^2 + s(B_1 + B) + (k+k_1)} \quad (2)$$

Free body diagram of M_2



From Newton's second law, we have

$$f_{M_2} + f_B + f_{k_1} + f_{B_2} = f(t)$$

$$\Rightarrow M_2 \frac{d^2 x_2}{dt^2} + B_1 \frac{d}{dt}[x_2 - x_1] + k_1 [x_2 - x_1] + B_2 \frac{d x_2}{dt} = f(t) \quad (4)$$

Applying Laplace transformation

$$M_2 s^2 x_2(s) + B s [x_2(s) - x_1(s)] + k_1 [x_2(s) - x_1(s)] + B_2 s x_2(s) = F(s) \quad (5)$$

$$\Rightarrow x_2(s) [M_2 s^2 + B s + k_1 + B_2 s] - x_1(s) [B s + k_1] = F(s) \quad (5)$$

Substitute $x_1(s)$ value in above equation

$$x_2(s) [M_2 s^2 + B s + k_1 + B_2 s] - \frac{[B s + k_1] x_2(s) + (s+2s) s + f_2(M)}{M_2 s^2 + s(B_1 + B) + (k+k_1)} = F(s) \quad (5)$$

$$X_2(S) [M_2 S^2 + BS + k_1 + B_2 S] [M_1 S^2 + S(B_1 + B) + (k+k_1)] \rightarrow [BS+k_1]^2 X_2(S)$$

$$M_1 S^2 + S(B_1 + B) + (k+k_1) \overset{?}{=} M_1 S^2 + S(B_1 + B) + (k+k_1)$$

$$0 = [S8 - 18]x + [S8 - 18] \frac{b}{fb} s + \frac{18b}{fb} M \Rightarrow F(S) + 28x + \frac{18b}{fb} M$$

$$X_2(S) [M_2 S^2 + BS + k_1 + B_2 S] [M_1 S^2 + S(B_1 + B) + (k+k_1)] - [BS+k_1]^2 X_2(S)$$

$$[c_2 x - (c_2)x] 28 + \Rightarrow F(S) [M_1 S^2 + S(B_1 + B) + (k+k_1)]$$

$$X_2(S) \left[[M_2 S^2 + BS + k_1 + B_2 S] [M_1 S^2 + S(B_1 + B) + (k+k_1)] - (BS+k_1)^2 \right]$$

$$0 = [1x + 28] \Rightarrow F(S) [M_1 S^2 + S(B_1 + B) + (k+k_1)] (c_2)x \Leftarrow$$

$$\frac{X_2(S)}{F(S)} = \frac{M_1 S^2 + S(B_1 + B) + (k+k_1)}{[M_2 S^2 + BS + k_1 + B_2 S] [M_1 S^2 + S(B_1 + B) + (k+k_1)] - (BS+k_1)^2}$$

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Mechanical Rotation System:-

<u>Parameter</u>	<u>Definition</u>
1) θ	Angular Displacement
2) $\frac{d\theta}{dt}$	Angular velocity
3) $\frac{d^2\theta}{dt^2}$	Angular Acceleration
4) T	Applied Torque
5) J	Moment of Inertia
6) B	Dashpot
7) k	Spring Constant

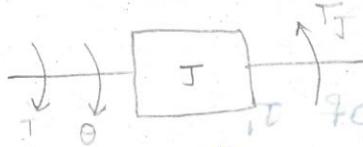
Mechanical + rotation System consists of three basic elements in series (c_2)x substituted

- 1) Moment of Inertia (J) $- [c_2 x + x + c_2 x + c_2 M] (c_2)x$
- 2) Dash-Pot (B) $+ (B + \theta) x + c_2 M$

Spring Constant

Moment of Inertia (J):-

when a torque T is applied to moment of inertia with some angular displacement θ , then an opposing torque T_J is developed and it is proportional to angular acceleration.

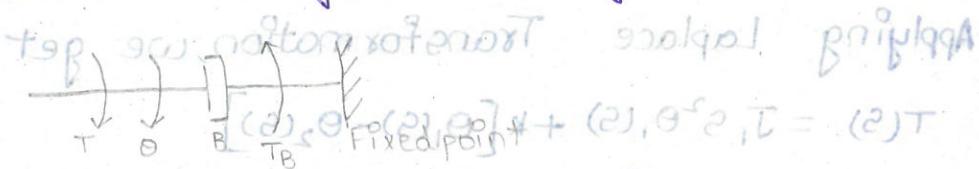


$$T_J \propto \frac{d^2\theta}{dt^2}$$

$$T_J = J \frac{d^2\theta}{dt^2}$$

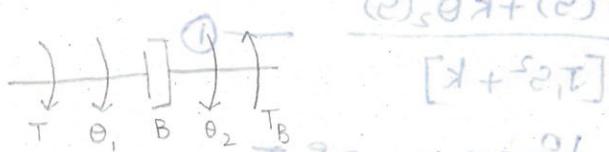
Dash-pot (B):-

when a torque T is applied to given dash-pot then, an opposite torque T_B is developed which is proportionate to angular velocity.



$$T_B \propto \frac{d\theta}{dt} \Rightarrow T_B = B \frac{d\theta}{dt}, \quad (2), \theta = (2)T$$

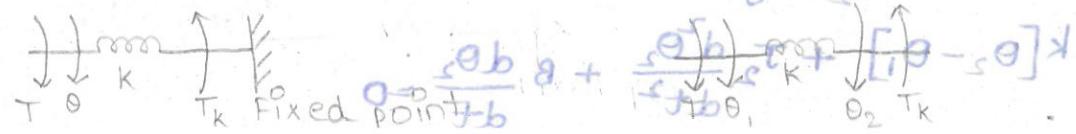
$$\frac{(2), \theta + (2)T}{(2), \theta + (2)T} = (2), \theta \Leftarrow$$



$$T_B \propto \frac{d}{dt} [\theta_1 - \theta_2] \Rightarrow T_B = B \frac{d}{dt} [\theta_1 - \theta_2]$$

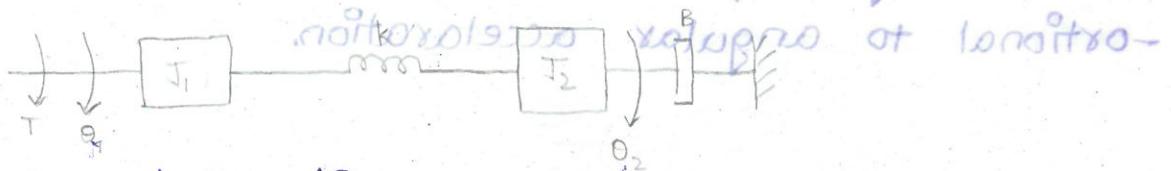
Spring Constant (k):-

when a torque T is applied to given spring, then an opposite torque T_k is developed which is proportionate to angular displacement.

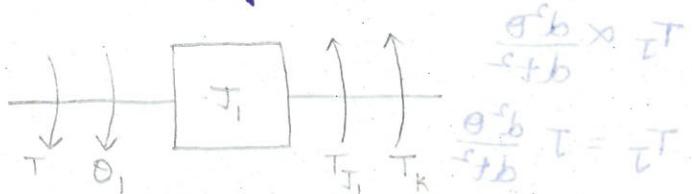


Problems

i) write the differential equations governing the mechanical rotation of System shown in figure and find its transfer function.



Sol: Free body diagram of J_1 ,



i) From Newton's Second Law

$$T = T_{J_1} + T_K$$

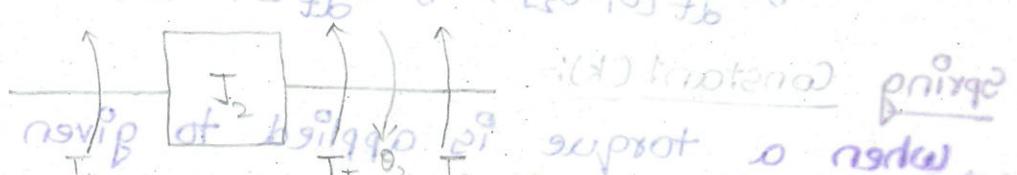
$$T = J_1 \frac{d^2\theta_1}{dt^2} + k[\theta_1 - \theta_2]$$

Applying Laplace Transformation, we get

$$T(s) = J_1 s^2 \theta_1(s) + k[\theta_1(s) - \theta_2(s)]$$

$$\Rightarrow \theta_1(s) = \frac{T(s) + k\theta_2(s)}{[J_1 s^2 + k]} \quad \text{--- (1)}$$

Free body diagram of J_2



i) From Newton's Second Law

$$T_K + T_{J_2} + T_B = 0$$

$$k[\theta_2 - \theta_1] + J_2 \frac{d^2\theta_2}{dt^2} + B \frac{d\theta_2}{dt} = 0$$

$$K[\Theta_2(s) - \Theta_1(s)] + J_2 s^2 \Theta_2(s) + BS \Theta_2(s) = 0$$

$$\Theta_2(s)[K + J_2 s^2 + BS] - K \Theta_1(s) = 0$$

$$\Theta_2(s) = \frac{K \Theta_1(s)}{K + J_2 s^2 + BS}$$

$$\Theta_2(s) = \frac{K \Theta_1(s)}{J_2 s^2 + BS + K}$$

From ① and see now because $\Theta_2(s)$ more

$$\Theta_2(s) = \frac{K}{[J_2 s^2 + BS + K]} \times \frac{T(s) + K \Theta_2(s)}{J_1 s^2 + K} = \frac{K T(s) + K^2 \Theta_2(s)}{J_1 s^2 + K + J_2 s^2 + BS + K}$$

$$\Theta_2(s) [J_2 s^2 + BS + K][J_1 s^2 + K] = K T(s) + K^2 \Theta_2(s)$$

$$\Theta_2(s) \{ [J_2 s^2 + BS + K][J_1 s^2 + K] - K^2 \} = K T(s)$$

$$\frac{\Theta_2(s)}{T(s)} = \frac{K}{[(J_2 s^2 + BS + K)(J_1 s^2 + K) - K^2]} = (2), V$$

$$\frac{\Theta_2(s)}{T(s)} = \frac{K}{J_1 J_2 s^4 + J_1 B s^3 + J_1 K s^2 + K J_2 s^2 + K B s + K - K^2}$$

$$\frac{\Theta_2(s)}{T(s)} = \frac{K}{J_1 J_2 s^4 + J_1 B s^3 + [J_1 + J_2] K s^2 + B K s}$$

27/6/17 now because $\Theta_2(s)$ more

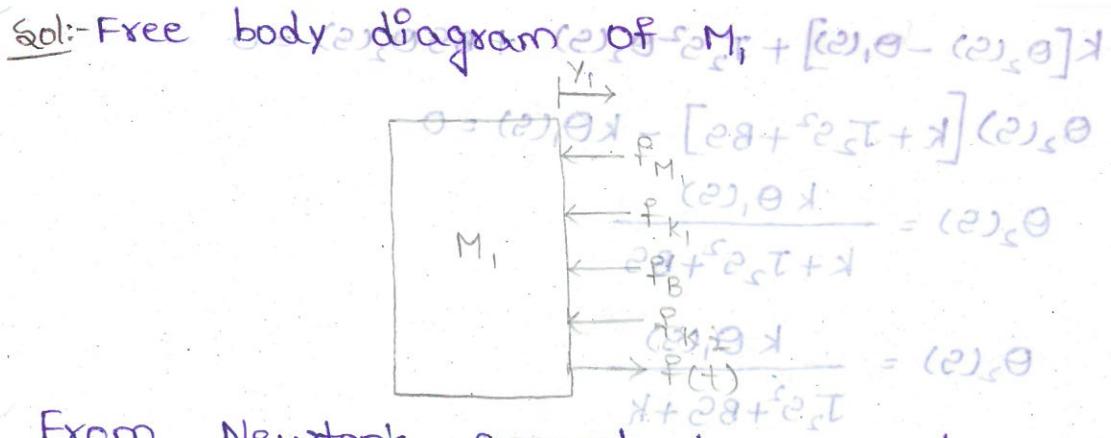
2) Determine the transfer function of the given system.

$$f(t) \rightarrow B \rightarrow M$$

$$O = [(2), X - (2), sV] J_1 + (2), X - (2), sM$$

$$O = (2), Y \rightarrow M = [sI + (2), sM] (2), Y$$

$$O = \left[\frac{(2), sY + (2), Y}{sI + (2), sM} \right] k_2 - [sI + (2), sM] (2), Y \leftarrow$$



From Newton's second law, we have $\text{for } M_1$

$$f(t) = f_{M_1} + f_{k_1} + f_B + f_{k_2}$$

$$f(t) = M_1 \frac{d^2 y_1}{dt^2} + k_1 y_1 + B \frac{dy_1}{dt} + k_2 [y_1 - y_2] = (e_2, \theta)$$

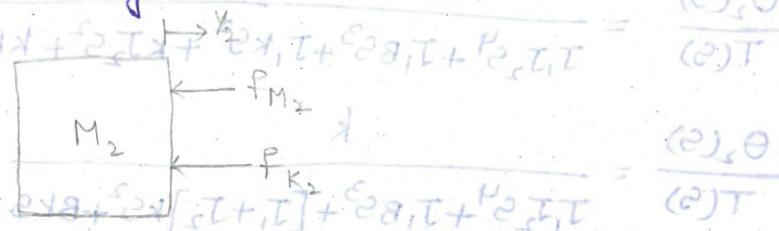
Applying Laplace transformation, we get

$$F(s) = M_1 s^2 Y_1(s) + k_1 Y_1(s) + BS Y_1(s) + k_2 [Y_1(s) - Y_2(s)]$$

$$F(s) = Y_1(s) [M_1 s^2 + k_1 + BS + k_2] - k_2 Y_2(s)$$

$$Y_1(s) = \frac{F(s) + k_2 Y_2(s)}{M_1 s^2 + BS + k_1 + k_2} \quad \text{for } ①$$

Free body diagram of M_2



From Newton's second law, we have for M_2

$$f_{M_2} + f_{k_2} = 0$$

$$M_2 \frac{d^2 y_2}{dt^2} + k_2 [y_2 - y_1] = 0$$

Applying Laplace transformation, we get

$$M_2 s^2 Y_2(s) + k_2 [Y_2(s) - Y_1(s)] = 0$$

$$Y_2(s) [M_2 s^2 + k_2] - k_2 Y_1(s) = 0$$

From ①

$$\Rightarrow Y_2(s) [M_2 s^2 + k_2] - k_2 \left[\frac{F(s) + k_2 Y_2(s)}{M_1 s^2 + BS + k_1 + k_2} \right] = 0$$

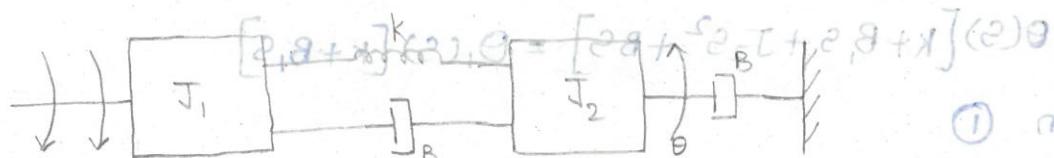
$$\Rightarrow Y_2(s) [M_2 s^2 + K_2] [M_1 s^2 + B s + K_1 + K_2] + K_2 Y(s) = K_2 F(s)$$

∴ Transfer Function

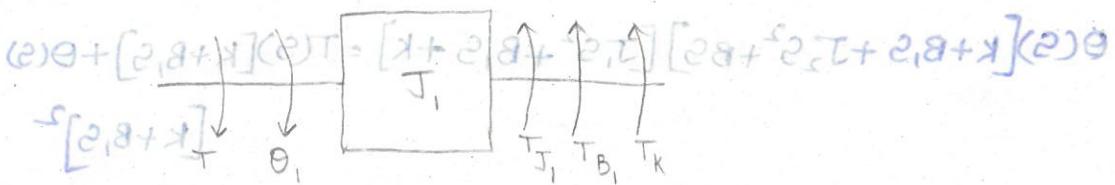
$$\frac{Y_2(s)}{F(s)} = \frac{K_2}{[M_2 s^2 + K_2] [M_1 s^2 + B s + K_1 + K_2] - K_2^2}$$

$$\Rightarrow \frac{Y_2(s)}{F(s)} = \frac{K_2}{M_1 M_2 s^4 + B M_2 s^3 + K_1 M_2 s^2 + K_2 M_2 s^2 + K_2 M_1 s^2 + B K_2 s + K_1 K_2 + K_2^2 - K_2^2}$$

3) Write the differential equations governing mechanical rotation system and find its transfer function.



Sol:- Free body diagram of J_1



From Newton's second law, we have

$$\begin{aligned} T &= T_{J_1} + T_{B_1} + T_K \\ \Rightarrow T &= J_1 \frac{d^2 \theta_1}{dt^2} + B_1 \frac{d}{dt} [\theta_1 - \theta_2] + k [\theta_1 - \theta_2] \end{aligned}$$

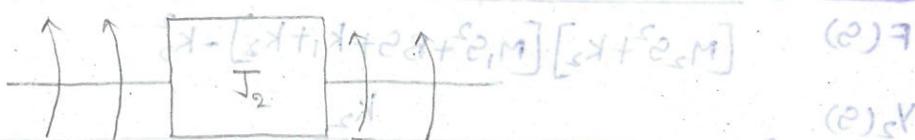
Applying Laplace Transformation, we get

$$T(s) = J_1 s^2 \theta_1(s) + B_1 s [\theta_1(s) - \theta_2(s)] + k [\theta_1(s) - \theta_2(s)]$$

$$\Rightarrow T(s) = J_1 s^2 \theta_1(s) + B_1 s [\theta_1(s) - \theta_2(s)] + k [\theta_1(s) - \theta_2(s)]$$

$$\Theta_1(S) = \frac{T(S) + \Theta(S)[B_1S + K]}{J_1S^2 + B_1S + K} \quad (1)$$

Free body diagram of J_2



From Newton's Second law, we have

$$T_k + T_{B_1} + T_{J_2} + T_B = 0$$

$$\Rightarrow k[\theta - \theta_1] + B_1 \frac{d}{dt} [\theta - \theta_1] + J_2 \frac{d^2 \theta}{dt^2} + B \frac{d\theta}{dt} = 0 \quad (2)$$

Applying Laplace transformation, we get

$$K[\Theta(S) - \Theta_1(S)] + B_1 S [\Theta(S) - \Theta_1(S)] + J_2 S^2 \Theta(S) + B S \Theta(S) = 0$$

$$\Theta(S) [K + B_1 S + J_2 S^2 + B S] - \Theta_1(S) [K + B_1 S] = 0$$

$$\Theta(S) [K + B_1 S + J_2 S^2 + B S] = \Theta_1(S) [K + B_1 S]$$

From (1)

$$\Theta(S) [K + B_1 S + J_2 S^2 + B S] = \frac{T(S) + \Theta(S)[B_1 S + K]}{J_1 S^2 + B_1 S + K} [K + B_1 S]$$

$$\Theta(S) [K + B_1 S + J_2 S^2 + B S] [J_1 S^2 + B_1 S + K] = T(S) [K + B_1 S] + \Theta(S) [K + B_1 S]^2$$

\therefore Transfer function

$$\frac{\Theta(S)}{T(S)} = \frac{k + B_1 S}{[K + B_1 S + J_2 S^2 + B S] [J_1 S^2 + B_1 S + K] - [K + B_1 S]^2}$$

$T = \omega, \theta = \theta_1, \dot{\theta} = \dot{\theta}_1, \ddot{\theta} = \ddot{\theta}_1$

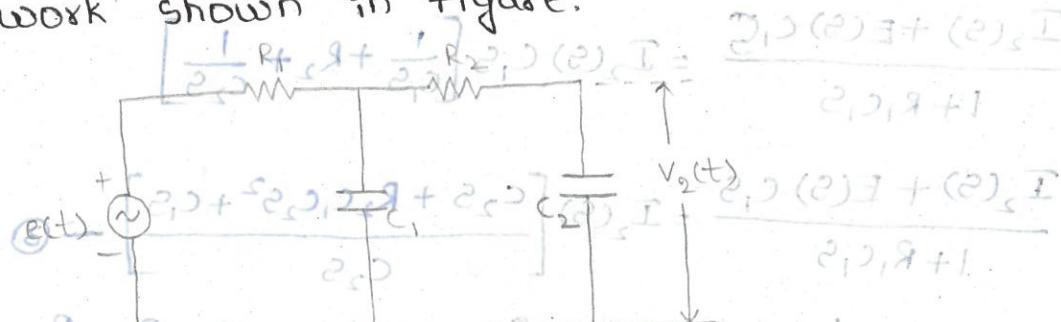
$$[(\ddot{\theta})\theta - (\ddot{\theta}_1)\theta_1] + [(\ddot{\theta})\theta - (\ddot{\theta}_1)\theta_1]^2, \theta + (\ddot{\theta})\theta^2, T = (\ddot{\theta})T$$

$$28/6/17 \quad (2) I \cdot \frac{1}{C_1} = \left[\frac{1}{R_1} + \frac{1}{R_2} + \frac{1}{C_2} \right] (2) I \quad (2) I \leftarrow$$

Electrical Systems:-

① — Problem

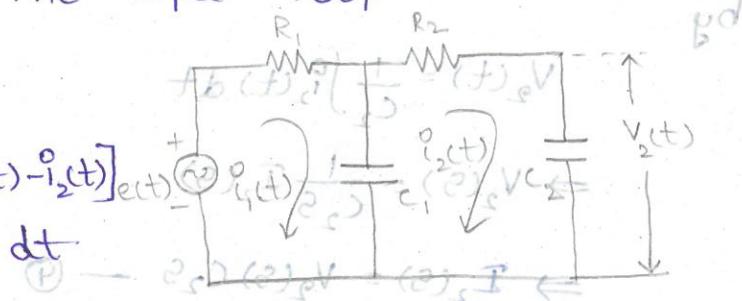
i) obtain the transfer function of electrical network shown in figure.



Sol:- Apply KVL to the input loop.

$$E(t) = I_1(t) \cdot R_1 +$$

$$E(t) = I_1(t) \cdot R_1 + \frac{1}{C_1} \int [I_1(t) - I_2(t)] dt + V_2(t)$$



Applying Laplace Transformation, we get

$$E(s) = R_1 I_1(s) + \frac{1}{C_1 s} [I_1(s) - I_2(s)] \quad (2) \leftarrow$$

$$\Rightarrow E(s) = I_1(s) \left[R_1 + \frac{1}{C_1 s} \right] - \frac{1}{C_1 s} I_2(s) \quad (2) \leftarrow$$

$$\Rightarrow I_1(s) = \frac{E(s) + \frac{1}{C_1 s} I_2(s)}{R_1 + \frac{1}{C_1 s}} \quad (2) \leftarrow$$

$$\Rightarrow I_1(s) = \frac{I_2(s) + E(s) C_1 s}{R_1 + C_1 s} \quad (1) \quad (2) \leftarrow$$

$$\frac{1}{C_2} \int [I_2(t) - I_1(t)] dt + R_2 I_2(t) + \frac{1}{C_2} \int I_2(t) dt = 0$$

Apply KVL to the output loop

$$\frac{1}{C_2} \int [I_2(t) - I_1(t)] dt + R_2 I_2(t) + \frac{1}{C_2} \int I_2(t) dt = 0$$

Applying Laplace Transformation, we get

$$\Rightarrow I_2(s) \left[\frac{1}{C_1 s} + R_2 + \frac{1}{C_2 s} \right] = \frac{1}{C_1 s} \cdot I_1(s)$$

filalss

$$\Rightarrow I_1(s) = I_2(s) C_1 s \left[\frac{1}{C_1 s} + R_2 + \frac{1}{C_2 s} \right] \quad \text{--- ②}$$

eq-① = eq-② without reasrt att niotdo(i)

$$\therefore \frac{I_2(s) + E(s) C_1 s}{1 + R_1 C_1 s} = I_2(s) C_1 s \left[\frac{1}{C_1 s} + R_2 + \frac{1}{C_2 s} \right]$$

$$\frac{I_2(s) + E(s) C_1 s}{1 + R_1 C_1 s} = I_2(s) \left[\frac{C_2 s + R_2 C_1 C_2 s^2 + C_1 s}{C_2 s} \right] \quad \text{--- ③}$$

The output voltage from the circuit is given by

$$V_2(t) = \frac{1}{C_2} \int_{t_0}^t I_2(t) dt$$

$$\Rightarrow V_2(s) = \frac{1}{C_2 s} I_2(s)$$

$$\Rightarrow I_2(s) = V_2(s) C_2 s \quad \text{--- ④}$$

Substitute eq-④ in eq-③

$$\frac{V_2(s) C_2 s + E(s) C_1 s}{1 + R_1 C_1 s} = V_2(s) C_2 s \left[\frac{C_2 s + R_2 C_1 C_2 s^2 + C_1 s}{C_2 s} \right] \quad \text{--- ⑤}$$

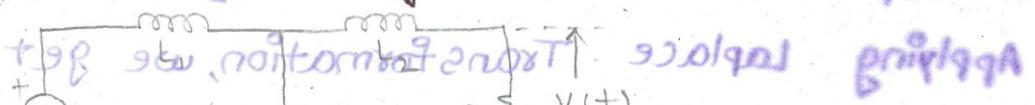
$$\Rightarrow V_2(s) C_2 s + E(s) C_1 s = V_2(s) \left[C_2 s + R_2 C_1 C_2 s^2 + C_1 s \right] [1 + R_1 C_1 s]$$

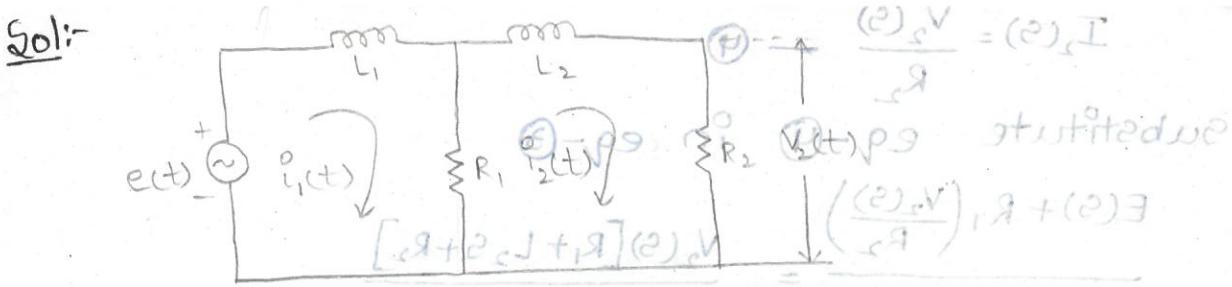
$$\Rightarrow E(s) C_1 s = V_2(s) \left\{ \left[C_2 s + R_2 C_1 C_2 s^2 + C_1 s \right] [1 + R_1 C_1 s] - C_2 s \right\}$$

∴ the transfer function of the given circuit is

$$\frac{V_2(s)}{E(s)} = \frac{C_1 s}{\left[C_2 s + R_2 C_1 C_2 s^2 + C_1 s \right] [1 + R_1 C_1 s] - C_2 s}$$

2) Obtain the transfer function of electrical network shown in figure.





Apply KVL to the input loop

$$e(t) = L_1 \frac{d}{dt} i_1(t) + R_1 [i_1(t) - i_2(t)] \quad \left(\frac{(e)_s V}{s} \right), A + (e) E$$

Applying Laplace Transformation, we get

$$E(s) = L_1 s I_1(s) + R_1 [I_1(s) - I_2(s)] \quad \left[\frac{(e)_s V}{s} \right], A + (e) E$$

$$E(s) = I_1(s) [L_1 s + R_1] - R_1 I_2(s). \quad \left(\frac{(e)_s V}{s} \right), A + (e) E$$

$$\Rightarrow I_1(s) = \frac{E(s) + R_1 I_2(s)}{L_1 s + R_1} \quad \left\{ \frac{(e)_s V}{s} \right\}, A + (e) E$$

Apply KVL to the output loop

$$R_1 [i_2(t) - i_1(t)] + L_2 \frac{d}{dt} i_2(t) + R_2 i_2(t) = 0 \quad (e)_s V$$

Applying Laplace Transformation, we get

$$R_1 [I_2(s) - I_1(s)] + L_2 s I_2(s) + R_2 I_2(s) = 0. \quad \text{false}$$

$$I_2(s) [R_1 + L_2 s + R_2] - R_1 I_1(s) = 0 \quad \text{option A}$$

$$I_2(s) [R_1 + L_2 s + R_2] = R_1 I_1(s) \quad \text{option B}$$

$$I_1(s) = \frac{I_2(s) [R_1 + L_2 s + R_2]}{R_1} \quad \text{option C}$$

$$\therefore \text{eq-1} = \text{eq-2} \quad \text{option D}$$

$$\frac{E(s) + R_1 I_2(s)}{L_1 s + R_1} = \frac{I_2(s) [R_1 + L_2 s + R_2]}{R_1} \quad \text{option E}$$

The output voltage from the circuit is given by.

$$V_2(t) = R_2 i_2(t) \quad \text{option F}$$

$$I_2(s) = \frac{V_2(s)}{R_2} - \textcircled{4}$$

Substitute eq-④ in eq-③

$$\frac{E(s) + R_1 \left(\frac{V_2(s)}{R_2} \right)}{L_1 s + R_1} = \frac{V_2(s) [R_1 + L_2 s + R_2]}{R_1 R_2}$$

$$\frac{E(s) + R_1 \left(\frac{V_2(s)}{R_2} \right)}{L_1 s + R_1} = [(t)_c^o - (t)_s^o] R + (t)_s^o \frac{R}{R_1} s + (t)_s^o = (t)_s^o$$

$$\frac{E(s) + R_1 \left(\frac{V_2(s)}{R_2} \right)}{L_1 s + R_1} = V_2(s) \left[\frac{1}{R_2} + \frac{L_2 s}{R_1 R_2} + \frac{1}{R_1} \right]$$

$$\frac{E(s) + R_1 V_2(s)}{R_1 + R_2} = V_2(s) \left[\frac{1}{R_2} + \frac{L_2 s}{R_1 R_2} + \frac{1}{R_1} \right] \left[\frac{L_1 s + R_1}{L_1 s + R_1} \right]$$

$E(s) \neq R_1 R_2$

$$E(s) = V_2(s) \left\{ \left[\frac{1}{R_2} + \frac{L_2 s}{R_1 R_2} + \frac{1}{R_1} \right] \left[\frac{L_1 s + R_1}{L_1 s + R_1} \right] - \frac{R_1}{R_1 R_2} \right\}$$

∴ the transfer function of the given circuit is.

$$\frac{V_2(s)}{E(s)} = \frac{0 = (t)_c^o R + (t)_s^o \frac{R}{R_1} s + [(t)_s^o - (t)_c^o] R}{\left[\frac{1}{R_2} + \frac{L_2 s}{R_1 R_2} + \frac{1}{R_1} \right] \left[\frac{L_1 s + R_1}{L_1 s + R_1} \right] - \frac{R_1}{R_1 R_2}}$$

29/6/17 $0 = (e)_c I_c R + (e)_s I_s R + [(e)_s - (e)_c] R$

Analogous Systems $(e)_c = (e)_s I_c R - [R + e_s I_s + R] (e)_s I$

The systems with identical mathematical expressions are known as Analogous Systems.

i) Electrical Analogous Systems with Mechanical Translation Systems.

$$(e)_c - p e = (e)_s - p e$$

a) Force Voltage Analogy $[R + e_s I_s + R] (e)_s I = (e)_s I_c R + (e)_c E$

b) Force Current Analogy

ii) Electrical Systems with Mechanical Rotation Systems.

a) Torque Voltage Analogy

$$(t)_c^o R = (t)_s^o V$$

Force voltage Analogy:-

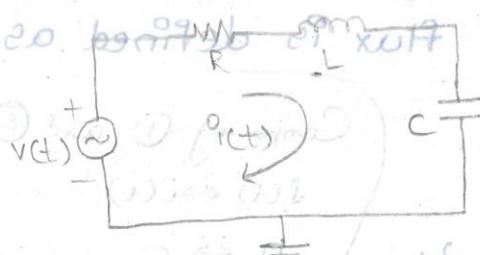
Consider a mechanical Translation System



$$f(t) = F_M + F_B + F_K$$

$$f(t) = M \frac{d^2x}{dt^2} + B \frac{dx}{dt} + kx \quad \text{--- (1)}$$

Consider Series RLC Circuit



Replace displacement

with velocity

$$\text{i.e. } \frac{d^2x}{dt^2} = \frac{dv}{dt}$$

$$\frac{dx}{dt} = v$$

Apply KVL, we get

$$V(t) = Rv(t) + L \frac{di(t)}{dt} + \frac{1}{C} \int i(t) dt \quad \text{--- (2)}$$

we know that, rate of change of charge is

Current

$\phi \leftrightarrow x; \frac{1}{L} \leftrightarrow \frac{1}{C}$ Comparing (1) and (2) we get

$$\therefore i(t) = \frac{dq}{dt}$$

$$f(t) \leftrightarrow v(t) \quad B \leftrightarrow R \quad k \leftrightarrow \frac{1}{C}$$

$$\Rightarrow V(t) = R \frac{dq}{dt} + L \frac{d^2q}{dt^2} + \frac{1}{C} \cdot q$$

$$\Rightarrow V(t) = L \frac{d^2q}{dt^2} + R \frac{dq}{dt} + \frac{1}{C} \cdot q \quad \text{--- (2)}$$

By Comparing (1) & (2), we get

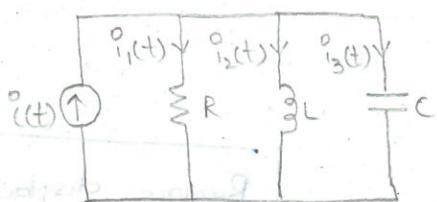
$$M \leftrightarrow L; B \leftrightarrow R; k \leftrightarrow \frac{1}{C}; x \leftrightarrow q \quad T + \theta b + \frac{\theta^2 b}{f b} T = T$$

Force Current Analogy:- $\theta \cdot k + \frac{\theta b}{f b} \theta + \frac{\theta^2 b}{f b} T = T$

Consider a mechanical Translation System



$f(t) = M \frac{d^2\varphi}{dt^2} + B \frac{d\varphi}{dt} + K\varphi$ Replacing displacement with velocity
 Consider shunt RLC circuit.



$$\therefore f(t) = M \frac{dv}{dt} + BV + K \int v dt \quad \text{--- (1)}$$

$$L \frac{d^2i}{dt^2} + B \frac{di}{dt} + \frac{1}{C} i = V(t)$$

$$i(t) = i_1(t) + i_2(t) + i_3(t) \quad \text{--- (2)}$$

$$\Rightarrow i(t) = \frac{V(t)}{R} + \frac{1}{L} \int V(t) dt + C \frac{dv(t)}{dt} \quad \text{Comparing}$$

The rate of change of flux is defined as voltage.

$$\therefore V(t) = \frac{d\phi}{dt}$$

$$\Rightarrow i(t) = \frac{1}{R} \frac{d\phi}{dt} + \frac{1}{L} \cdot \phi + C \cdot \frac{d^2\phi}{dt^2} \quad \text{--- (3)}$$

$$\Rightarrow i(t) = C \frac{d^2\phi}{dt^2} + \frac{1}{R} \frac{d\phi}{dt} + \frac{1}{L} \phi \quad \text{--- (4)}$$

By comparing (1) & (2), we get,

$$M \leftrightarrow C; B \leftrightarrow \frac{1}{R}; K \leftrightarrow \frac{1}{L}; \varphi \leftrightarrow \phi$$

Torque Voltage Analogy:-

Consider a mechanical Rotational System



$$J \frac{d^2\theta}{dt^2} + B \frac{d\theta}{dt} + K\theta = T_M \quad \text{--- (5)}$$

$$J \frac{d^2\theta}{dt^2} + B \frac{d\theta}{dt} + \frac{1}{C} \theta = T_M \quad \text{--- (6)}$$

$$T = T_J + T_B + T_R \quad \text{--- (7)}$$

$$\Rightarrow T = J \frac{d^2\theta}{dt^2} + B \frac{d\theta}{dt} + K\theta \quad \text{Comparing}$$

Consider Series RLC Circuit



Comparing (1) and (2) we get

$$f(t) \leftrightarrow i(t)$$

$$M \leftrightarrow C$$

$$B \leftrightarrow \frac{1}{R}$$

$$K \leftrightarrow \frac{1}{L}$$

Apply KVL, we get $\phi \frac{1}{L} + \frac{\phi b}{R} + \frac{\phi b}{C} = (t) i$

$$v(t) = R i(t) + L \frac{di(t)}{dt} + \frac{1}{C} \int i(t) dt \quad \text{Eqn ①}$$

we know that ~~rate of change of charge is current~~

$$i(t) = \frac{dq}{dt}$$

$$\Rightarrow v(t) = R \frac{dq}{dt} + L \frac{d^2 q}{dt^2} + \frac{1}{C} \cdot q$$

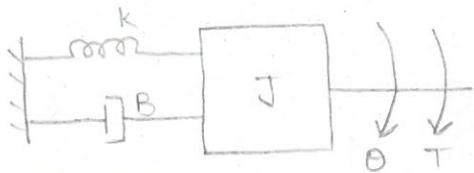
$$\Rightarrow v(t) = L \frac{d^2 q}{dt^2} + R \frac{dq}{dt} + \frac{1}{C} \cdot q \quad \text{Eqn ②}$$

Comparing ① & ②, we get

$$M \leftrightarrow L; B \leftrightarrow R; k \leftrightarrow \frac{1}{C}; \theta \leftrightarrow q.$$

Torque Current Analogy:-

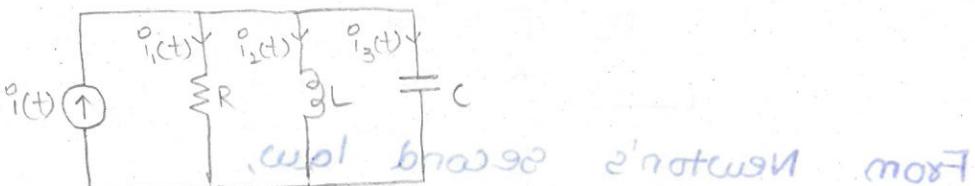
Consider a mechanical Rotational system.



$$T = T_J + T_B + T_K.$$

$$T = J \frac{d^2 \theta}{dt^2} + B \frac{d\theta}{dt} + K \cdot \theta \quad \text{Eqn ①}$$

Consider shunt RLC circuit



$$i(t) = i_1(t) + i_2(t) + i_3(t)$$

$$\Rightarrow i(t) = \frac{v(t)}{R} + \frac{1}{L} \left\{ v(t) dt \right\} + C \frac{dv(t)}{dt} + \frac{1}{C} \cdot v(t)$$

the rate of change of flux is defined as voltage.

$$\therefore v(t) = \frac{d\phi}{dt}$$

$$\Rightarrow i(t) = \frac{1}{L} \frac{d\phi}{dt}, \frac{1}{C} \cdot \phi + R \cdot \frac{d^2 \phi}{dt^2}$$

$$\Rightarrow i(t) = C \frac{d^2\phi}{dt^2} + \frac{1}{R} \frac{d\phi}{dt} + \frac{1}{L} \phi - \textcircled{2}$$

top we have A/A

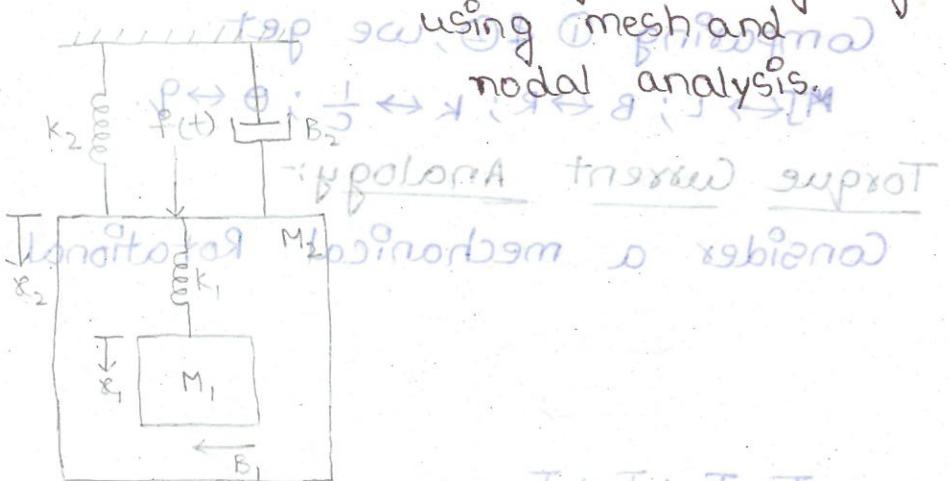
Comparing $\textcircled{1}$ & $\textcircled{2}$ $\left[\frac{1}{C} \right] \frac{1}{2} + \frac{(t)b}{fb} + (t)i = (t)v$

$\Rightarrow C \leftrightarrow R; B \leftrightarrow \frac{1}{L}; t \leftrightarrow \phi$ to fit word we know

3/7/17

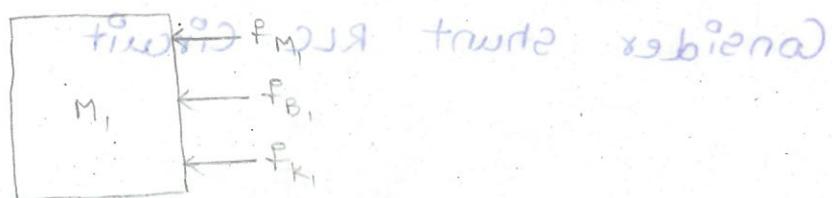
Problems

i) write the differential equations governing the mechanical system shown in figure. Draw the force voltage and force current electrical analogous circuits shown in figure. Verify by using mesh and nodal analysis.



Sol:- Free body diagram of M_1

$$f_{M_1} + f_{B_1} + f_{k_1} = T$$

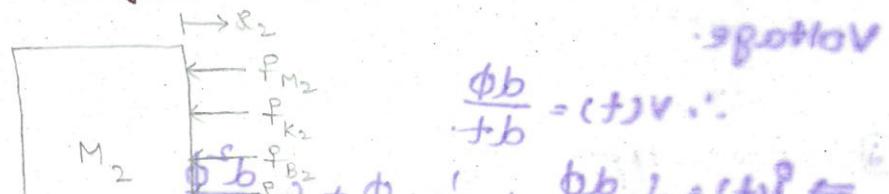
$$f_{M_1} + \frac{\partial f_b}{\partial x_1} + \frac{\partial f_b}{\partial x_2} + \frac{\partial f_b}{\partial \phi} T = T$$


From Newton's second law,

$$f_{M_1} + f_{B_1} + f_{k_1} = 0 \quad (\textcircled{1}) \quad \dot{x}_1 + (\textcircled{2})_1 + (\textcircled{3})_1 = (\textcircled{4})_1$$

$$\Rightarrow M_1 \frac{d^2x_1}{dt^2} + B_1 \frac{dx_1}{dt} + k_1 [x_1 - x_2] = 0 \quad \textcircled{1} \quad \dot{x}_1 + (\textcircled{2})_1 + (\textcircled{3})_1 = (\textcircled{4})_1$$

Free body diagram of M_2 to get str



From Newton's second Law,

$$f(t) = F_{M_2} + F_{B_2} + f_{k_2} + F_{B_1} + f_{k_1}$$

$$\Rightarrow f(t) = M_2 \frac{d^2 \delta_2}{dt^2} + B_2 \frac{d\delta_2}{dt} + k_2 \delta_2 + B_1 \frac{d}{dt} [\delta_2 - \delta_1] + k_1 [\delta_2 - \delta_1] \quad \text{--- (1)}$$

Force Voltage Analogy (or) Force voltage Electri-

-(col) Analogous Circuit $\left[\frac{1}{C_2} + \frac{1}{R_2} + \frac{1}{tb} \right] = (t) \quad \text{--- (2)}$

$$M_1 \leftrightarrow L_1 \quad B_1 \leftrightarrow R_1 \quad k_1 \leftrightarrow \frac{1}{C_1} \quad \delta \leftrightarrow \varphi \quad f(t) = e(t)$$

$$M_2 \leftrightarrow L_2 \quad B_2 \leftrightarrow R_2 \quad k_2 \leftrightarrow \frac{1}{C_2} \quad - \frac{d\delta}{dt} = \frac{d\varphi}{dt} \quad \rho V = (t) \quad \text{--- (3)}$$

+ current equation $\delta = \int (t) dt + C_0$

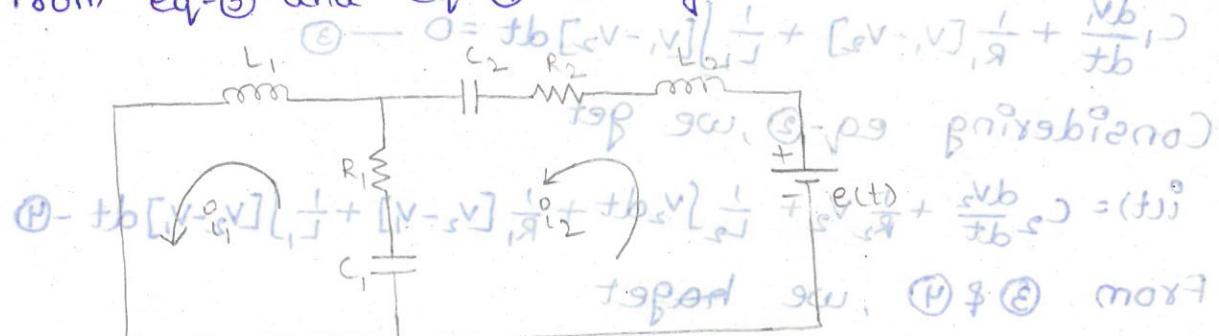
Considering eq-①

$$L_1 \frac{di_1}{dt} + R_1 [i_1 - i_2] + \frac{1}{C_1} \int [i_1 - i_2] dt = 0 \quad \text{--- (4)}$$

Considering eq-②

$$e(t) = L_2 \frac{di_2}{dt} + R_2 i_2 + \frac{1}{C_2} \int i_2 dt + R_1 [i_2 - i_1] + \frac{1}{C_1} \int [i_2 - i_1] dt \quad \text{--- (5)}$$

From eq-③ and eq-④ we get



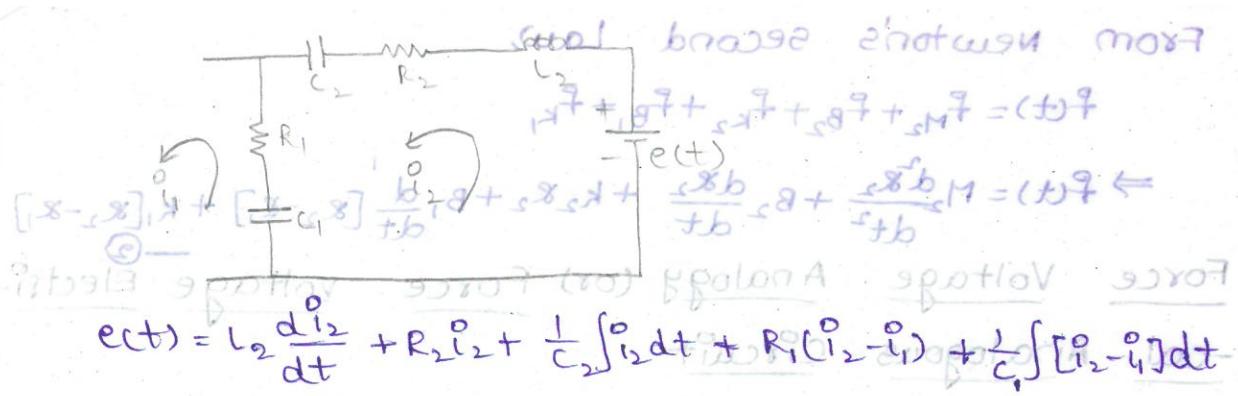
Mesh Analysis: Considering ① loop.



$$① - 0 = tb [ev - v] \left[\frac{1}{i_1} + [ev - v] \frac{1}{R_1} + \frac{vb}{tb} \right] \quad \text{--- (6)}$$

$$L_1 \frac{di_1}{dt} + R_1 [i_1 - i_2] + \frac{1}{C_1} \int [i_1 - i_2] dt = 0 \quad \text{--- (7)}$$

$$② \text{ Considering } \left[\frac{1}{R_2} + \frac{1}{tb} \right] + tb [v - v] \left[\frac{1}{i_2} + vb \frac{1}{R_2} + \frac{vb}{tb} \right] = (t) \quad \text{--- (8)}$$



$$(f) \varphi = (f) \varphi$$

AS $\text{eq-3} = \text{eq-5}$; $\text{eq-4} = \text{eq-6}$ thus Verified.

Force : Current & Electrical Analogous Circuit

$$M_1 \leftrightarrow C_1 \quad B_1 \leftrightarrow \frac{1}{R_1} \quad k_1 \leftrightarrow \frac{1}{L_1} \quad x \leftrightarrow \phi$$

$$M_2 \leftrightarrow C_2 \quad B_2 \leftrightarrow \frac{1}{R_2} \quad k_2 \leftrightarrow \frac{1}{L_2} \quad \frac{d\phi}{dt} = V$$

①-ρσ βηνιχεβιδανα

$$\text{Considering eq-① we get } \frac{dx}{dt} = \frac{d\phi}{dt} = v$$

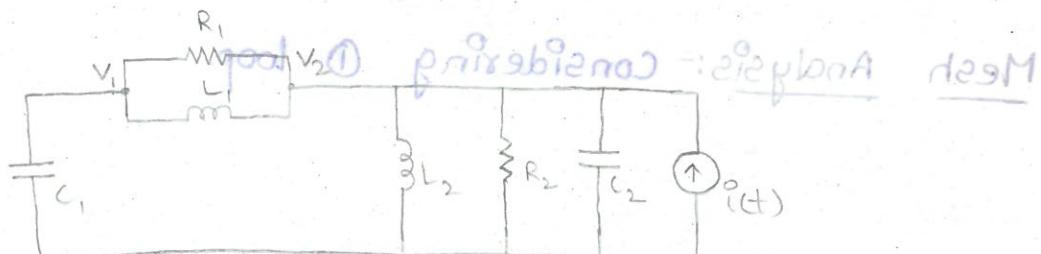
Considering eq-①, we get

$$C_1 \frac{dV_1}{dt} + \frac{1}{R_1} [V_1 - V_2] + \frac{1}{L_1} \int [V_1 - V_2] dt = 0 \quad \text{--- (3)}$$

Considering eq.-②, we get

$$v_2(t) = C_2 \frac{dv_2}{dt} + \frac{1}{R_2} v_2 + \frac{1}{L_2} \int v_2 dt + \frac{1}{R_1} [v_2 - v_1] + \frac{1}{L_1} \int [v_2 - v_1] dt \quad (4)$$

From ③ & ④ , we have



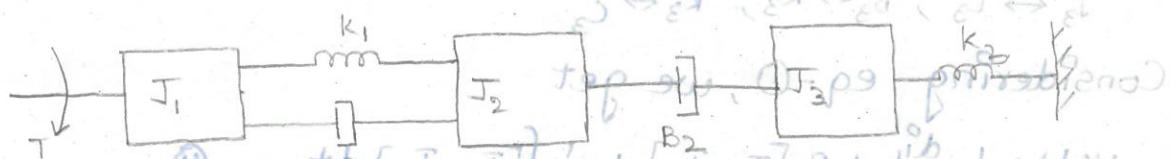
Considering Node V_1 , we get

$$C_1 \frac{dV_1}{dt} + \frac{1}{R_1}[V_1 - V_2] + \frac{1}{L_1} \int [V_1 - V_2] dt = 0. \quad \text{--- (5)}$$

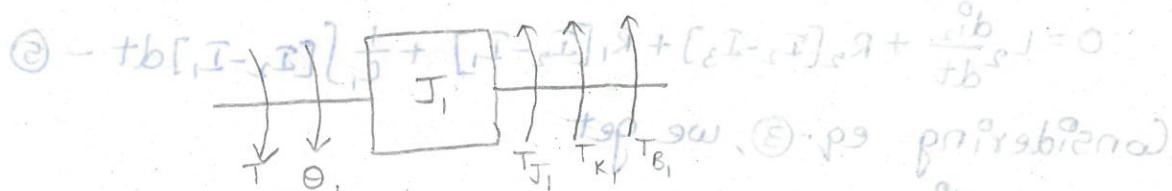
Considering Node v_2 , we get $[e_{ij}]_R + \frac{v_2}{t_h} =$

$$i(t) = C_2 \frac{dV_2}{dt} + \frac{1}{R_2} V_2 + \frac{1}{L_2} \int V_2 dt + \frac{1}{R_1} [V_2 - V_1] + \frac{1}{L_1} \int [V_2 - V_1] dt \quad (6)$$

6/17
 2) write the differential equations governing mechanical rotation system shown in figure. Draw the Torque voltage and Torque Current analog-ous Circuits. $(t)v - T$



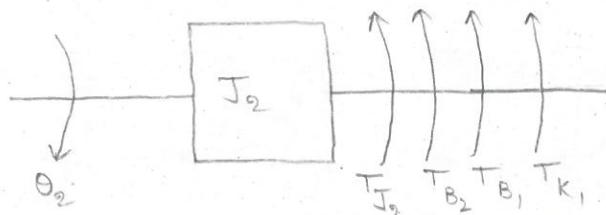
Sol: Free body diagram of J_1 :



$$T = T_{J_1} + T_{B_1} + T_{K_1} - [I_1 - \varepsilon I] \frac{d\theta_1}{dt} + [I_2 - \varepsilon I] \frac{d\theta_1}{dt} + [\varepsilon I - \varepsilon I] \frac{d\theta_1}{dt} + \frac{\varepsilon b}{tb} \varepsilon I = 0$$

$$T = J_1 \frac{d^2\theta_1}{dt^2} + B_1 \frac{d}{dt} [\theta_1 - \theta_2] + K[\theta_1 - \theta_2] \quad \text{--- (1)}$$

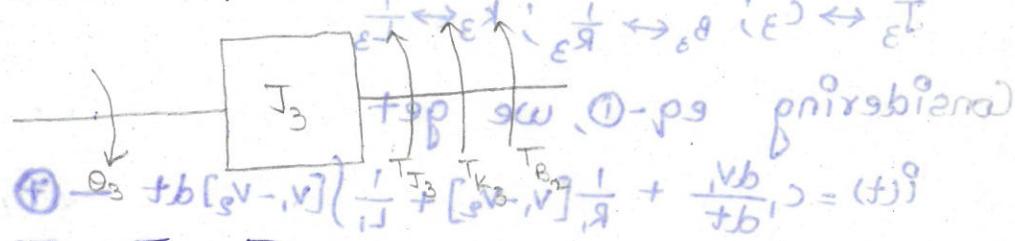
Free body diagram of J_2 :



$$0 = T_{J_2} + T_{B_2} + T_{K_2} + T_{K_1}$$

$$0 = J_2 \frac{d^2\theta_2}{dt^2} + B_2 \frac{d}{dt} [\theta_2 - \theta_3] + B_1 \frac{d}{dt} [\theta_2 - \theta_1] + K_1 [\theta_2 - \theta_1] \quad \text{--- (2)}$$

Free body diagram of J_3 :



$$0 = T_{J_3} + T_{B_3} + T_{K_3} + T_{K_2} + T_{K_1} - [I_3 - \varepsilon I] \frac{d\theta_3}{dt} - [I_2 - \varepsilon I] \frac{d\theta_3}{dt} - [\varepsilon I - \varepsilon I] \frac{d\theta_3}{dt} + \frac{\varepsilon b}{tb} \varepsilon I = 0$$

$$\Theta = J_3 \frac{d^2 \Theta}{dt^2} + K_3 \Theta_3 + B_2 \frac{d}{dt} [\Theta_3 - \Theta_2] \quad \text{--- ③}$$

Torque Voltage Analogous Circuit

$$J_1 \leftrightarrow L_1; B_1 \leftrightarrow R_1; K_1 \leftrightarrow \frac{1}{C_1}; \Theta \leftrightarrow \varphi$$

$$J_2 \leftrightarrow L_2; B_2 \leftrightarrow R_2;$$

$$T = V(t)$$

$$J_3 \leftrightarrow L_3; B_3 \leftrightarrow R_3; K_3 \leftrightarrow \frac{1}{C_3}$$

Considering eq-①, we get

$$V(t) = L_1 \frac{dI_1}{dt} + R_1 [I_1 - I_2] + \frac{1}{C_1} \int [I_1 - I_2] dt \quad \text{--- ④}$$

Considering eq-②, we get

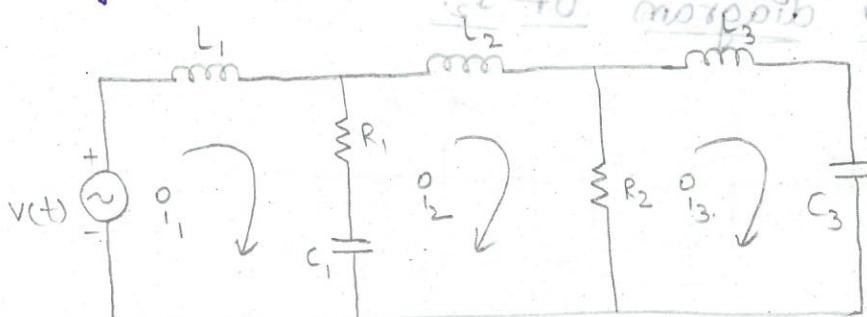
$$\Theta = L_2 \frac{d\Theta_2}{dt} + R_2 [I_2 - I_3] + R_1 [I_2 - I_1] + \frac{1}{C_1} \int [I_2 - I_1] dt \quad \text{--- ⑤}$$

Considering eq-③, we get

$$\Theta = L_3 \frac{d\Theta_3}{dt} + \frac{1}{C_3} \int \Theta_3 dt + R_2 [I_3 - I_2] \quad \text{--- ⑥}$$

By Considering ④, ⑤, ⑥ equations

we get Torque Voltage Analogous Circuit



Torque Current Analogous Circuit $\dot{\Theta} + \frac{1}{L} \dot{I} = 0$

$$④ - J_1 \leftrightarrow C_1; B_1 \leftrightarrow \frac{1}{R_1}; K_1 \leftrightarrow \frac{1}{L_1}; \dot{\Theta}_1 \leftrightarrow \dot{I}_1; \dot{\Theta}_2 \leftrightarrow \dot{I}_2; \dot{\Theta}_3 \leftrightarrow \dot{I}_3; T = I(t); \dot{I} = \dot{I}_1 + \dot{I}_2 + \dot{I}_3 \quad \text{--- ⑦}$$

$$J_2 \leftrightarrow C_2; B_2 \leftrightarrow \frac{1}{R_2};$$

$$J_3 \leftrightarrow C_3; B_3 \leftrightarrow \frac{1}{R_3}; K_3 \leftrightarrow \frac{1}{L_3}$$

Considering eq-①, we get

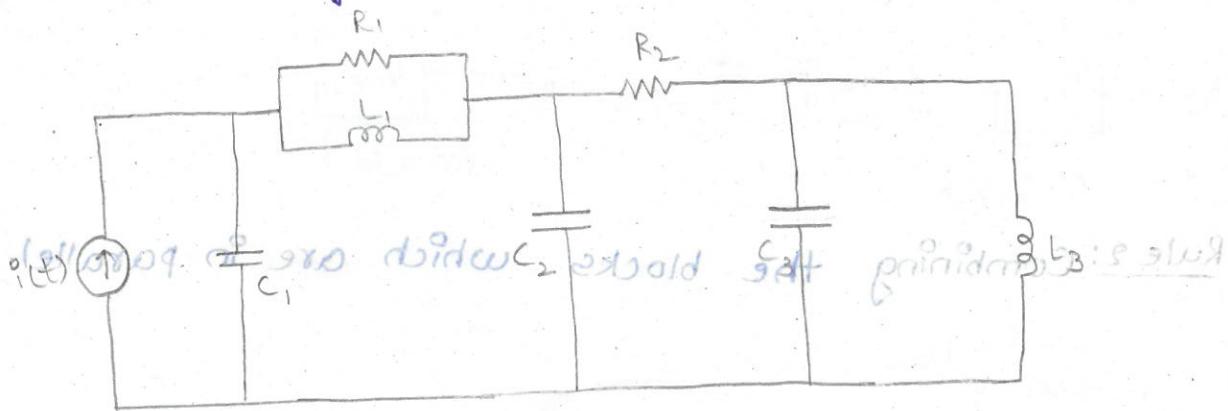
$$I(t) = C_1 \frac{dV_1}{dt} + \frac{1}{R_1} [V_1 - V_2] + \frac{1}{L_1} \int [V_1 - V_2] dt \quad \text{--- ⑧}$$

$$0 = C_2 \frac{dV_2}{dt} + \frac{1}{R_2} [V_2 - V_3] + \frac{1}{R_1} [V_2 - V_1] + \frac{1}{L_1} \int [V_2 - V_1] dt - \textcircled{8}$$

Considering eq-③, we get

$$0 = C_3 \frac{dV_3}{dt} + \frac{1}{L_3} \int V_3 dt + \frac{1}{R_2} [V_3 - V_2] - \textcircled{9}$$

By Considering 7,8,9 equations, we get Torque
current Analogous circuit bold set principle: 1.3 kN



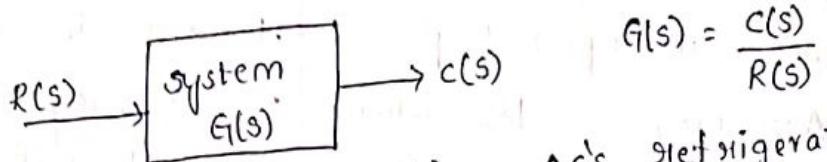
Page 3: Moving the source point set the bold set principle

bold set before moving the source point: Page 3

Control Systems → Hagoor Kani

A control system manages, commands, directs (or) regulates the behaviour of other devices (or) systems using control loops.

A control system is a system which provides the desired response by controlling the output. The simple block diagram of control system is shown by



$$G(s) = \frac{C(s)}{R(s)}$$

Ex:- Traffic light, control machine, AC's, refrigerators, control Systems.

Classification of continuous time & discrete time control systems

→ Continuous time & discrete time control systems

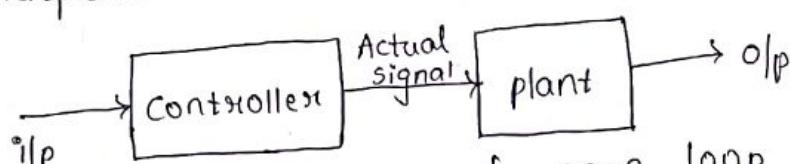
→ single input single output (SISO)

→ multi input multi output (MIMO) control systems

→ Open loop and closed loop Control Systems

* Control systems can be classified as open loop and closed loop controlled systems based on the feed back path.

→ In open loop control system output is not fed back to the input. So, the control action is independent of the desired output

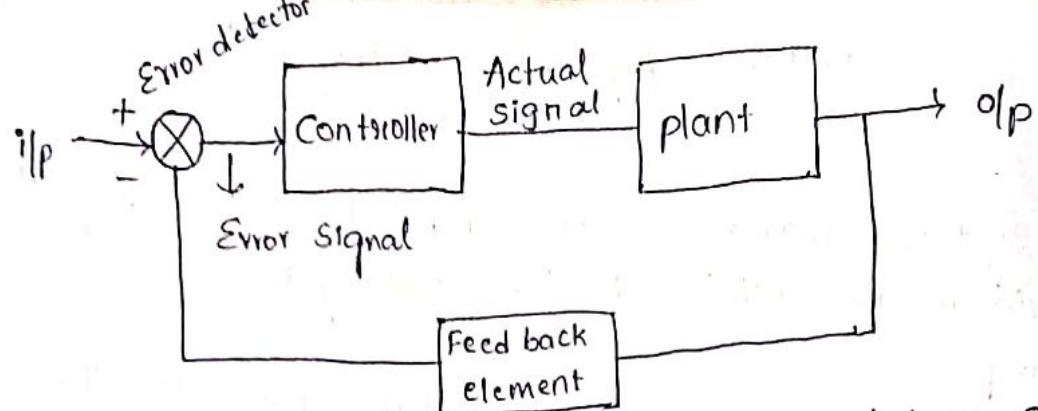


The block diagram of open loop control system

is shown above

→ In closed loop control system output is fed back to the input. So, the control action is dependent on the desired output.

The block diagram of closed loop control system is as shown:



Differences b/w open loop and closed loop control system

- | | |
|---|---|
| 1. There is no feed back element in open loop. | 1. There is feed back element in closed loop. |
| 2. Accuracy is less when compared with closed loop. | 2. Accuracy is more when compared with open loop. |
| 3. The control action is independent on the o/p. | 3. The control action is dependent on the o/p. |
| 4. Cost is less. | 4. Cost is more. |
| 5. There is less complexity. | 5. Complexity is more when compared with open loop. |

Control Systems

Systems:

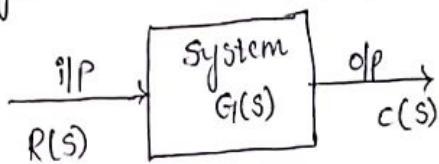
It is a combination of elements / components that are arranged in sequence to perform a specific function, the formation of this group is called a system.

Control Systems:

To achieve the desired output of the system by varying the input of the system is called controlled system.

Input of the control systems are command signal (or) excitation signal. It is represented by $R(s)$.

Output of the control systems are controlled signal (or) response signal. It is represented by $c(s)$.



Control Systems

Open loop
control system

(OLCS)

(or)
conditional controlled
system

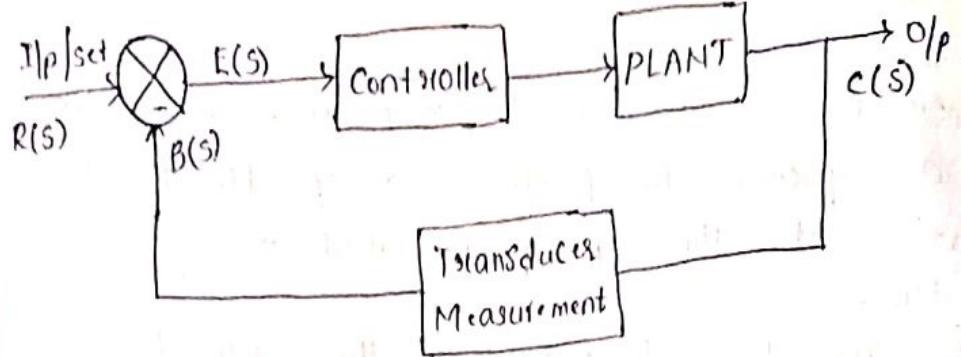
* OLCS are more stable than CLCS when condition is applied.

* CLCS are more stable than OLCS without condition.

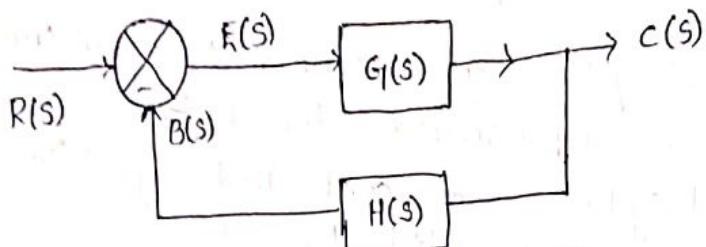
Closed loop control systems:

In closed loop control system, the changes in the O/p are measured through feed back and compared with the I/p (or) set point, to achieve the controlled objective.





Canonical form :-



Transfer Function for CLCS :-

$$E(s) = R(s) - B(s)$$

$$E(s) \cdot G(s) = c(s)$$

$$E(s) = \frac{c(s)}{G(s)}$$

$$\frac{c(s)}{G(s)} = R(s) - B(s)$$

$$c(s) = G(s) [R(s) - B(s)]$$

$$c(s) = G(s) R(s) - G(s) B(s)$$

$$c(s) = G(s) R(s) - E(s) \cdot H(s) c(s)$$

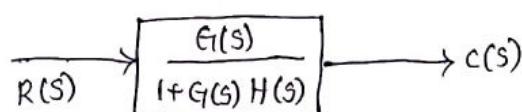
$$c(s) + G(s) H(s) c(s) = G(s) R(s)$$

$$c(s) [1 + G(s) H(s)] = G(s) R(s)$$

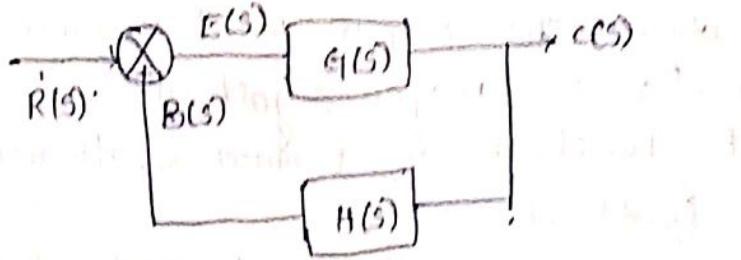
$$\frac{c(s)}{R(s)} = \frac{G(s)}{1 + G(s) H(s)}$$

$$TF = F(s) = \frac{c(s)}{R(s)} = \frac{G(s)}{1 + G(s) H(s)}$$

Mathematical form :-



Feedback with positive:



$$E(s) = R(s) + B(s)$$

$$\therefore E(s) G(s) = c(s)$$

$$E(s) = \frac{c(s)}{G(s)}$$

$$\frac{c(s)}{E(s)} = R(s) + B(s)$$

$$c(s) = G(s) [R(s) + B(s)]$$

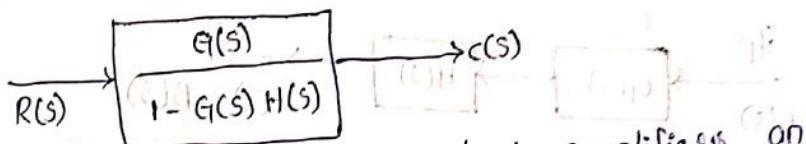
$$c(s) = G(s) R(s) + G(s) B(s)$$

$$c(s) = G(s) R(s) + G(s) H(s) c(s)$$

$$c(s) [1 - G(s) H(s)] = G(s) R(s)$$

$$TF = F(s) = \frac{c(s)}{R(s)} = \frac{G(s)}{1 - G(s) H(s)}$$

Mathematical form:



→ By combining the +ve feed back amplifiers and -ve feed back amplifiers, the gain is more in +ve feed back amplifiers rather than -ve feed back amplifiers.

→ -ve feed back amplifiers are more stable than +ve feed back amplifiers.

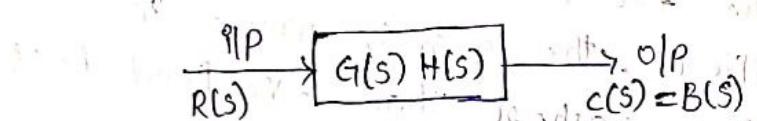
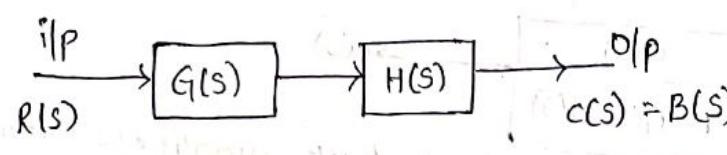
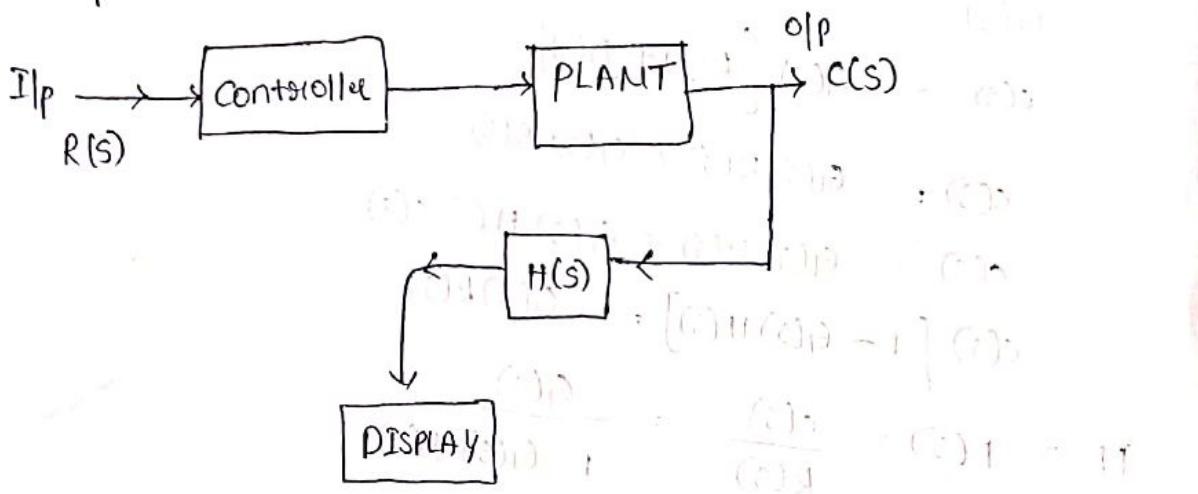
Open Loop Control System (OLCS):
They are conditional controlled systems formulated under the condition that the system is not subjected to any type of disturbances.

In this configuration, the feed back loop measurement is not connected to the forward path (or) controller.

→ Feedback is in open loop systems except for displaying the information about the output which is same as the setpoint has no major significance. This insignificance of feedback is termed as elimination (or) removal of feedback.

→ Open loop performance analysis is not applicable for open loop control systems because they are highly stable systems and they are not subjected to any type of disturbances.

→ Representation of openloop control system is



$$R(s) G(s) H(s) = c(s)$$

$$G(s) H(s) = \frac{c(s)}{R(s)}$$

$$T.F = P(s) = \frac{c(s)}{R(s)}$$

(a) Nonlinearity is due to presence of backlash with nonlinearity due to saturation with gain.

Differences b/w OLCS and CLCS

OLCS

- 1) It is easy to design.
- 2) It requires less cost.
- 3) Accuracy is less in OLCS.

OLCS

- 4) Gain is more.
- 5) Under condition, OLCS are more stable.
- 6) without condition, OLCS are unstable.

Examples of OLCS & CLCS :-

↳ Temperature control system

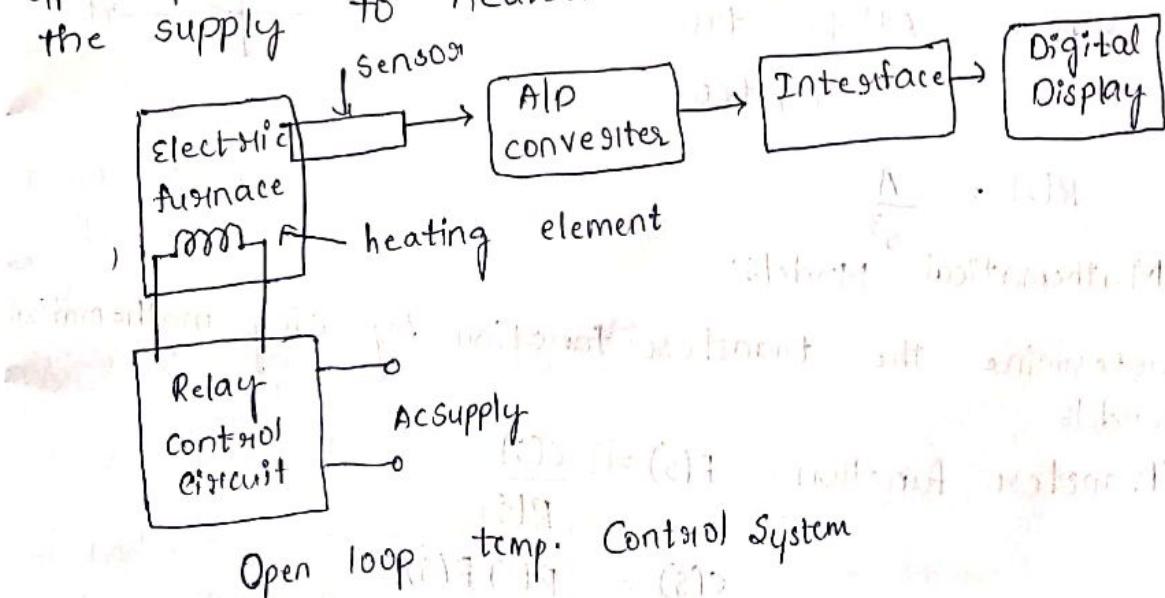
→ Traffic control system

→ Numerical control system

Temperature Control System:

Open loop System:

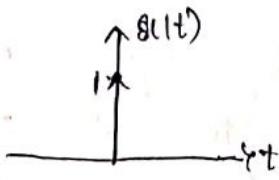
The electric furnace shown in fig 1.3 is an open loop system. The o/p in the system is the desired temp. The temp. of the system is raised by heat generated by the heating element. The o/p temperature depends on the time during which the supply to heater remains ON.



Weighting Functions :-

1. Impulse Ramp Signal :-

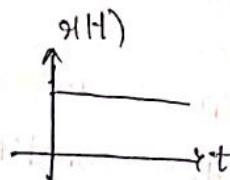
$$g(t) = \begin{cases} 1, & t=0 \\ 0, & \text{else} \end{cases}$$



$$R(s) = 1$$

2. Step Signal :-

$$u(t) = \begin{cases} 1, & t \geq 0 \\ 0, & \text{else} \end{cases}$$



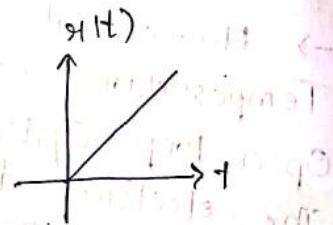
$$h(t) = \begin{cases} A u(t), & t \geq 0 \\ 0, & \text{else} \end{cases}$$

$$g(t) = A u(t) = \begin{cases} 1, & t \geq 0 \\ 0, & \text{else} \end{cases}$$

$$R(s) = \frac{A}{s}$$

3. Ramp Signal :-

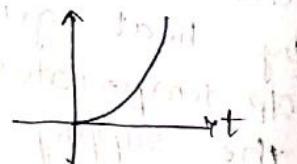
$$g(t) = \begin{cases} At \neq 0, & t \neq 0 \\ 0, & t=0 \end{cases}$$



$$R(s) = \frac{A}{s^2}$$

4) Parabolic signal :-

$$h(t) = \begin{cases} \frac{At^2}{2} \neq 0, & t \neq 0 \\ 0, & t=0 \end{cases}$$



$$R(s) = \frac{A}{s^3}$$

Mathematical Models:

Determine the transfer function by using mathematical models:

$$\text{Transfer function} : F(s) = \frac{C(s)}{R(s)}$$

$$C(s) = R(s) F(s)$$

$$R(s) = 1$$

$$C(s) = F(s)$$

Transfer function is defined as the Laplace transform

of impulse response. is known as weighting function.

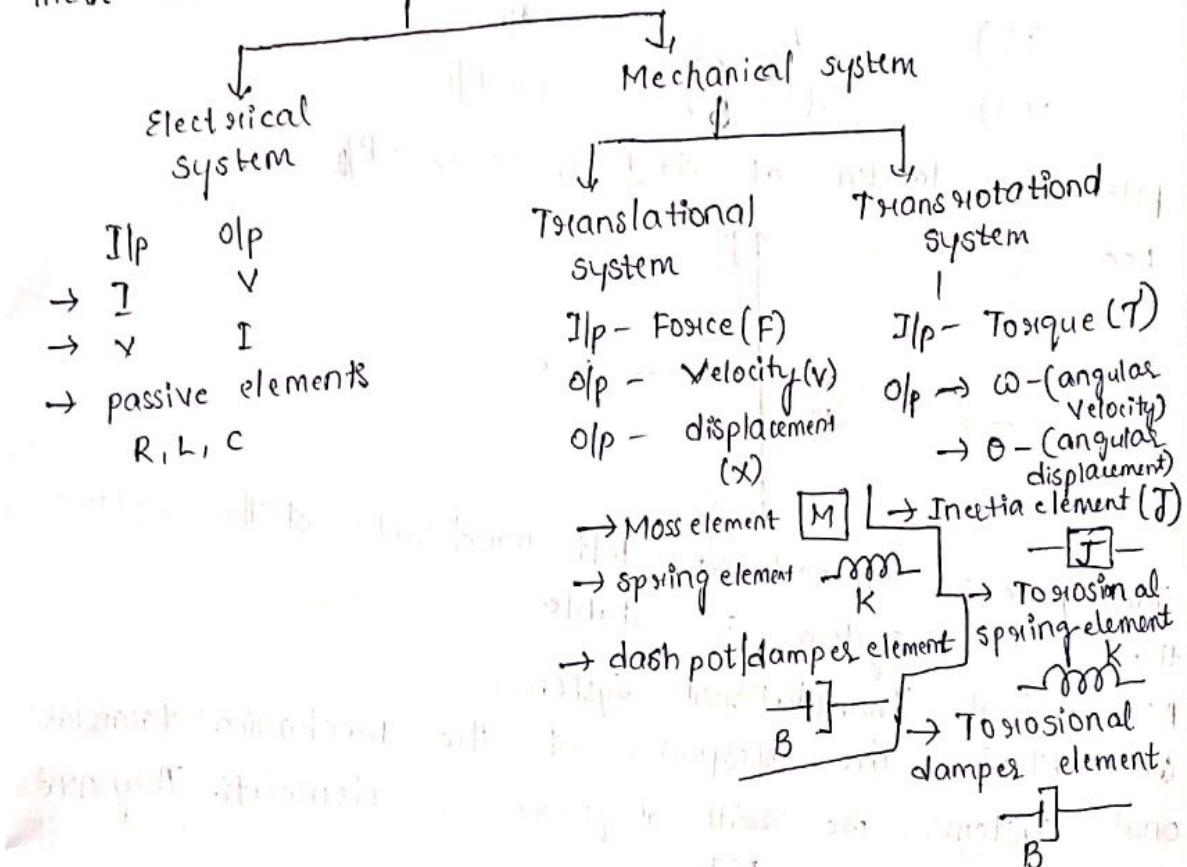
Mathematical Model of a control System:-

→ Mathematical Model of a C.S is a combination of various components are connected in sequence of a system to serve an objective.

→ The input & output relation of a system is represented by differential equations.

→ To obtain the response of the system with the help of studying the response of the system with the help of by solving the differential equations.

These are classified into 2 types.



Electrical System:-

→ It consists of I, V, R, L, C

$$R \rightarrow V = IR \rightarrow I = V/R$$

$$L \rightarrow V = L \frac{di}{dt} \rightarrow I = \frac{1}{L} \int V dt$$

$$C \rightarrow V = \frac{1}{C} \int i dt \rightarrow I = C \frac{dv}{dt}$$

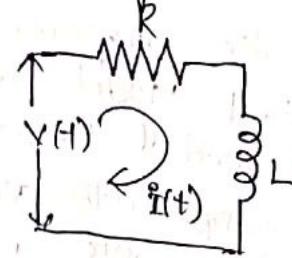
→ Inductor stores the energy in the form of current

→ Capacitor stores the energy in the form of voltage

Determine the T.F. for given ckt diagram.

$$V(t) = iR + L \frac{di}{dt}$$

$$V(t) = i(t)R + L \frac{di(t)}{dt}$$



Apply LS

$$V(s) = I(s)R + LS I(s)$$

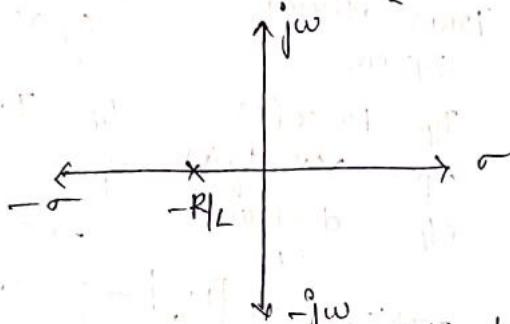
$$V(s) = I(s)(R + LS)$$

$$\frac{I(s)}{V(s)} = \frac{1}{R + LS}$$

$$\frac{I(s)}{V(s)} = \frac{1}{s(L + \frac{R}{S})} = \frac{1/L}{s + R/L}$$

Poles are located at $s + \frac{R}{L} = 0 \Rightarrow s = -R/L$

Roc



The pole is located at left hand side of the s-plane.
Hence the system is stable

Mechanical Translational Systems:

We obtain the response of the mechanical translational systems. We will require 3 elements. They are

Mass element - \boxed{M}

Spring element - \boxed{mK} (Elasticity is present)

damper element - $\boxed{I_B}$ (Friction is present)

displacement - x

Velocity, $v = \frac{dx}{dt}$

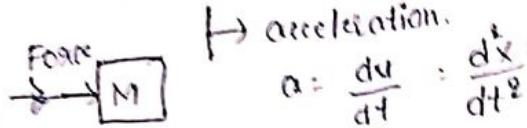
Acceleration, $a = \frac{d^2x}{dt^2}$

Note: $a = \frac{dv}{dt} = \frac{d}{dt}\left(\frac{dx}{dt}\right) = \frac{d^2x}{dt^2}$

All these 3 elements will exhibits Newton's 2nd law.

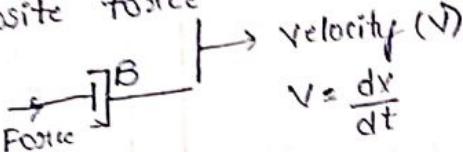
Mass element :-

If force applied on mass element it will exhibits opposite force from the mass element i.e., acceleration.



Damper element :-

If force applied on damper element it will exhibits opposite force which is in the form of velocity.



Spring element :-

If force applied on spring element it will exhibits opposite force which is in the form of displacement.



→ The sum of the forces acting on the body is equal to zero.

Note :- If 2 elements are connected in series consider a dummy node in b/w these elements.

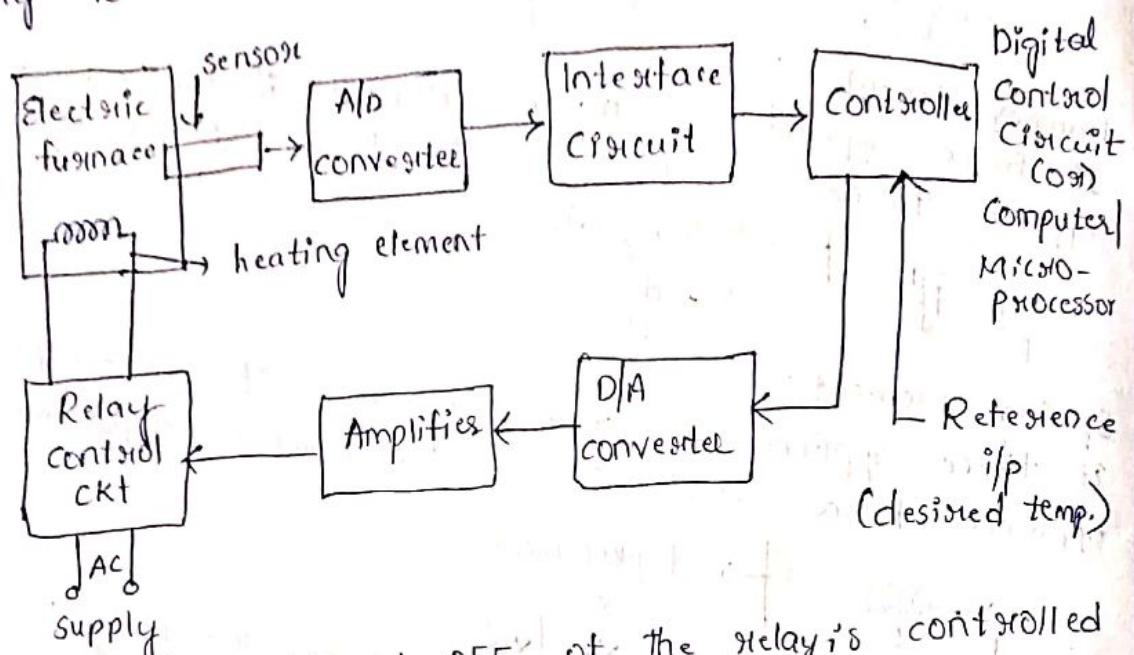
Ex:-

M_1 - \bullet - M_2

⇒ Examples of CS :-

Temperature control system :-
The ON and OFF of the supply is governed by the temperature is.
The time setting of the relay. The temperature gives an analog signal corresponding to the temperature of the furnace. The analog signal is converted to digital signal by an Analog-to-digital converter (ADC). The digital signal is given to the digital display device to display the temperature. In this system if there is any change in output temperature then the time setting of the relay is not altered automatically.

Closed loop System:
 The electric furnace shown in fig. is a closed loop system. The o/p of the system is the desired temperature and it depends on the time during which the supply to heater remains ON.



The switching ON and OFF of the relay's controlled by a controller which is a digital system (or) computer. The desired temperature is fed to the system through keyboard or as a signal corresponding to desired temp. via ports. The actual temp. is sensed by sensor and converted to digital signal by the A/D converter. The computer reads the actual temperature and compares with desired temperature. If it finds any difference then it sends signal to switch ON or OFF the relay through D/A converter and amplifier. Thus the system automatically corrects any changes in o/p. Hence it is a closed loop system.

Traffic Control System:

Open loop System:

Traffic control by means of traffic signals operated on a time basis constitutes an open-loop control system. The sequence of control signals are based on a time slot given for each signal. The time slots are decided based on a traffic study. The system will not measure the density of the traffic before giving the signals. Since the timeslot does not change according to traffic density, the system is

open loop system.

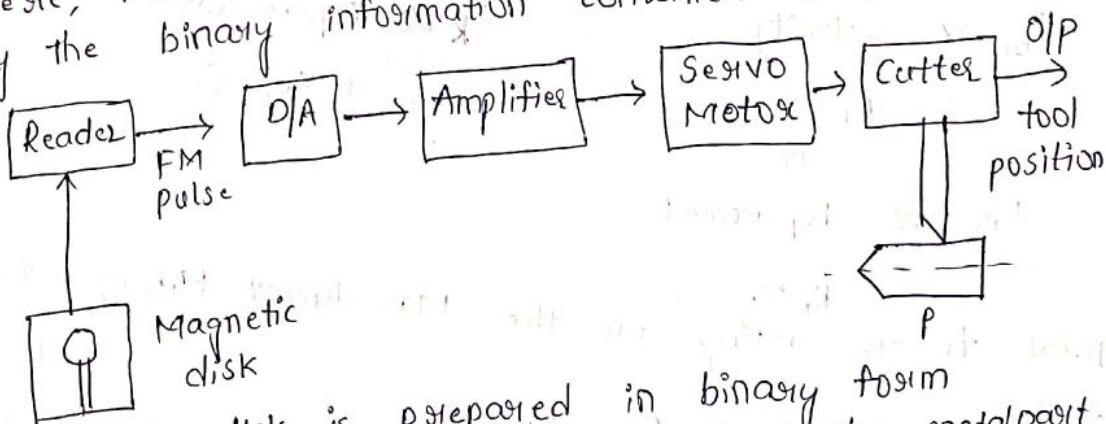
closed loop System:

Traffic control system can be made as a closed loop system if the time slots of the signals are decided based on the density of traffic. In closed loop traffic control system, the density of the traffic is measured on all the sides and the information is fed to a computer. The timings of the control signals are decided by the computer. Since the closed system based on the density of traffic changes the timings, the loop system dynamically changes the timings, the flow of vehicles will be better than open loop system.

Numerical control System:-

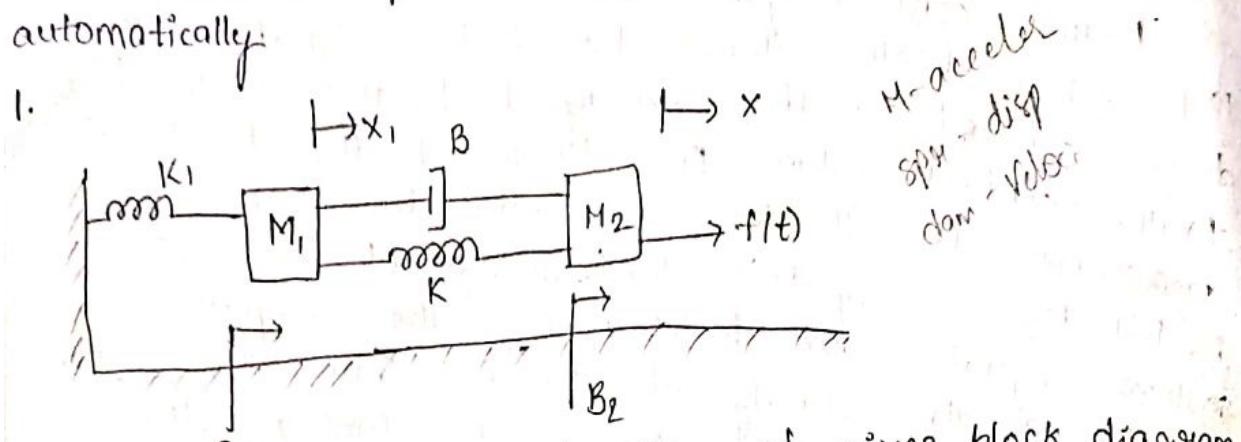
Open loop system:

Numerical control is a method of controlling the motion of machine components using numbers. The position of work head tool is controlled by the binary information contained in a disk.



A magnetic disk is prepared in binary form representing the desired part P (P is the metal part to be machined). The tool will operate on the desired part P. To start the system, the disk is fed through the reader to the D/A converter. The D/A converter converts the FM O/p of the reader to an analog signal. It is amplified and fed to the servomotor which positions the cutter on the desired part P. The position of the cutter head is controlled by the angular motion of the servomotor. This is an open loop system since no feedback path exists b/w the o/p and i/p. The system

positions the tool for a given I/p command. Any deviation in the desired position is not checked and corrected automatically.



Determine the transfer function of given block diagram. No. of differential equations is dependent on no. of mass elements.

At node M₁:

$$F_M \propto \text{acceleration}$$

$$= M_1 \frac{d^2 x_1}{dt^2}$$

$$F_B \propto \text{velocity}$$

$$= B \frac{d}{dt} (x_1 - x)$$

$$F_{B_1} \propto \text{velocity}$$

$F_k \propto$ displacement

$$= B_1 \frac{dx_1}{dt} = K(x_1 - x)$$

$F_K \propto$ displacement

$\vec{F}_1 = k_1 \vec{x}_1$ acting on the mass element M_1 is equal to

$$F_{M_1} + F_{k_1} + F_{B_1} + F_B + F_k = 0$$

$$\Rightarrow M_1 \frac{d^2x_1}{dt^2} + k_1 x_1 + B_1 \frac{dx_1}{dt} + B \frac{d}{dt}(x_1 - x) + k(x_1 - x) = 0$$

Apply Laplace Transform on both sides.

$$\Rightarrow M, s^2 x_1(s) + k, x_1(s) + B, s x_1(s) + B [s x_1(s) - x_1(s)] +$$

$$k[x_1(s) - x(s)] = 0$$

$$\Rightarrow x_1(s) \left[M_1 s^2 + K_1 + B_1 s + BS + k \right] + x(s) \left[-BS - k \right] = 0$$

$$\Rightarrow x_1(s) [M_1 s^2 + k_1 + B_1 s + BS + k] - x(s) [BS + k] = 0 \quad \text{---(1)}$$

At node M_2 :

$$F_{M_2} \propto \text{acceleration}$$
$$= M_2 \frac{d^2x}{dt^2}$$

$F_{B_2} \propto \text{velocity}$

$$= B_2 \frac{dx}{dt}$$

At Mass M_2 , applied force is equal to sum of opposite forces.

$$f(t) = M_2 \frac{d^2x}{dt^2} + B_2 \frac{dx}{dt} + B \frac{d}{dt}(x - x_1) + k(x - x_1)$$

Apply L/T on LHS

$$F(s) = M_2 s^2 x(s) + B_2 s x(s) + B [s x(s) - s x_1(s)] + k(x(s) - x_1(s))$$

$$F(s) = x(s) [M_2 s^2 + B_2 s + B s + k] - x_1(s) [B s + k] \quad \text{--- (2)}$$

$$\text{From eq } \text{--- (1)} \quad x(s) [M_1 s^2 + k_1 + B_1 s + B s + k] = x(s) [B s + k]$$

$$\Rightarrow x_1(s) [M_1 s^2 + k_1 + B_1 s + B s + k] = x(s) [B s + k]$$

$$\Rightarrow x_1(s) = \frac{x(s) (B s + k)}{M_1 s^2 + k_1 + B_1 s + B s + k} \quad \text{--- (3)}$$

Sub. eq (3) in eq (2)

$$F(s) = x(s) [M_2 s^2 + B_2 s + B s + k] - \frac{(B s + k)^2 x(s)}{M_1 s^2 + k_1 + (B_1 + B) s + k}$$

$$= x(s) \left\{ M_2 s^2 + s(B + B_2) + k - \frac{(B s + k)^2}{M_1 s^2 + k_1 + (B_1 + B) s + k} \right\}$$

$$F(s) = x(s) \left\{ \frac{[M_2 s^2 + s(B + B_2) + k] [M_1 s^2 + k_1 + (B + B_1) s + k] - (B s + k)^2}{M_1 s^2 + k_1 + (B_1 + B) s + k} \right\}$$

$$\Rightarrow \frac{x(s)}{F(s)} = \frac{M_1 s^2 + k_1 + (B_1 + B) s + k}{[M_2 s^2 + s(B + B_2) + k] [M_1 s^2 + k_1 + (B + B_1) s + k] - (B s + k)^2}$$

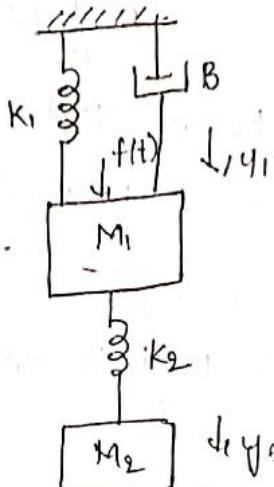
$F_B \propto \text{Velocity}$

$$= B \frac{d}{dt}(x - x_1)$$

$F_K \propto \text{displacement}$

$$= k(x - x_1)$$

Q. Write the eqns of motion in s-domain for the system shown. Determine the transfer fn of the system.



No. of differential equations is dependent on no. of mass elements.

At node M_1 :

$F_{M_1} \propto$ acceleration

$$= M_1 \frac{d^2 y_1}{dt^2}$$

$F_{K_1} \propto$ displacement

$$\rightarrow K_1 y_1$$

$F_B \propto$ Velocity

$$= B \frac{dy_1}{dt}$$

$F_{K_2} \propto$ displacement

$$= K_2 (y_1 - y_2)$$

Applied force is equal to sum of opposite forces.

$$\Rightarrow F_{M_1} + F_{K_1} + F_B + F_{K_2} = f(t)$$

$$\Rightarrow f(t) = M_1 \frac{d^2 y_1}{dt^2} + K_1 y_1 + B \frac{dy_1}{dt} + K_2 (y_1 - y_2)$$

Apply LIT on b.s

$$F(s) = M_1 s^2 Y_1(s) + K_1 Y_1(s) + B s Y_1(s) + K_2 [Y_1(s) - Y_2(s)]$$

$$F(s) = Y_1(s) [M_1 s^2 + K_1 + B s + K_2] - K_2 Y_2(s) \quad \text{---(1)}$$

At node M_2 :

$F_{M_2} \propto$ acceleration

$$= M_2 \frac{d^2 y_2}{dt^2}$$

$F_{K_2} \propto$ displacement

$$= K_2 (y_2 - y_1)$$

Total force acting on the Mass M_2 is equal to zero.

$$F_{M_2} + F_{K_2} = 0$$

$$M_2 \frac{d^2 y_2}{dt^2} + K_2(y_2 - y_1) = 0$$

Apply LT on b.s

$$M_2 s^2 Y_2(s) + K_2 [Y_2(s) - Y_1(s)] = 0$$

$$[M_2 s^2 + K_2] Y_2(s) - K_2 Y_1(s) = 0 \quad \text{--- (2)}$$

From eq (2)

$$Y_1(s) = \frac{[M_2 s^2 + K_2] Y_2(s)}{K_2} \quad \text{--- (3)}$$

Sub. eq (3) in eq (1)

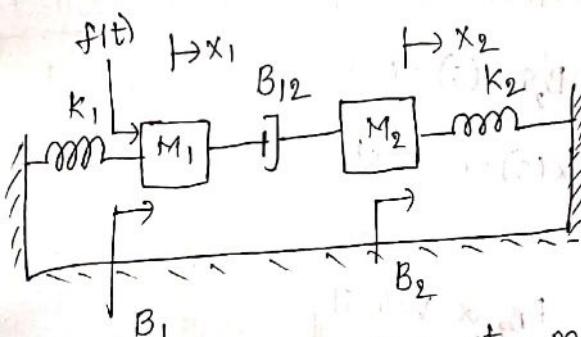
$$F(s) = [M_2 s^2 + K_2] Y_2(s) [M_1 s^2 + K_1 + BS + K_2] - K_2 Y_2(s)$$

$$= Y_2(s) \left[\frac{(M_2 s^2 + K_2)(M_1 s^2 + (K_1 + K_2 + BS)) - K_2}{K_2} \right]$$

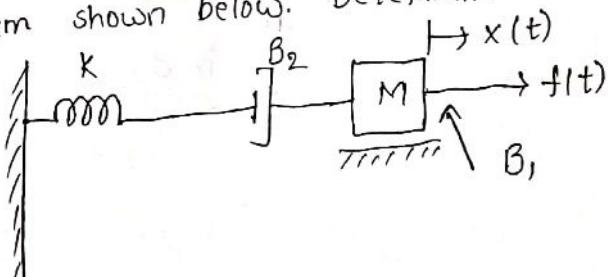
$$F(s) = Y_2(s) \left[\frac{(M_2 s^2 + K_2)(M_1 s^2 + K_1 + K_2 + BS) - K_2^2}{K_2} \right]$$

$$\Rightarrow \text{TF} = \frac{Y_2(s)}{F(s)} = \frac{K_2}{(M_2 s^2 + K_2)(M_1 s^2 + K_1 + K_2 + BS) - K_2^2}$$

3. Determine the transfer function $\frac{x_1(s)}{F(s)}$ and $\frac{x_2(s)}{F(s)}$ for the system shown below:

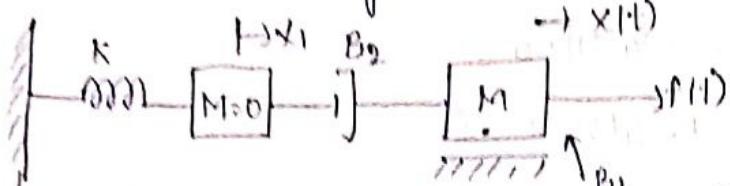


4. Write the equation of motion in s-domain for the system shown below. Determine the TF of the system.



Conse
By observing the above diagram, when elements are connected in series consider a dummy node in b/w these elements.

\therefore The above diagram can be equal to



No. of differential eqn's is dependent on no. of mass elements.

At node $M=0$:

$F_K \propto$ displacement

$$= Kx_1$$

$F_M \propto$ acceleration

$$= M \frac{d^2x_1}{dt^2} \quad (\because M=0)$$

$F_{B2} \propto$ Velocity

$$= B_2(x_1 - x)$$

$$= B_2 \frac{d}{dt}(x_1 - x)$$

$$F_M = 0$$

Total force acting on Mass $M=0$ is equal to zero.

$$F_K + F_M + F_{B2} = 0$$

$$Kx_1 + M \frac{d^2x_1}{dt^2} + B_2 \frac{d(x_1 - x)}{dt} = 0$$

Apply L/T

$$Kx_1(s) + M s^2 x_1(s) + B_2 s [x_1(s) - x(s)] = 0$$

$$x_1(s) [K + Ms^2 + B_2 s] - B_2 s x(s) = 0$$

$$\Rightarrow x_1(s) [K + B_2 s] - B_2 s x(s) = 0 \quad \text{---(1)}$$

At node M :

$F_M \propto$ acceleration

$$= M \frac{d^2x}{dt^2}$$

$F_{B2} \propto$ Velocity

$$= B_2 \frac{d}{dt}(x - x_1)$$

$F_{B1} \propto$ Velocity

$$= B_1 \frac{dx}{dt}$$

Applied force is equal to sum of opposition forces

$$F_M + F_{B_1} + F_{B_2} = f(t)$$

$$M \frac{d^2x}{dt^2} + B_1 \frac{dx}{dt} + B_2 \frac{d}{dt}(x - x_1) = f(t)$$

Apply L.T on L.S

$$M s^2 X(s) + B_1 s X(s) + B_2 [X(s) - x_1(s)]s = F(s)$$

$$X(s) [M s^2 + B_1 s + B_2 s] - B_2 s x_1(s) = F(s) \quad \text{---(2)}$$

From eq ①

$$x_1(s) = \frac{B_2 s X(s)}{K + B_2 s} \quad \text{---(3)}$$

eq ③ in eq ②

$$X(s) (M s^2 + B_1 s + B_2 s) - B_2 s \left(\frac{B_2 s X(s)}{K + B_2 s} \right) = F(s)$$

$$X(s) \left[(M s^2 + B_1 s + B_2 s) - \frac{(B_2 s)^2}{K + B_2 s} \right] = F(s)$$

$$\Rightarrow \frac{X(s)}{F(s)} = \frac{[M s^2 + (B_1 + B_2)s][K + B_2 s] - (B_2 s)^2}{K + B_2 s} = F(s)$$

$$\Rightarrow \text{TF} = \frac{X(s)}{F(s)} = \frac{K + B_2 s}{(M s^2 + B_1 s + B_2 s)(K + B_2 s) - (B_2 s)^2}$$

3) No. of differential equations is dependent on no. of mass elements.

At node M_1 :

$F_M \propto$ acceleration $F_{B_{12}} \propto$ Velocity

$$= M_1 \frac{d^2 x_1}{dt^2}$$

$$= B_{12} \frac{d(x_1 - x)}{dt}$$

$F_{K_1} \propto$ displacement

$$= K_1 x_1$$

$F_{B_1} \propto$ Velocity

$$= B_1 \frac{dx_1}{dt}$$

Applied force is equal to sum of opposite forces.

$$F_M + F_{K_1} + F_{B_1} + F_{B_{12}} = f(t)$$

$$F(t) = M_1 \frac{d^2 x_1}{dt^2} + K_1 x_1 + B_1 \frac{dx_1}{dt} + B_{12} \frac{d}{dt}(x_1 - x_2)$$

Apply LT on b.s

$$F(s) = M_1 s^2 X_1(s) + K_1 X_1(s) + B_1 s X_1(s) + B_{12} [X_1(s) - X_2(s)] s$$

$$F(s) = X_1(s) [M_1 s^2 + K_1 + B_1 s + B_{12} s] - B_{12} s X_2(s) \quad \text{--- (1)}$$

At node M_2 :-

$F_{M_2} \propto$ acceleration

$$= M_2 \frac{d^2 x_2}{dt^2}$$

$F_{B_2} \propto$ Velocity

$$= B_2 \frac{dx_2}{dt}$$

$F_{K_2} \propto$ displacement

$$= K_2 x_2$$

$F_{B_{12}} \propto$ Velocity

$$= B_{12} \frac{d}{dt}(x_2 - x_1)$$

Total force acting on mass M_2 is equal to zero.

$$F_{M_2} + F_{K_2} + F_{B_2} + F_{B_{12}} = 0$$

$$\Rightarrow M_2 \frac{d^2 x_2}{dt^2} + K_2 x_2 + B_2 \frac{dx_2}{dt} + B_{12} \frac{d}{dt}(x_2 - x_1) = 0$$

Apply LT on b.s

$$\Rightarrow M_2 s^2 X_2(s) + K_2 X_2(s) + B_2 s X_2(s) + B_{12} [X_2(s) - X_1(s)] s = 0$$

$$\Rightarrow [M_2 s^2 + K_2 + B_2 s + B_{12} s] X_2(s) - B_{12} s X_1(s) = 0 \quad \text{--- (2)}$$

From eq (2).

$$[M_2 s^2 + K_2 + B_2 s + B_{12} s] X_2(s) = B_{12} s X_1(s)$$

$$X_1(s) = \frac{M_2 s^2 + K_2 + B_2 s + B_{12} s (X_2(s))}{B_{12} s} \quad \text{--- (3)}$$

$$X_2(s) = \frac{B_{12} s X_1(s)}{M_2 s^2 + K_2 + B_2 s + B_{12} s} \quad \text{--- (4)}$$

Sub. eq (3) in eq (1)

$$F(s) = \left(\frac{M_2 s^2 + K_2 + B_2 s + B_{12} s}{B_{12} s} \right) [M_1 s^2 + K_1 + B_1 s + B_{12} s] - B_{12} s X_2(s)$$

$$F(s) = X_2(s) \left[\frac{[M_2 s^2 + K_2 + (B_2 + B_{12}) s]}{B_{12} s} [M_1 s^2 + K_1 + B_1 s + B_{12} s] - B_{12} s \right]$$

$$F(s) = X_2(s) \left\{ \frac{[M_2 s^2 + K_2 + (B_2 + B_{12})s][M_1 s^2 + K_1 + B_1 s + B_{12}s] - (B_{12}s)^2}{B_{12}s} \right\}$$

$$\Rightarrow \frac{X_2(s)}{F(s)} = \frac{B_{12}s}{(M_2 s^2 + K_2 + B_2 s + B_{12}s)(M_1 s^2 + K_1 + B_1 s + B_{12}s) - (B_{12}s)^2}$$

Sub eq (4) in eq (1)

$$F(s) = X_1(s) (M_1 s^2 + K_1 + B_1 s + B_{12}s) - B_{12}s \left[\frac{B_{12}s X_1(s)}{M_2 s^2 + K_2 + B_2 s + B_{12}s} \right]$$

$$= X_1(s) \left[M_1 s^2 + K_1 + B_1 s + B_{12}s - \frac{(B_{12}s)^2}{M_2 s^2 + K_2 + B_2 s + B_{12}s} \right]$$

$$F(s) = X_1(s) \left[\frac{(M_1 s^2 + K_1 + B_1 s + B_{12}s)(M_2 s^2 + K_2 + B_2 s + B_{12}s) - (B_{12}s)^2}{M_2 s^2 + K_2 + B_2 s + B_{12}s} \right]$$

$$\frac{X_1(s)}{F(s)} = \frac{M_2 s^2 + K_2 + B_2 s + B_{12}s}{(M_1 s^2 + K_1 + B_1 s + B_{12}s)(M_2 s^2 + K_2 + B_2 s + B_{12}s) - (B_{12}s)^2}$$

Mathematical Model of Mechanical Transrotational system :-
We determine the mathematical model of a mechanical transrotational systems we require 3 elements. Those are Inertia of mass (J) , torsional spring element (K) , torsional damper / dashpot element (B) .

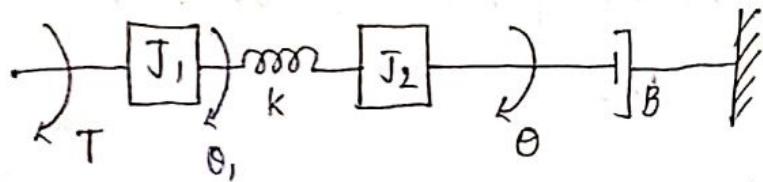
I/p - Torque (T)

O/p - angular displacement (θ)
angular velocity (ω)

$$\omega = \frac{d\theta}{dt}$$

angular acceleration $a = \frac{d\omega}{dt} \equiv \frac{d^2\theta}{dt^2}$
the transfer function for the given transrotational system:

I. Determine mechanical



No. of differential eq's is dependent on no. of inertia of mass elements i.e., 2.

At node J₁:

$$T_{J_1} \propto \text{angular acceleration}$$

$$= J_1 \frac{d^2\theta_1}{dt^2}$$

$$T_K \propto \text{displacement}$$

$$= K\theta_1 - K\theta$$

Applied Torque is equal to sum of opposite torques.

$$T_{J_1} + T_K = T$$

$$J_1 \frac{d^2\theta_1}{dt^2} + \frac{K}{J_1}(\theta_1 - \theta) = T$$

Apply LT on b.s

$$J_1 s^2 \theta_1(s) + \frac{K}{J_1} [s\theta_1(s) - \theta(s)] = T(s)$$

$$\theta_1(s) [J_1 s^2 + \frac{K}{J_1}] - \frac{K}{J_1} \theta(s) = T(s) \quad \text{--- (1)}$$

At node J₂:

$$T_{J_2} \propto \text{angular acceleration}$$

$$= J_2 \frac{d^2\theta}{dt^2}$$

$$T_B \propto \text{Velocity}$$

$$= B \frac{d\theta}{dt}$$

$$T_K \propto \text{angular displacement}$$

$$= K(\theta - \theta_1)$$

Total torque acting on

$$T_{J_2} + T_B + T_K = 0$$

$$J_2 \frac{d^2\theta}{dt^2} + B \frac{d\theta}{dt} + K(\theta - \theta_1) = 0$$

Apply LT on b.s

$$J_2 s^2 \theta(s) + B s \theta(s) + K[s\theta(s) - \theta_1(s)] = 0$$

$$\theta(s) [J_2 s^2 + B s + K] - K \theta_1(s) = 0 \quad \text{--- (2)}$$

$$\text{--- (2)} \Rightarrow \theta_1(s) = \frac{(J_2 s^2 + B s + K) \theta(s)}{K} \quad \text{--- (3)}$$

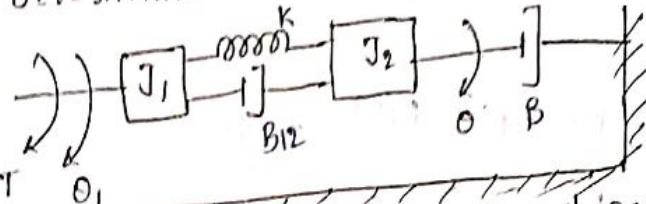
sub eq (3) in eq (1)

$$T(s) = \frac{(J_2 s^2 + B_2 s + k)}{s} \cdot \theta(s) \cdot (J_1 s^2 + k) - k \theta(s)$$

$$T(s) = \theta(s) \left[\frac{(J_2 s^2 + B_2 s + k)(J_1 s^2 + k) - k^2}{k} \right]$$

$$TF = \frac{\theta(s)}{T(s)} \cdot \frac{k}{(J_2 s^2 + B_2 s + k)(J_1 s^2 + k) - k^2}$$

Q. Determine the TF for a given translational system.



No. of differential eq's is dependent on no. of inertia elements.

At node J_1 :

$T_{J_1} \propto$ ang. acceleration

$$= J_1 \frac{d^2 \theta_1}{dt^2}$$

$T_{B_{12}} \propto$ ang. velocity

$$= B_{12} \frac{d \theta_1}{dt}$$

$T_K \propto$ ang. displacement
= $k(\theta_1 - \theta)$

Applied torque is equal to sum of opposite torques.

$$T = T_{J_1} + T_{B_{12}} + T_K$$

$$T = J_1 \frac{d^2 \theta_1}{dt^2} + B_{12} \frac{d \theta_1}{dt} + k(\theta_1 - \theta)$$

Apply LT on b.s

$$T(s) = J_1 s^2 \theta_1(s) + B_{12} s [\theta_1(s) - \theta(s)] + k[\theta_1(s) - \theta(s)]$$

$$T(s) = \theta_1(s) [J_1 s^2 + B_{12} s + k] - [B_{12} s + k] \theta(s) \quad \text{--- (1)}$$

At node J_2 :

$T_{J_2} \propto$ ang. acceleration

$$= J_2 \frac{d^2 \theta_2}{dt^2}$$

$T_B \propto$ ang. Velocity

$$= B \frac{d \theta_2}{dt}$$

$T_K \propto$ ang. displacement

$$= k(\theta - \theta_2)$$

$T_{B_{12}} \propto$ ang. Velocity

$$= B_{12} \frac{d \theta}{dt}$$

Total torque acting on J_2 is τ_{c90} .

$$T_{J_2} + T_B + T_K + T_{B_{12}} > 0$$

$$J_2 \frac{d^2\theta}{dt^2} + B \frac{d\theta}{dt} (0 - \theta_1) + K(0 - \theta_1) + B_{12} \frac{d\theta}{dt} (0 - \theta_1) = 0$$

Apply L/H on b.s

$$\Rightarrow J_2 s^2 \theta(s) + B s [\theta(s) - \theta_1] + K[\theta(s) - \theta_1] + B_{12} s [\theta(s) - \theta_1] = 0$$

$$\theta(s) [J_2 s^2 + B s + K + B_{12} s] - \theta_1 (s) [B s + K + B_{12} s] = 0 \quad \text{--- (2)}$$

From eq (2)

$$\Rightarrow \theta_1(s) = \frac{[J_2 s^2 + B s + K + B_{12} s]}{B s + K + B_{12} s} \theta(s) \quad \text{--- (3)}$$

Substitute eq (3) in eq (1)

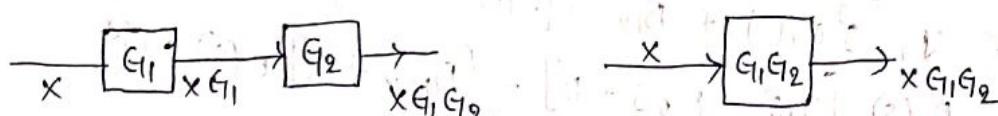
$$T(s) = \left[\frac{[J_2 s^2 + B s + K + B_{12} s]}{B s + K + B_{12} s} \right] [J_1 s^2 + B_{12} s + K] \theta(s) - [B_{12} s + K] \theta(s)$$

$$T(s) = \theta(s) \left\{ \frac{[J_2 s^2 + B s + K + B_{12} s] (J_1 s^2 + B_{12} s + K)}{B s + K + B_{12} s} - [B_{12} s + K] \right\}$$

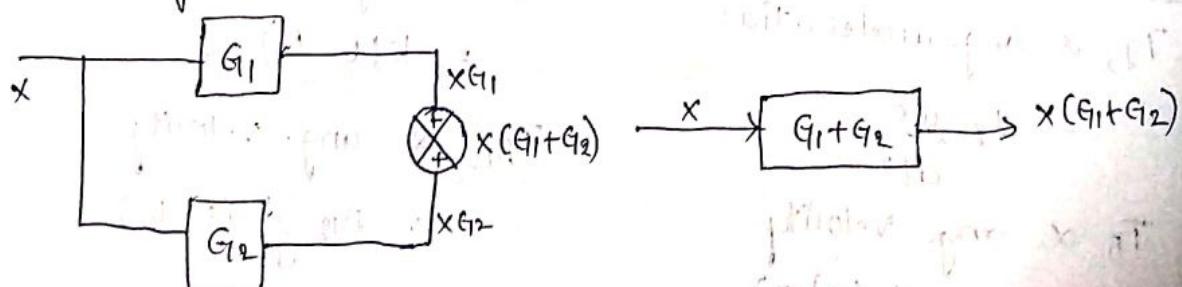
$$\text{TF} = \frac{\theta(s)}{T(s)} = \frac{B s + K + B_{12} s}{(J_2 s^2 + B s + K + B_{12} s)(J_1 s^2 + B_{12} s + K) - (B_{12} s + K)^2}$$

Mathematical Reduction Method :-

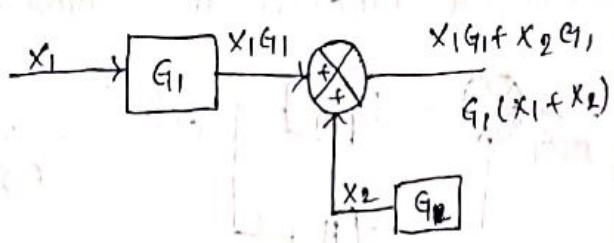
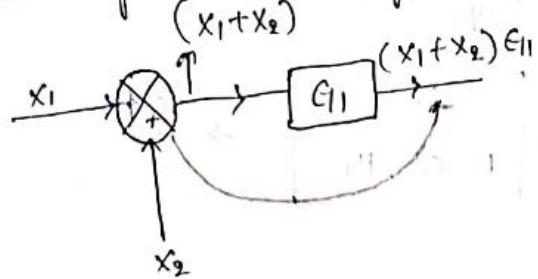
Rule-1 :-
Combining the blocks in series.



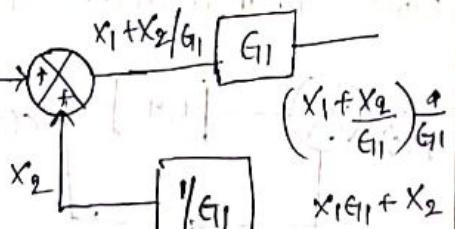
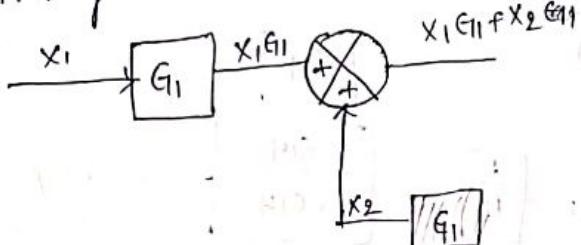
Rule-2 :-
Combining the blocks in parallel.



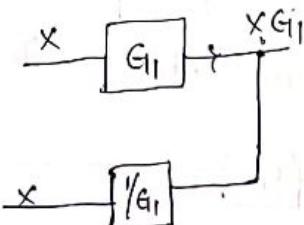
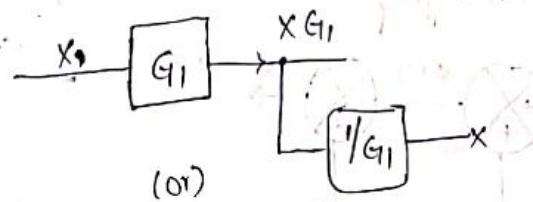
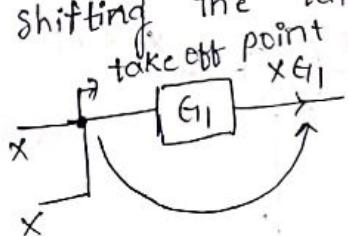
Rule-3 :- Shifting the summing point after the block.



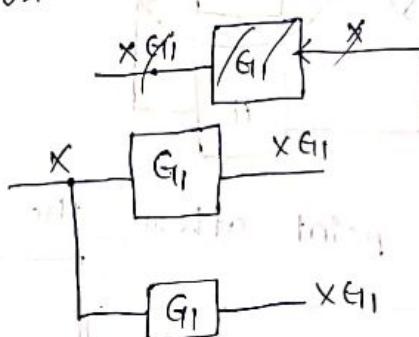
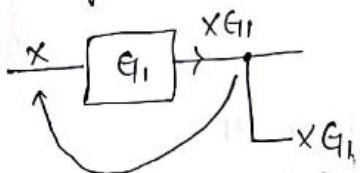
Rule-4 :- Shifting the summing point before the block.



Rule-5 :- Shifting the take off point after the point block.



Rule-6 :- Shifting the take off before the block.



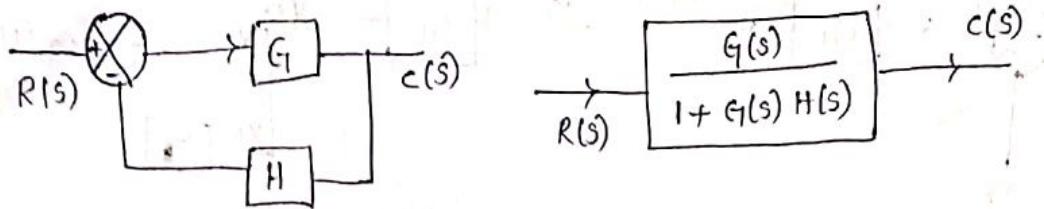
REMARKS :-

- * Summing point after the block
- * Summing point before the block
- * Take off point before the block
- * Take off point after the block

the block G
the block $1/G$
the block G
the block $1/G$

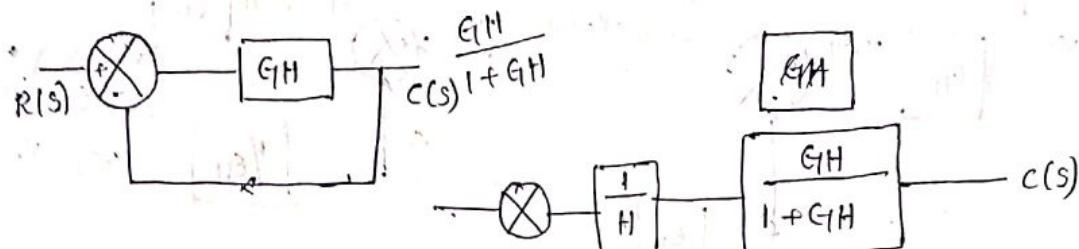
Rule - 7 :-

Transfer function for closed loop control system:



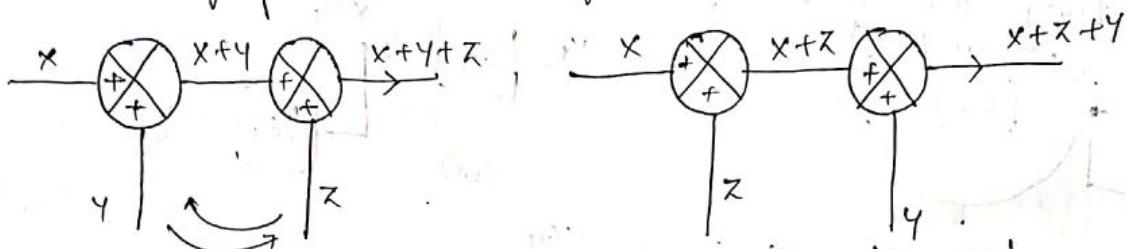
Rule - 8 :-

Block transformation / Interchanging the block.

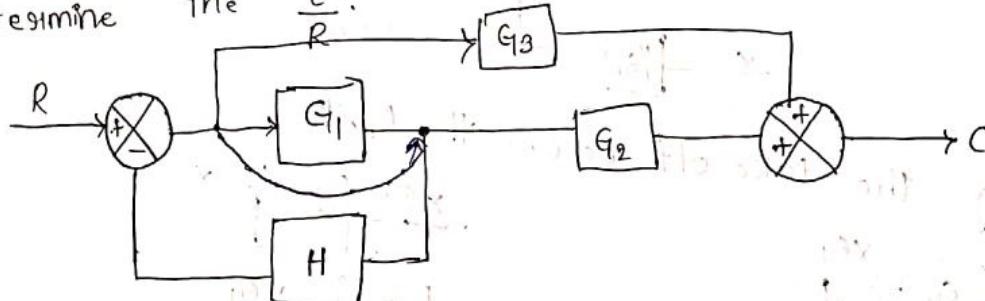


Rule - 9 :-

Interchanging the summing points :-

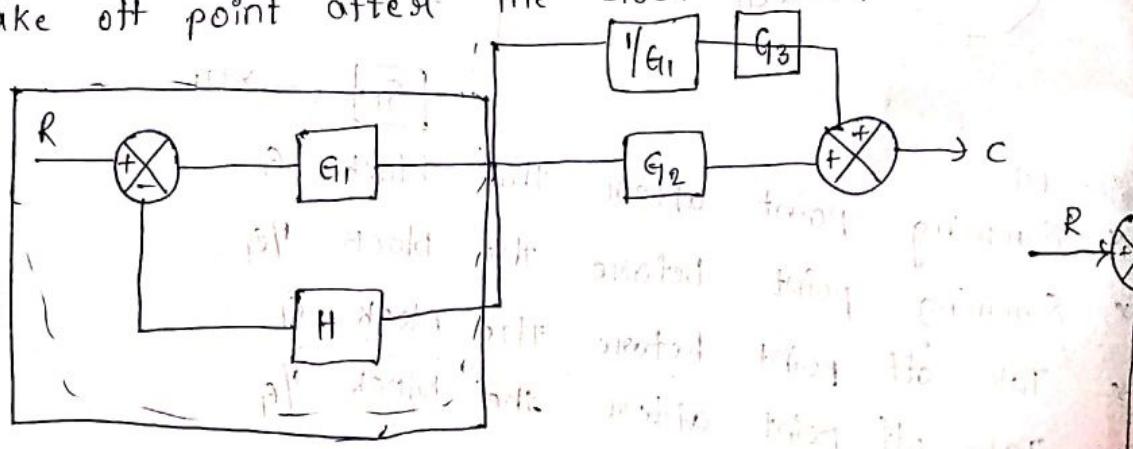


1. Reduce the block diagram shown in fig. and determine the $\frac{C}{R}$.

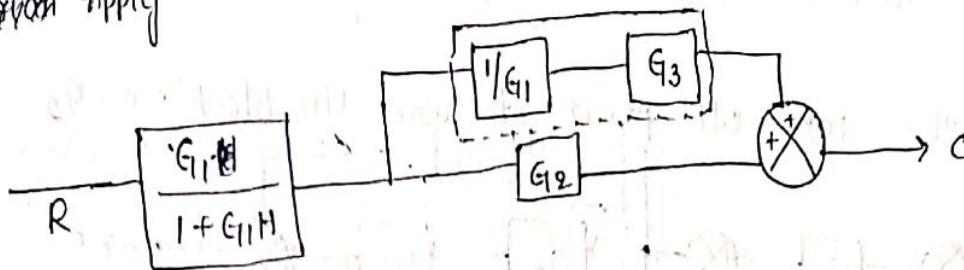


Step 1:

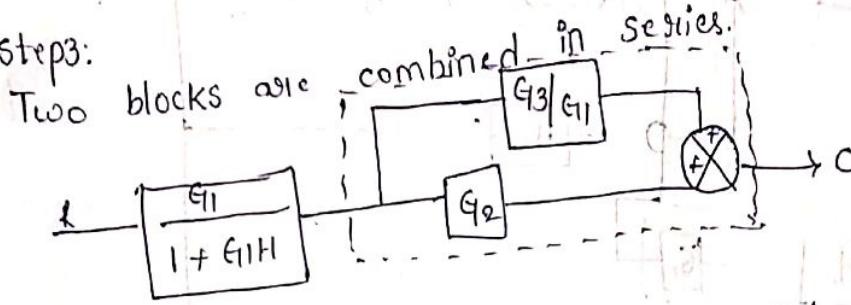
Take off point after the block i.e., G_1



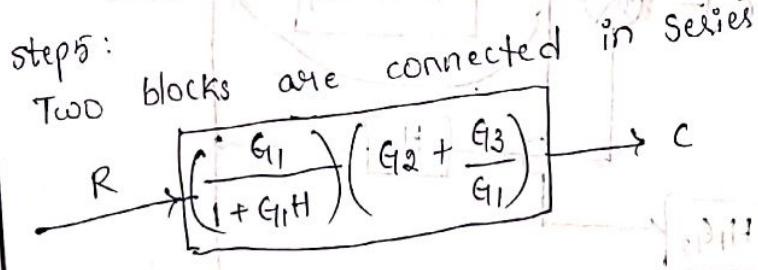
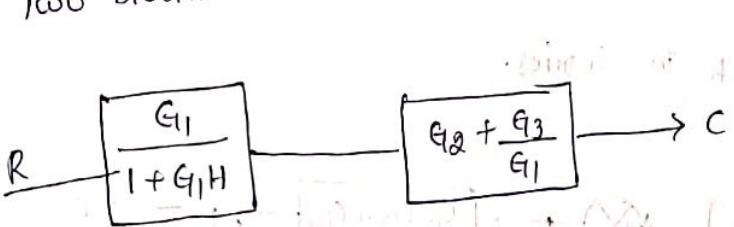
Step 2:
Now apply -ve feedback for closed loop CS transfer functions



Step 3:



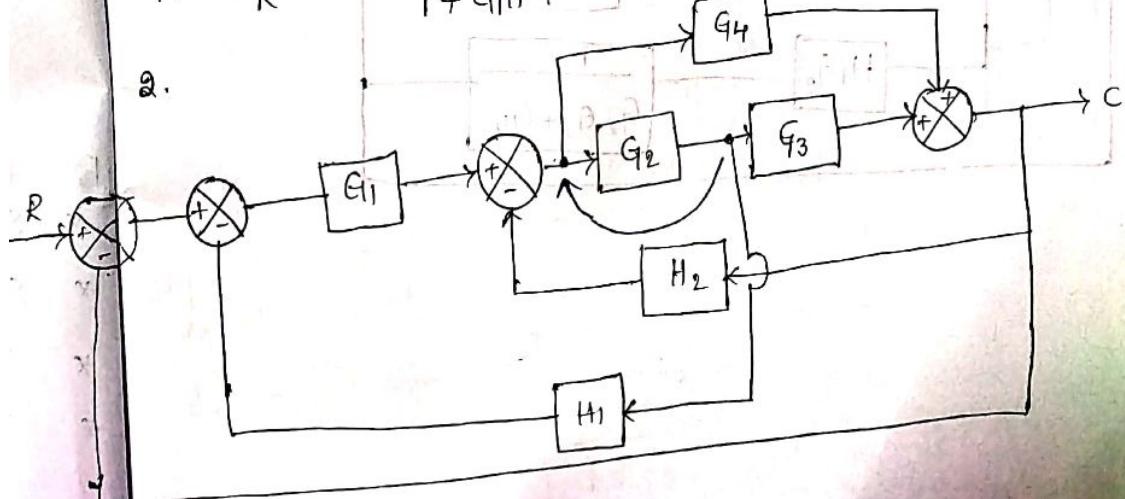
Step 4:



$$\left(\frac{G_1}{1+G_1H} \right) \left(G_2 + \frac{G_3}{G_1} \right) = \frac{C}{R}$$

$$\left(\frac{G_1}{1+G_1H} \right) \left(\frac{G_1G_2 + G_3}{G_1} \right) = \frac{C}{R}$$

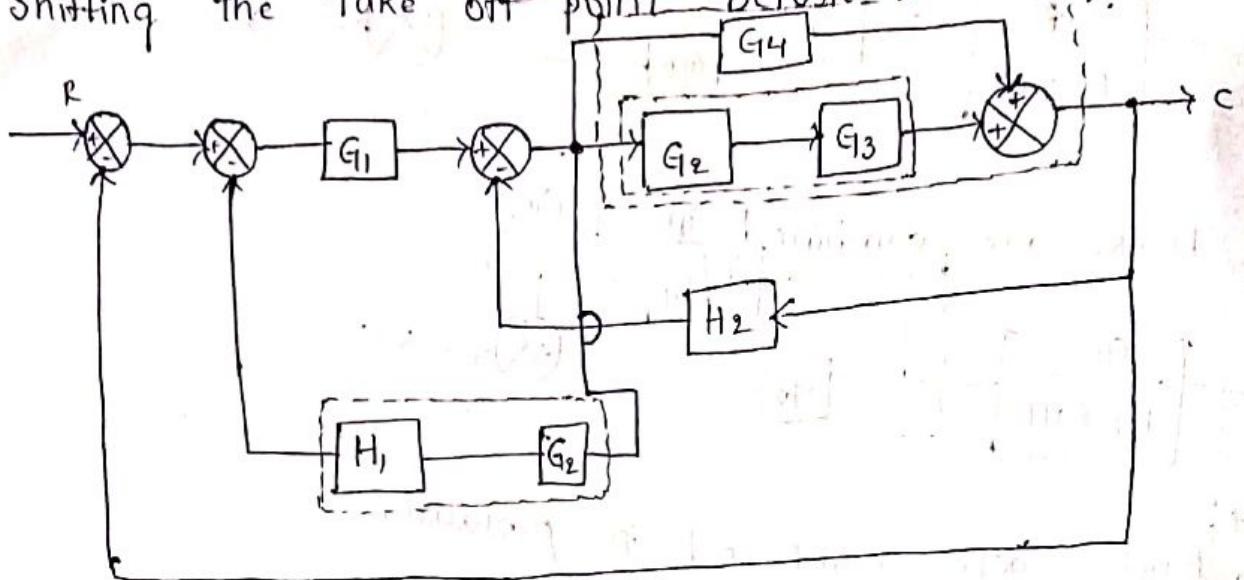
$$TF = \frac{C}{R} = \frac{G_1G_2 + G_3}{1+G_1H}$$



Using block diagram reduction technique, find closed loop transfer function for the given system.

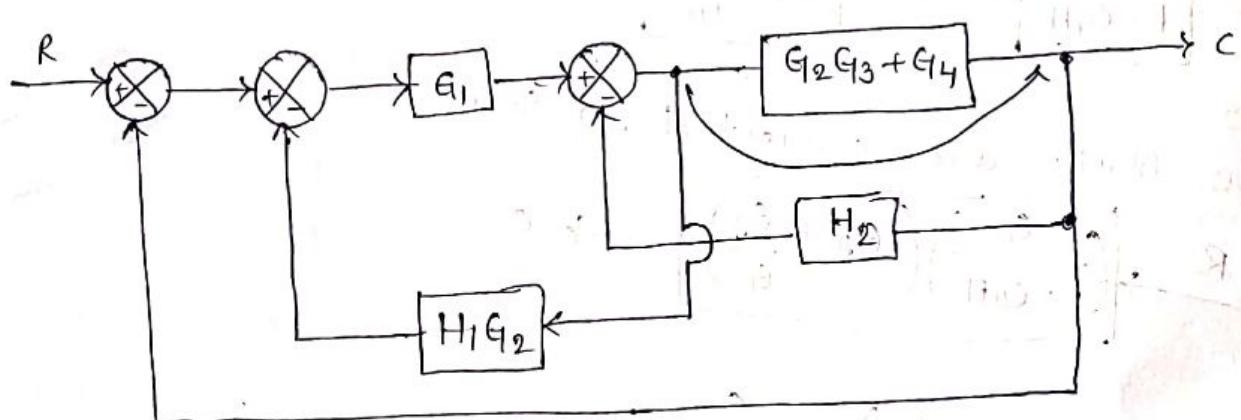
Step 1:

Shifting the take off point before the block i.e., G_2 .



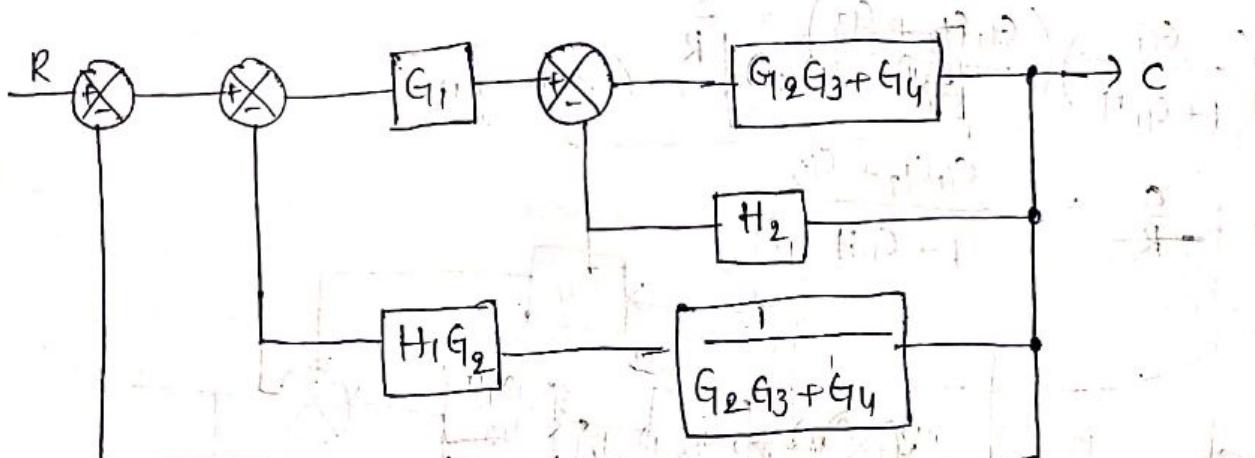
Step 2:

Combining the block in series.

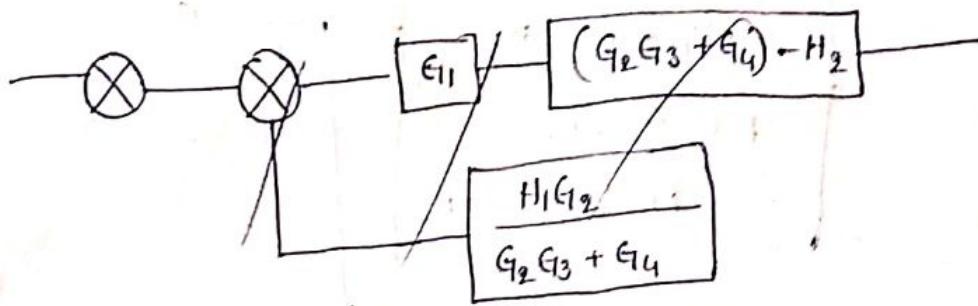


Step 3:

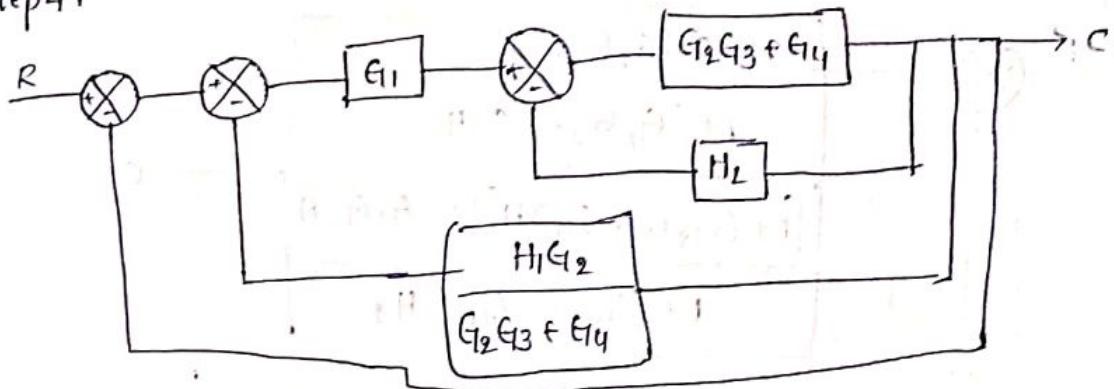
Shift the take off point after the block.



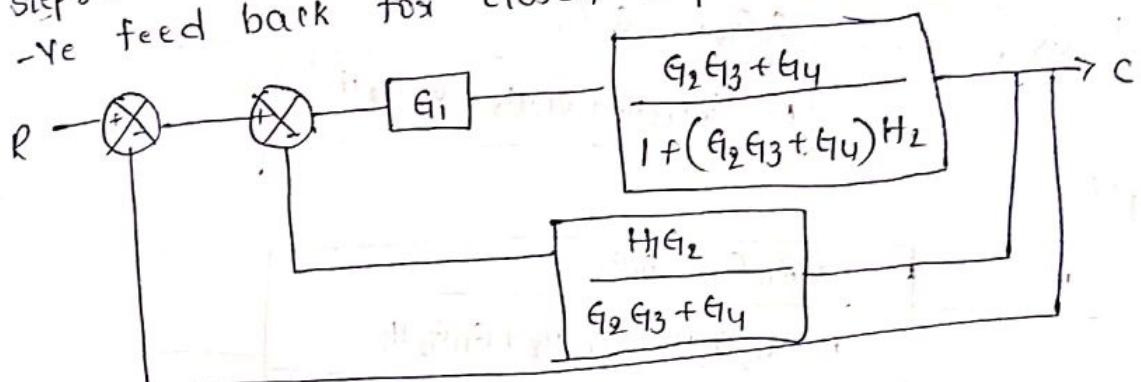
Step 4:



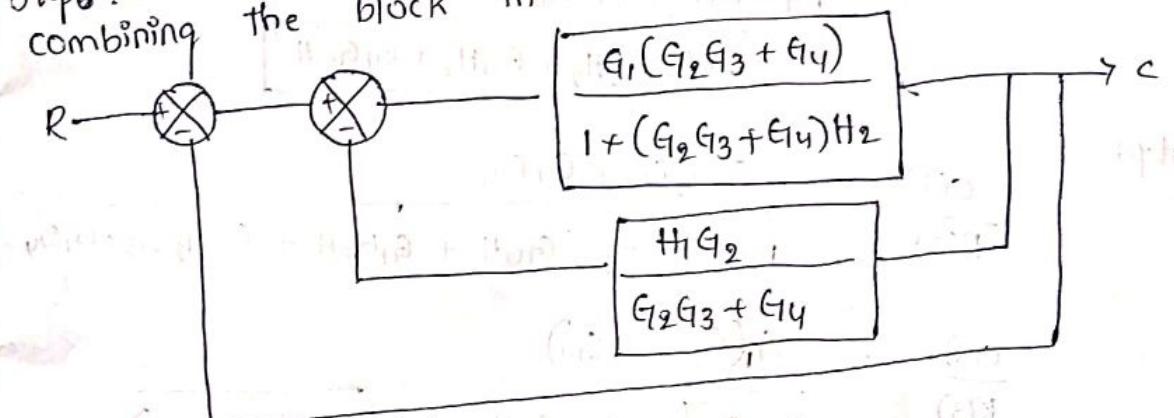
Step 4:



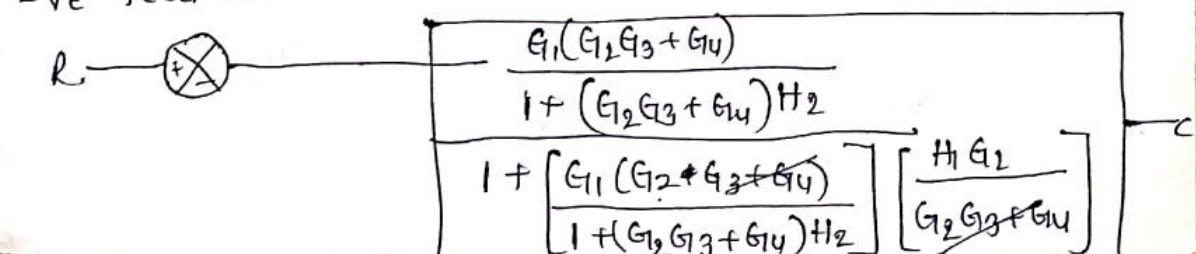
Step 5: -ve feed back for closed loop CS.



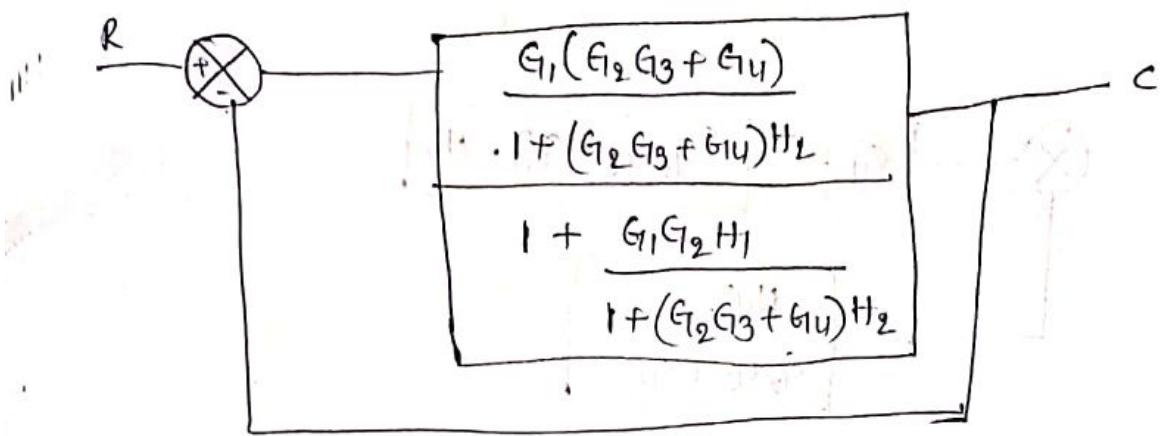
Step 6: combining the block in series.



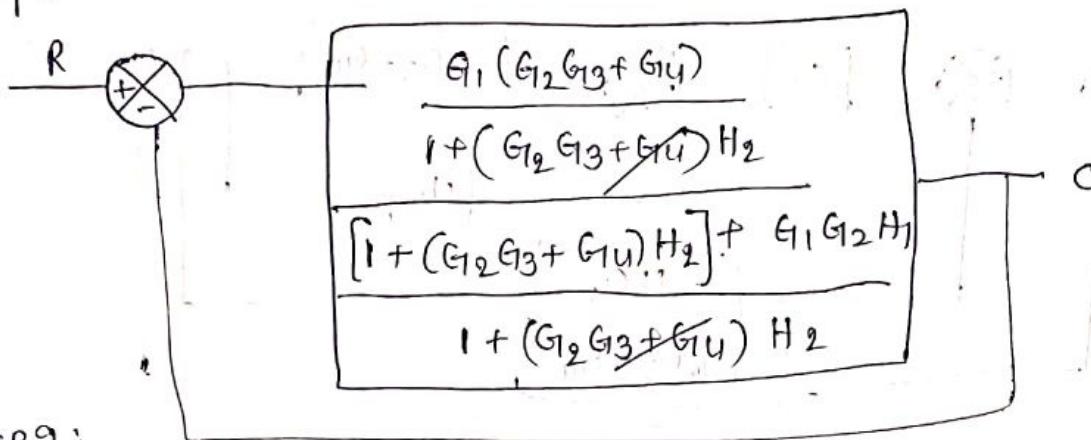
Step 7: -ve feed back CS



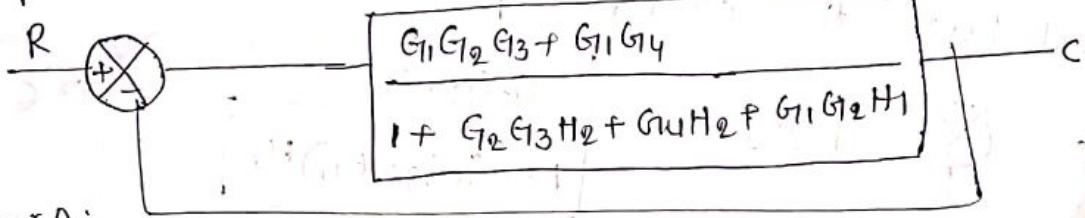
Step 7:



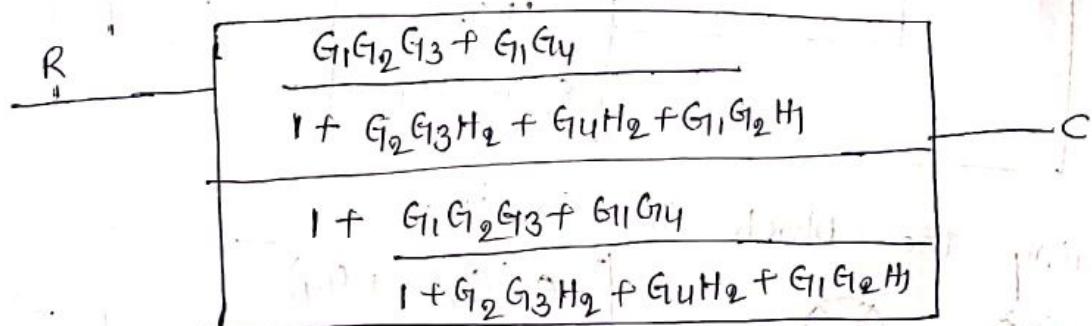
Step 8:



Step 9:



Step 10:

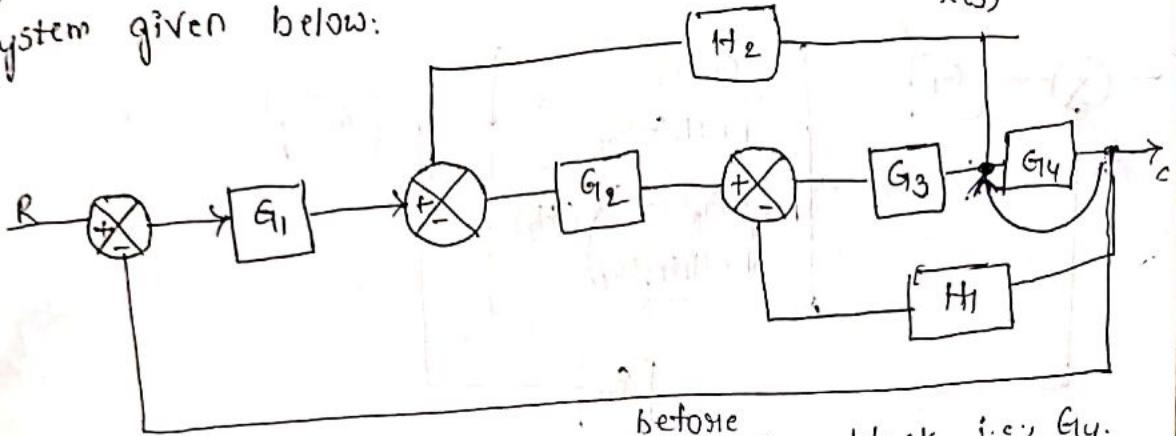


Step 11:

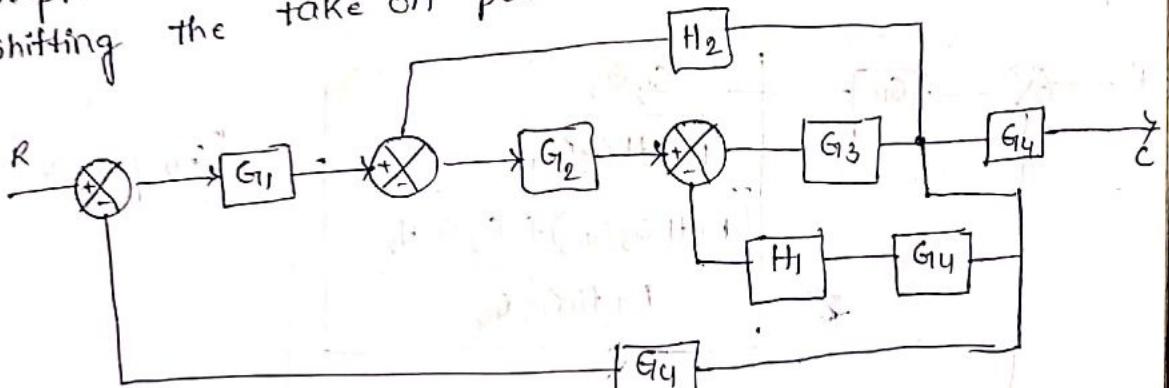
$$\frac{C(s)}{R(s)} = \frac{G_1G_2G_3 + G_1G_4}{1 + G_2G_3H_2 + G_4H_2 + G_1G_2H_1 + G_1G_2G_3 + G_1G_4}$$

$$\frac{C(s)}{R(s)} = \frac{G_1(G_2G_3 + G_4)}{1 + G_2(G_3H_2 + G_1H_1 + G_1G_3) + G_4(H_2 + G_1)}$$

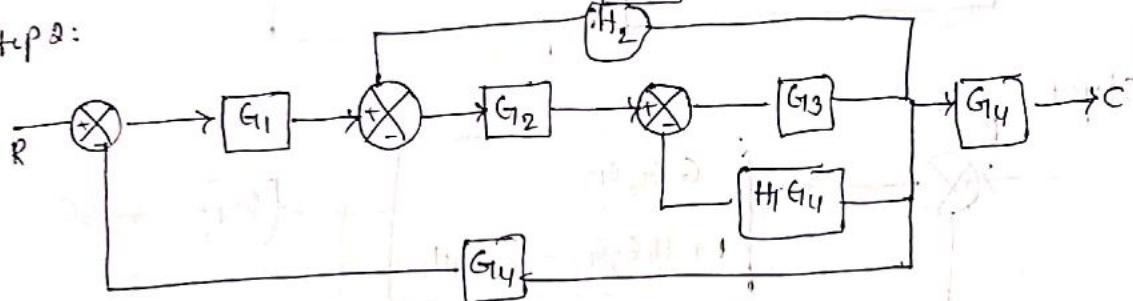
3) Determine the overall transfer function $\frac{C(s)}{R(s)}$ for the system given below:



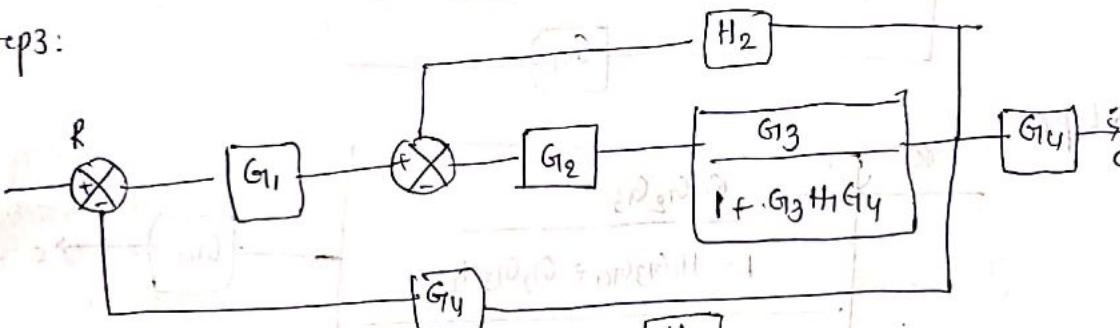
Step 1: shifting the take off point before the block. i.e., G_4 .



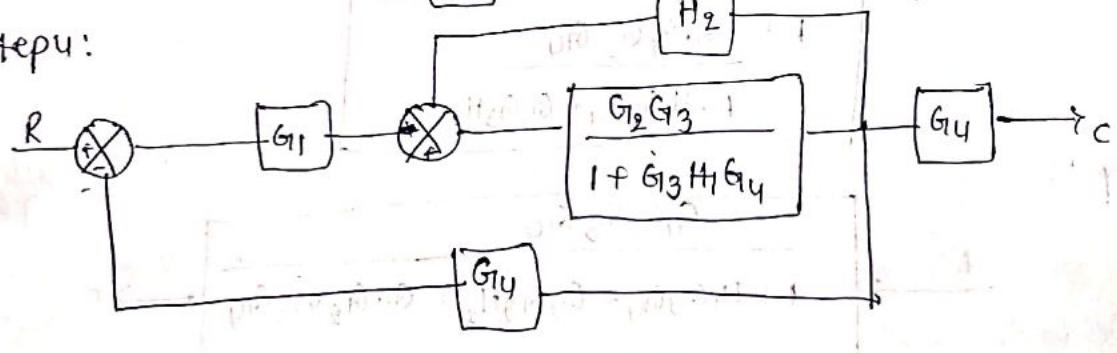
Step 2:



Step 3:

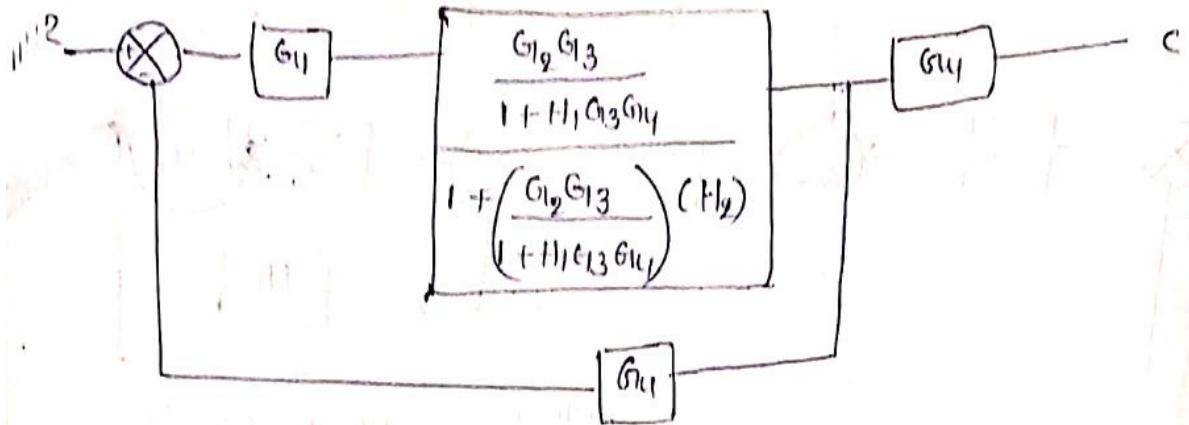


Step 4:

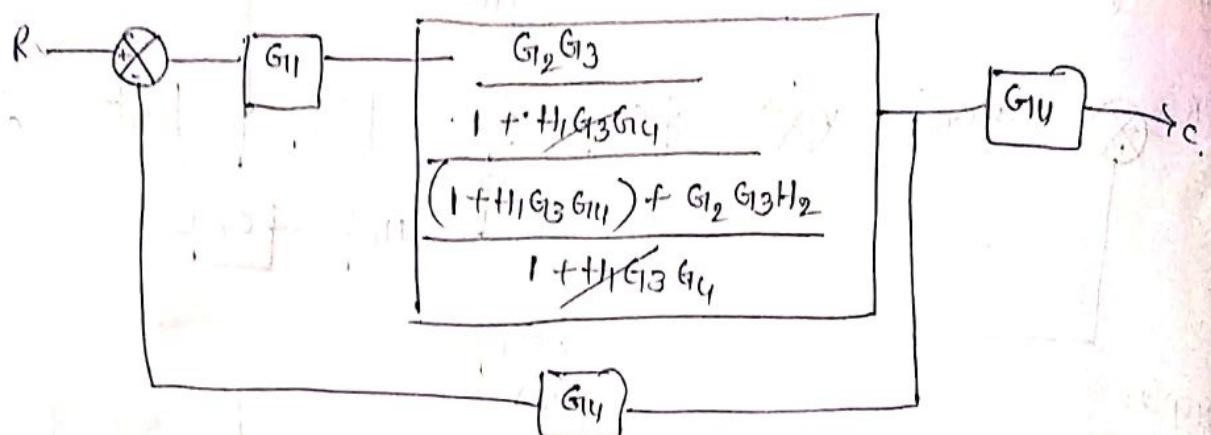


Step 5:

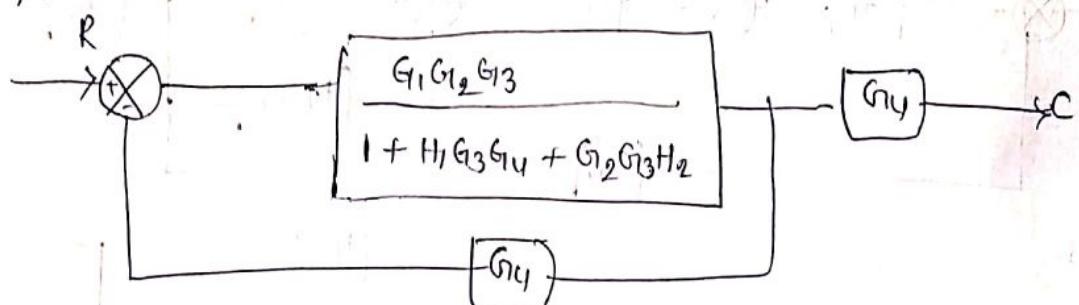




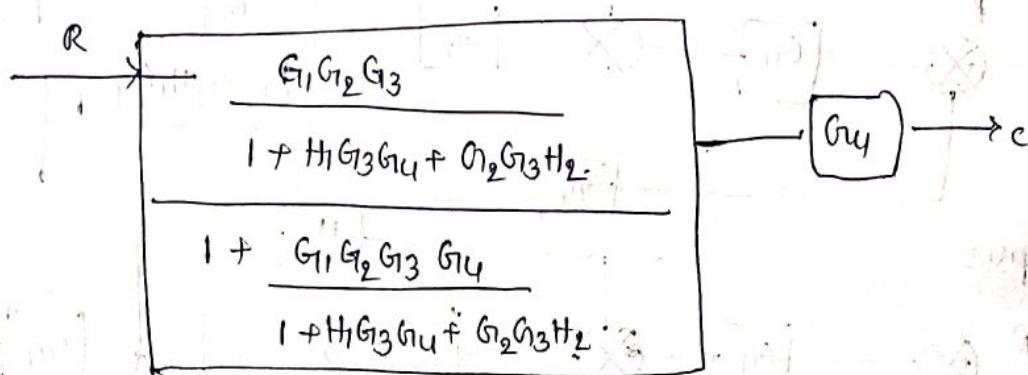
Step 6:



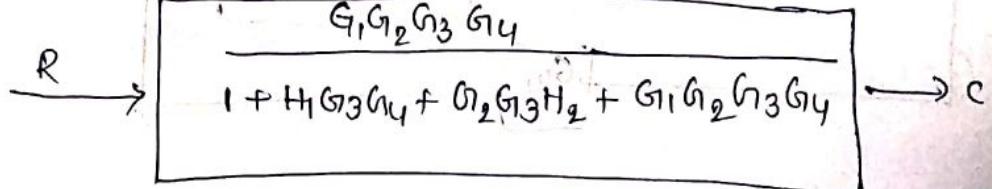
Step 7:



Step 8:



Step 9:

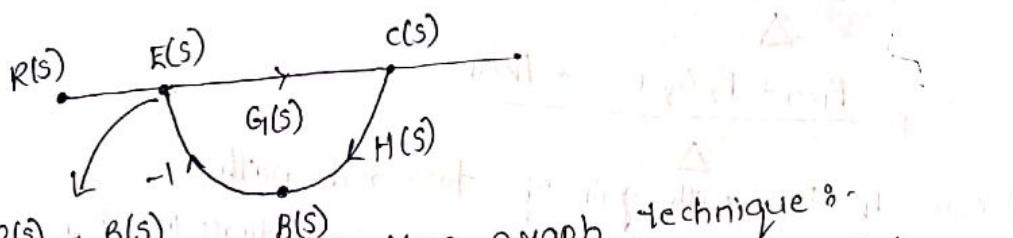
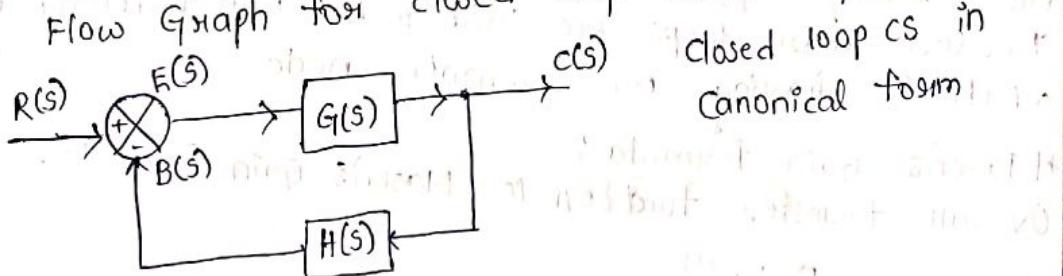


Step 10:

$$\frac{C(s)}{R(s)} = \frac{G_1 G_2 G_3 G_4}{1 + G_3 [H_1 G_4 + G_2 H_2 + G_1 G_2 G_4]}$$

Signal Flow Graph Method :-
 It is used to determine the transfer function of the system. In order to avoid the complexity existing in the block diagram reduction technique with the help of signal flow graph technique. It is graphical representation of control system in which it's node represents system variable and are connected direct branches.

Signal Flow Graph for closed loop Control System :-



Terminology for Signal flow graph technique :-
Node :- It represents system variable and it is having which is equal to the sum of all incoming signals at it.

Source Node / Input Node :- It is the node which is having outgoing branches.
Output Node / Sink Node :- It is the node which is having all incoming branches.

Mixed Node :- It is a node, which is having both incoming and outgoing branches.

Path :- It is the transversal of connecting branches in the direction of branch arrow such that node is traverse more than once.

Forward Path:

It is the path from ilp node to olp node.

Loop:

It is the path which originates and terminates at the same node.

Note:

- Self loop on ilp nodes are not valid loops and should not be considered when writing the transfer function.
- Loops (or) selfloops on olp nodes are valid loops while writing the transfer function.

Non touching loops:

Two (or) more loops are said to be non touching loops which is having no common node.

*Mason's Gain formula:-

Overall transfer function (or) Mason's Gain formula =

$$= \frac{\sum P_k \Delta_k}{\Delta}$$

$$= P_1 \Delta_1 + P_2 \Delta_2 + \dots + P_k \Delta_k$$

where P_k is path gain of forward path.

Δ is $1 - \{ \text{Sum of gain of individuals loops} \} +$

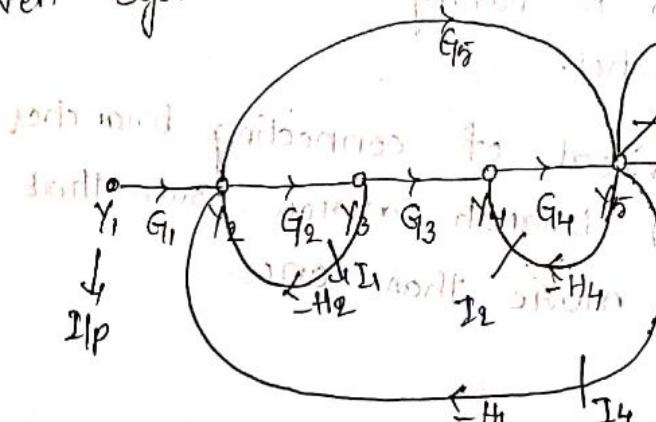
$\{ \text{Sum of gains of two non-touching loops} \} -$

$\{ \text{Sum of gains of three non-touching loops} \}$

Δ_k is the value of Δ obtained by removing all the loops touching the k^{th} forward path.

K = no. of forward paths

1. Determine the transfer function by using signal flow graph method (or) Mason's Gain formula for the given system



No. of forward paths (K) = 2

Mason's Gain formula = $\frac{\sum P_K \Delta_K}{\Delta}$

$$= \frac{P_1 \Delta_1 + P_2 \Delta_2}{\Delta}$$

$$P_1 = G_1 G_2 G_3 G_4 (1) = G_1 G_2 G_3 G_4$$

$$P_2 = G_1 G_5 (1) - G_1 G_5$$

No. of individual loops = 4

$$I_1 = -G_2 H_2$$

$$I_2 = -G_4 H_4$$

$$I_3 = -H_3$$

$$I_4 = -G_2 G_3 G_4 H_1$$

Non-touching loops (L):

$$L_1 = I_1 I_2 = (-G_2 H_2)(-G_4 H_4) = G_2 G_4 H_2 H_4$$

$$L_2 = I_1 I_3 = (-G_2 H_2)(-H_3) = G_2 H_2 H_3$$

$$\Delta = 1 - [-G_2 H_2 - G_4 H_4 - H_3 - G_2 G_3 G_4 H_1] + [G_2 G_4 H_2 H_4 + G_2 H_2 H_3] + 0$$

$$\Delta = 1 + G_2 H_2 + G_4 H_4 + H_3 + G_2 G_3 G_4 H_1 + G_2 G_4 H_2 H_4 + G_2 H_2 H_3$$

$$\Delta_1 = 1 - 0 = 1$$

$$\Delta_2 = 1 - 0 = 1$$

$$\Delta_1 = \Delta_2 = 1$$

Mason's Gain formula = $\frac{\sum P_K \Delta_K}{\Delta}$

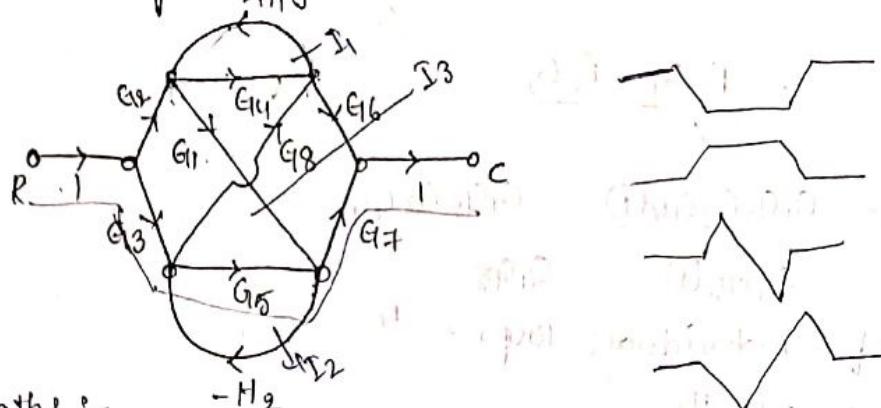
$$= \frac{P_1 \Delta_1 + P_2 \Delta_2}{\Delta}$$

$$= \frac{G_1 G_2 G_3 G_4 (1) + G_1 G_5 (1)}{1 + G_2 H_2 + G_4 H_4 + H_3 + G_2 G_3 G_4 H_1 + G_2 G_4 H_2 H_4 + G_2 H_2 H_3}$$

$$= \frac{G_1 (G_2 G_3 G_4 + G_5)}{1 + G_2 H_2 (1 + G_4 H_4 + H_3) + G_4 H_4 + H_3 + G_2 G_3 G_4 H_1}$$

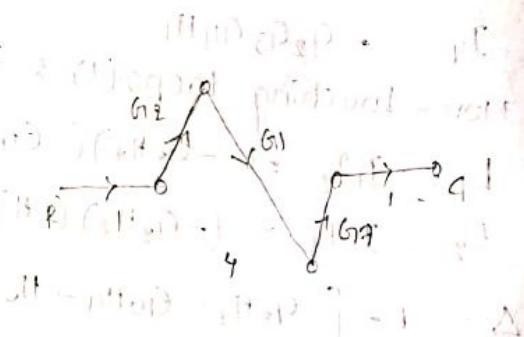
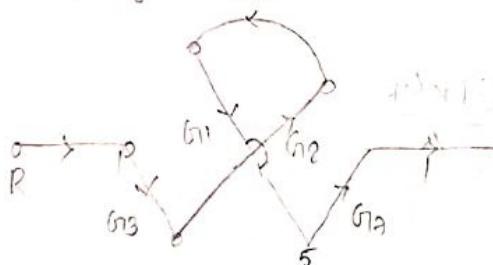
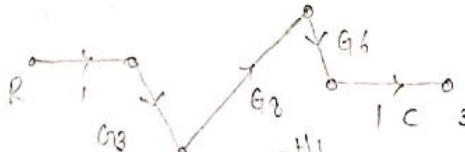
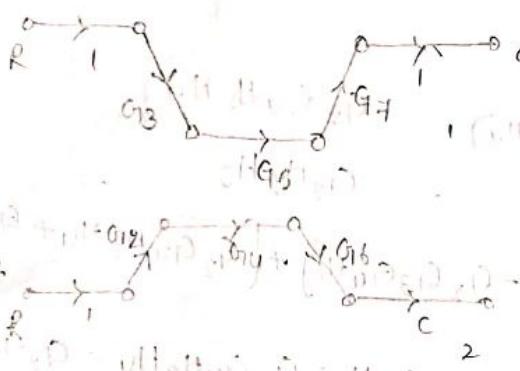
Q) Det

Q) Determine the transfer function by using signal flow graph method for the system given below:



Forward paths:

No. of forward paths (K) = 6



By observing the given system, we have six forward paths i.e., $K=6$.

$$P_1 = (1) G_3 G_5 G_1 G_7 (1) = G_3 G_5 G_1 G_7$$

$$P_2 = (1) G_3 G_5 G_1 G_7 (1) = G_3 G_5 G_1 G_7$$

$$P_3 = (1) G_3 G_5 G_1 G_7 (1) = G_3 G_5 G_1 G_7$$

$$P_4 = (1) G_3 G_5 G_1 G_7 (1) = G_3 G_5 G_1 G_7$$

$$P_5 = (1) G_3 G_5 G_1 G_7 G_2 G_4 G_9 (1) = -G_3 G_5 G_1 G_7 G_2 G_4 G_9$$

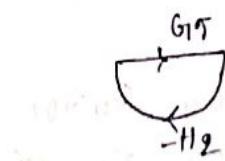
$$P_6 = (1) G_3 G_5 G_1 G_7 G_2 G_4 G_9 G_6 (1) = -G_3 G_5 G_1 G_7 G_2 G_4 G_9 G_6$$

Individual loops (I)

$$I_1 = G_{14}(-H_1) = -G_{14}H_1$$



$$I_2 = -G_{15}H_2$$



$$I_3 = G_{11}(-H_2)G_{18}(-H_1) = G_{11}G_{18}H_1H_2$$

Non touching loops (L):

$$L = I_1 I_2 = (-G_{14}H_1)(-G_{15}H_2) = G_{14}G_{15}H_1H_2$$

$$\Delta_1 = 1 - \{I_1\} + 0 = 1 + G_{14}H_1$$

$$\Delta_2 = 1 - I_2 = 1 + G_{15}H_2$$

$$\Delta_3 = 1 - 0 = 1$$

$$\Delta_4 = 1 - 0 = 1$$

$$\Delta_5 = 1 - 0 = 1$$

$$\Delta_6 = 1 - 0 = 1$$

$$\therefore \Delta_3 = \Delta_4 = \Delta_5 = \Delta_6 = 1$$

$$\Delta = 1 - (I_1 + I_2 + I_3) + (L)$$

$$\Delta = 1 - (-G_{14}H_1 - G_{15}H_2 + G_{11}G_{18}H_1H_2) + G_{14}G_{15}H_1H_2$$

$$\Delta = 1 + G_{14}H_1 + G_{15}H_2 - G_{11}G_{18}H_1H_2 + G_{14}G_{15}H_1H_2$$

wkT the Mason's Gain formula: $\sum_{k=1}^{\Delta} P_k \Delta_k$

$$= \frac{P_1 \Delta_1 + P_2 \Delta_2 + P_3 \Delta_3 + P_4 \Delta_4 + P_5 \Delta_5 + P_6 \Delta_6}{\Delta}$$

$$= (G_{13}G_{15}G_{17})(1 + G_{14}H_1) + (G_{12}G_{14}G_{16})(1 + G_{15}H_2) + G_{13}G_{18}G_{16} +$$

$$G_{12}G_{11}G_{17} + (-G_{13}G_{18}G_{11}G_{17}H_1) - G_{12}G_{11}H_2G_{18}G_{16}$$

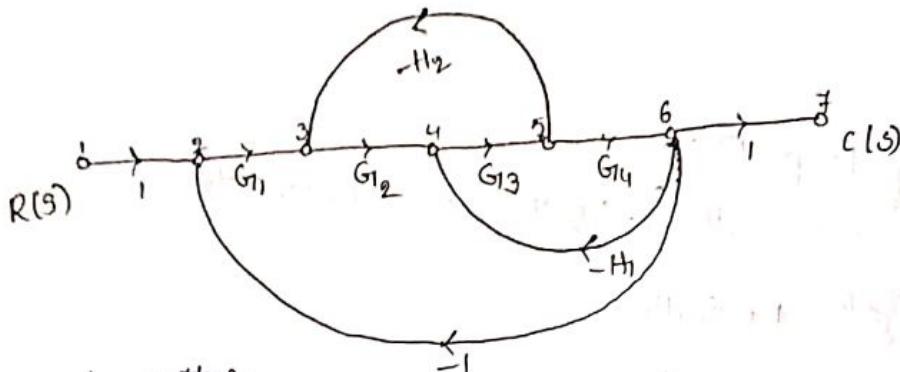
$$1 + G_{14}H_1 + G_{15}H_2 - G_{11}G_{18}H_1H_2 + G_{14}G_{15}H_1H_2$$

$$= G_{13}G_{15}G_{17} + G_{13}G_{14}G_{15}G_{17}H_1 + G_{12}G_{14}G_{16} + G_{12}G_{14}G_{15}G_{16}H_2 + G_{13}G_{16}G_{18} + G_{11}G_{12}G_{17} - G_{11}G_{13}G_{17}G_{18}H_1 - G_{11}G_{12}G_{16}G_{18}H_2$$

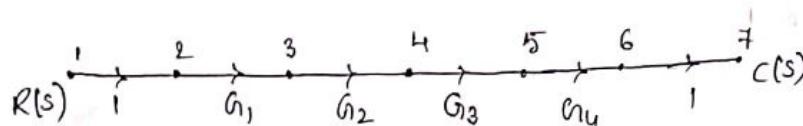
$$1 + G_{14}H_1 + G_{15}H_2 - G_{11}G_{18}H_1H_2 + G_{14}G_{15}H_1H_2$$

$$\begin{aligned}
 & G_3 G_{15} G_{17} (1 + G_{11} H_1) + G_{12} G_{14} G_{16} (1 + G_{15} H_2) + G_{13} G_8 (G_{16} - G_1 G_7 H_1) \\
 & + G_{11} G_{12} (G_{17} - G_{16} G_8 H_2) \\
 & 1 + G_{11} H_1 + G_{15} H_2 - H_1 H_2 (G_{11} G_8 - G_{16} G_5)
 \end{aligned}$$

3) Find the overall gain $\frac{C(s)}{R(s)}$ for the given system



Forward path:

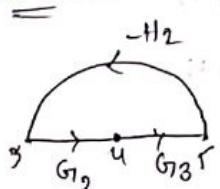


No. of forward paths = 1

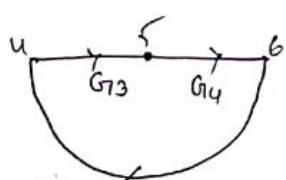
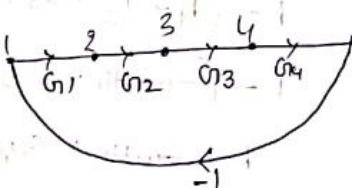
$$K = 1$$

$$P = (1) G_1 G_2 G_3 G_4 H_1 = G_1 G_2 G_3 G_4$$

Individual Loops:

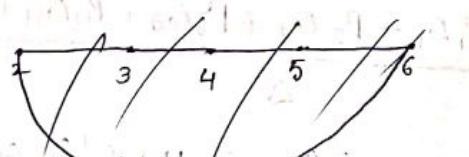


$$I_1 = -G_2 G_3 H_2$$



$$I_2 = -G_3 G_4 H_1$$

$$I_3 = -G_1 G_2 G_3 G_4$$



There are no non-touching loops.

$$\Delta = 1 - (I_1 + I_2 + I_3)$$

$$\Delta = 1 - [-G_2 G_3 H_2 - G_3 G_4 H_1 - G_1 G_2 G_3 G_4]$$

$$\Delta = 1 + G_2 G_3 H_2 + G_3 G_4 H_1 + G_1 G_2 G_3 G_4$$

$$\Delta = 1 - 0 \approx 1$$

By Mason's Gain formula, $\frac{\sum P_k \Delta_k}{\Delta}$

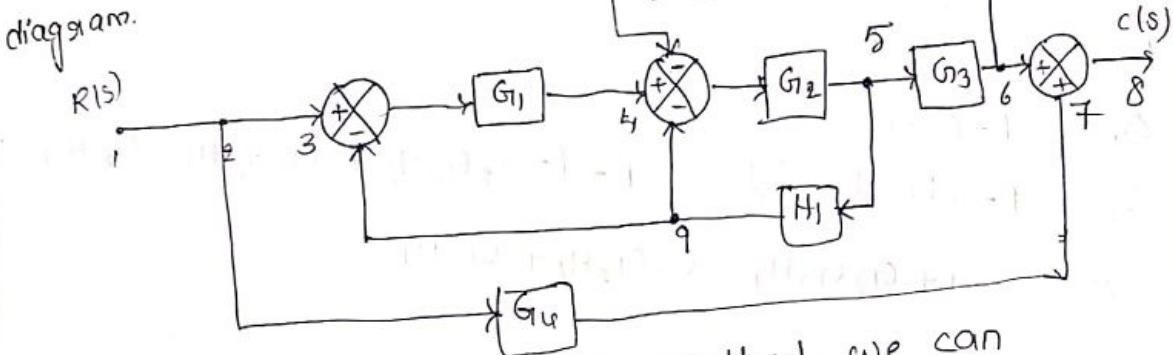
$$\cdot \frac{P_1 \Delta_1}{\Delta}$$

$$= \frac{G_1 G_{12} G_{13} G_{14} u(1)}{1 + G_{12} G_{13} H_2 + G_{13} G_{14} H_1 + G_1 G_{12} G_{13} G_{14}}$$

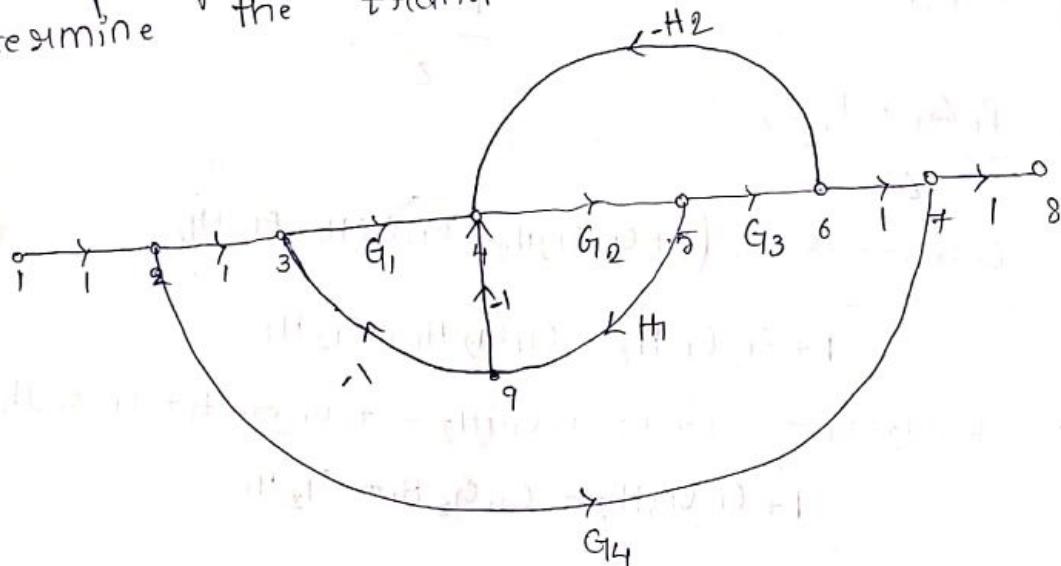
$$- \frac{G_1 G_{12} G_{13} G_{14}}{1 + G_{12} (G_{13} H_2 + G_1 G_{13} G_{14}) + G_3 G_{14} H_1}$$

Feed back characteristics and its advantages, linearizing
Nagarkar & Gopal
effect of feed back:

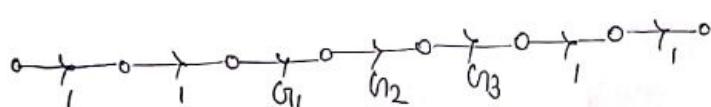
* Determine the transfer function for the given block diagram.



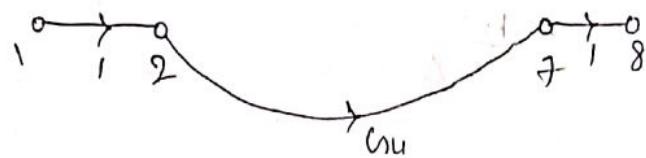
By using signal flow graph method, we can determine the transfer function.



Forward paths :-



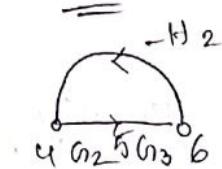
$$P_1 = G_1 G_2 G_3$$



$$P_2 = G_4$$

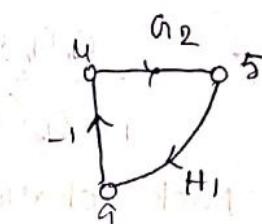
No. of forward paths (K) = 2

Individual loops (I) :-



$$I_1 = -G_2 G_3 H_2$$

$$I_2 = G_1 G_2 H_1$$



$$I_3 = -G_2 H_1$$

Non touching loops (L) :-

$$L = 0.$$

$$\Delta_1 = 1 - 0 = 1$$

$$\Delta_2 = 1 - (I_1 + I_2 + I_3) = 1 - (-G_2 G_3 H_2 + G_1 G_2 H_1 - G_2 H_1)$$

$$\Delta_2 = 1 + G_2 G_3 H_2 - G_1 G_2 H_1 + G_2 H_1$$

$$\Delta = \Delta_2$$

By Mason's gain formula, $\frac{\sum_{K=1}^2 P_K \Delta_K}{\Delta}$

$$= \frac{P_1 \Delta_1 + P_2 \Delta_2}{\Delta}$$

$$= G_1 G_2 G_3 + G_4 (1 + G_2 G_3 H_2 - G_1 G_2 H_1 + G_2 H_1)$$

$$1 + G_2 G_3 H_2 - G_1 G_2 H_1 + G_2 H_1$$

$$= \frac{G_1 G_2 G_3 + G_4 + G_2 G_3 G_4 H_2 - G_1 G_2 G_4 H_1 + G_2 G_4 H_1}{1 + G_2 G_3 H_2 - G_1 G_2 H_1 + G_2 H_1}$$

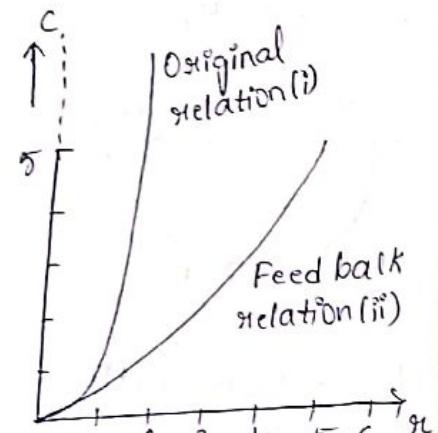
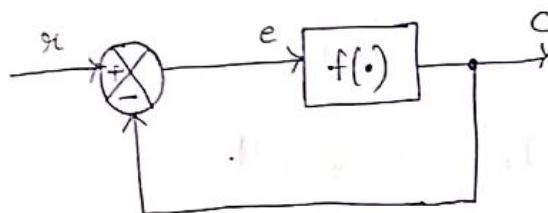
Linearizing effect of feed back :-
 Yet another property of feedback is its linearizing effect which is illustrated by means of the simple single-loop static system of fig (a). In a static system, various gains are independent of time. we shall assume that the forward block function is nonlinear expresses as

$e = f(e) \rightarrow e^2$, square law function
 when the feed back is open,

$e = sI \Rightarrow C = sI^2$
 which is plotted in fig (b). On the other hand when the loop is closed, we have

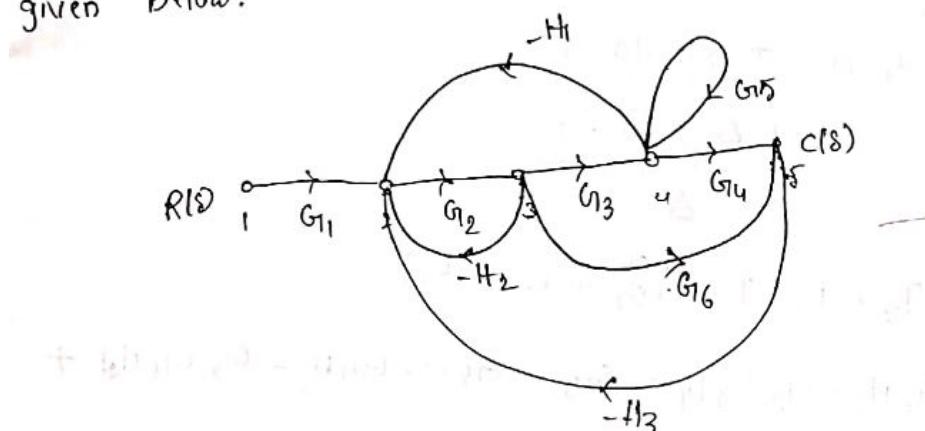
$$e = sI - C$$

$$\text{and so } e = f(e) = (sI - C)^2.$$

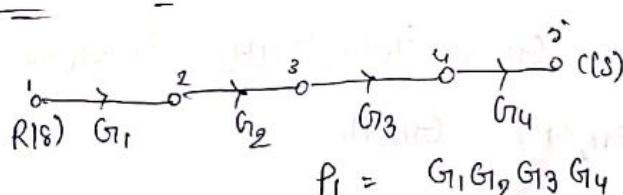


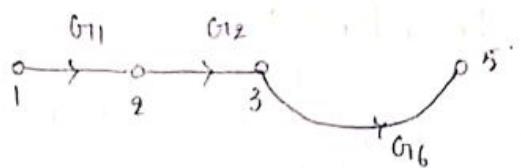
It is easily seen by comparison of the graphs (i) and (ii) that the input-output relation $[C(s)]$ is approximately linear over a much wider range for the closed-loop system compared to its open-loop behaviour

Find the overall gain $\frac{C(s)}{R(s)}$ for the signal flow graph given below:



Forward path :-

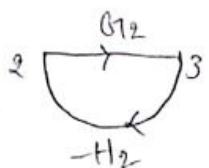




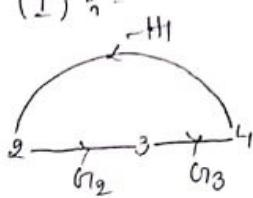
$$P_2 = G_{11} G_{12} G_{16}$$

No. of forward paths (K) = 2

Individual loops: (I) :-



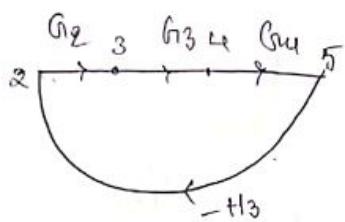
$$I_1 = -G_{12} H_2$$



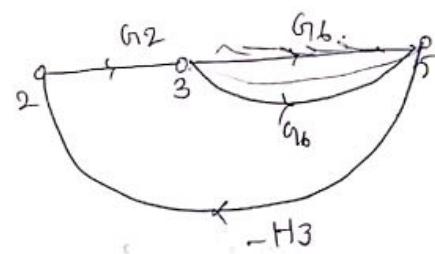
$$I_2 = -G_{12} G_{13} H_1$$



$$I_3 = G_{15}$$



$$I_4 = -G_{12} G_{13} G_{14} H_3$$



$$I_5 = -G_{12} G_{16} H_3$$

Non touching loops (L) :-

$$L_1 = I_1 I_3 = -G_{12} G_{15} H_2$$

$$L_2 = I_3 I_5 = -G_{15} H_3$$

$$\Delta_1 = 1 - 0 = 1$$

$$\Delta_2 = 1 - G_{15}$$

By Mason's Gain formula,

$$\frac{\sum P_K \Delta_K}{\Delta} = \frac{P_1 \Delta_1 + P_2 \Delta_2}{\Delta}$$

$$\Delta = 1 - (I_1 + I_2 + I_3 + I_4 + I_5) + (L_1 + L_2)$$

$$= 1 - (-G_{12} H_2 - G_{12} G_{13} H_1 + G_{15} - G_{12} G_{13} G_{14} H_3 - G_{12} G_{16} H_3) + (-G_{12} G_{15} H_2 - G_{15} H_3)$$

$$= 1 + G_{12} H_2 + G_{12} G_{13} H_1 - G_{15} + G_{12} G_{13} G_{14} H_3 + G_{12} G_{16} H_3 - G_{12} G_{15} H_2 - G_{15} H_3$$

$$\frac{P_1 \Delta_1 + P_2 \Delta_2}{\Delta}$$

$$\Rightarrow \frac{G_1 G_{12} G_{13} G_{14} + G_1 G_{12} G_{16} (1 - G_{15})}{1 + G_{12} H_2 + G_{12} G_{13} H_1 - G_{15} + G_{12} G_{13} G_{14} H_3 + G_{12} G_{16} H_3 - G_2 G_{15} H_2 - G_{15} H_3}$$

$$\Rightarrow \frac{G_1 G_{12} (G_{13} G_{14} + G_{16} - G_{15} G_{16})}{1 + G_{12} H_2 (1 - G_{15}) + G_{12} G_{13} (H_1 + G_{14} H_3) - G_{15} + G_{12} G_{16} H_3 - G_{15} H_3}$$