

Unit 1

Electromechanical Energy Conversion Principles

Introduction:

Electrical energy is seldom available naturally and is rarely directly utilized. There are two conversion takes place-----

- a. One form to electrical form
- b. Electrical form to original form or any other desired form

The device through which we convert one form to electrical form & back to original form or any other desired form is studied in EMEC.

Like—Transformers, D.C. Machines, A. C. Machines (Induction and Synchronous)

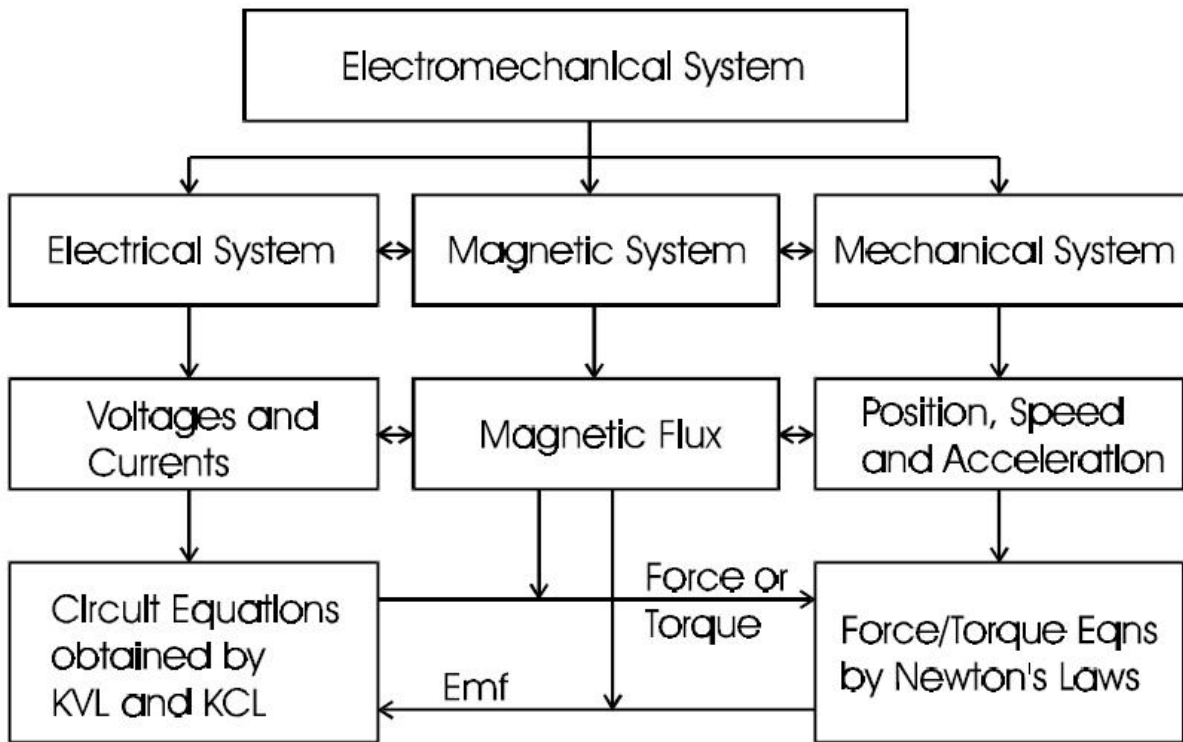
These devices can be transducers for low energy conversion processing and transporting. A second category of such devices is meant for production of force or torque with limited mechanical motion like electromagnets, relays, actuators etc.

A third category is the continuous energy conversion devices like motors or generators which are used for bulk energy conversion and utilization.

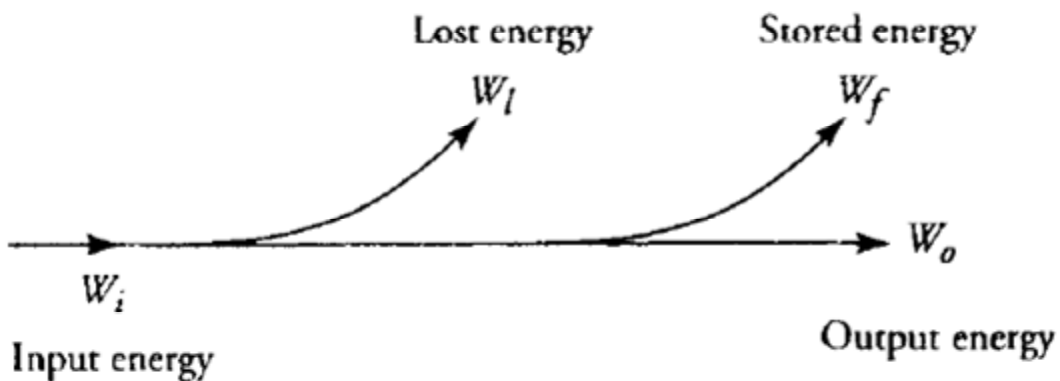
EMEC-----via-----Medium of magnetic or electric field. For practical devices magnetic medium is most suitable.

When we speak of electromechanical energy conversion, however, we mean either the conversion of electric energy into mechanical energy or vice versa.

Electromechanical energy conversion is a reversible process except for the losses in the system. The term "reversible" implies that the energy can be transferred back and forth between the electrical and the mechanical systems.



Concept map of electromechanical system modeling



Energy Flow Diagram

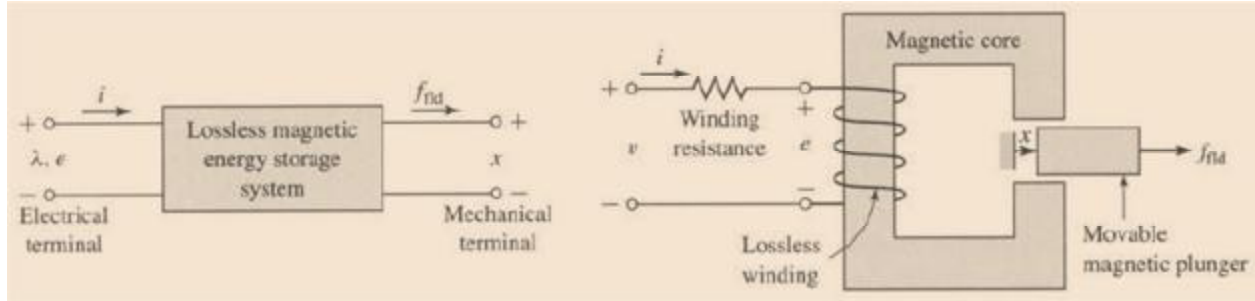
From energy diagram we can see that principle of energy conservation is accurately followed. i.e Input Energy=Losses + Stored Energy + Output Energy.

Singly Excited System:

Consider a singly excited linear actuator as shown below. The winding resistance is R . At a certain time instant t , we record that the terminal voltage applied to the excitation winding is v ,

the excitation winding current i , the position of the movable plunger x , and the force acting on the plunger F with the reference direction chosen in the positive direction of the x axis, as shown in the diagram. After a time interval dt , we notice that the plunger has moved for a distance dx under the action of the force F . The mechanical done by the force acting on the plunger during this time interval is thus

$$dw_m = Fdx$$



Singly Excited system energy conversion

The amount of electrical energy that has been transferred into the magnetic field and converted into the mechanical work during this time interval can be calculated by subtracting the power loss dissipated in the winding resistance from the total power fed into the excitation winding as

$$dw_e = dw_f + dw_m = vidt - Ri^2 dt$$

Since,

$$e = \frac{d\lambda}{dt} = v - Ri$$

So,

$$dw_f = dw_e - dw_m = eidt - Fdx = id\lambda - Fdx$$

we can also write,

$$e = \frac{d\lambda}{dt} = v - Ri$$

$$dw_f(\lambda, x) = \frac{dw_f(\lambda, x)}{d\lambda} d\lambda + \frac{dw_f(\lambda, x)}{dx} dx$$

the energy stored in a magnetic field can be expressed as

$$w_f(\lambda, x) = \int_0^\lambda i(\lambda, x) d\lambda$$

For a magnetically linear (with a constant permeability or a straight line magnetization curve such that the inductance of the coil is independent of the excitation current) system, the above expression becomes

$$W_f(\lambda, x) = \frac{1}{2} \frac{\lambda^2}{L(x)}$$

and the force acting on the plunger is then

$$F = -\frac{\partial W_f(\lambda, x)}{\partial x} = \frac{1}{2} \left[\frac{\lambda}{L(x)} \right]^2 \frac{dL(x)}{dx} = \frac{1}{2} i^2 \frac{dL(x)}{dx}$$

In the diagram below, it is shown that the magnetic energy is equivalent to the area above the magnetization or λ - i curve. Mathematically, if we define the area underneath the magnetization curve as the *coenergy* (which does not exist physically), i.e.

$$W_f'(i, x) = i\lambda - W_f(\lambda, x)$$

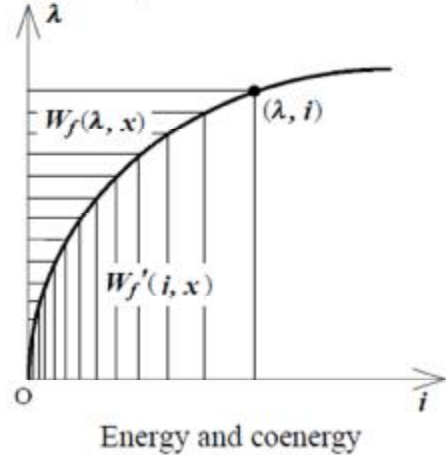
we can obtain

$$\begin{aligned} dW_f'(i, x) &= \lambda di + i d\lambda - dW_f(\lambda, x) \\ &= \lambda di + F dx \\ &= \frac{\partial W_f'(i, x)}{\partial i} di + \frac{\partial W_f'(i, x)}{\partial x} dx \end{aligned}$$

Therefore,

$$\lambda = \frac{\partial W_f'(i, x)}{\partial i}$$

and
$$F = \frac{\partial W_f'(i, x)}{\partial x}$$



From the above diagram, the coenergy or the area underneath the magnetization curve can be calculated by

$$W_f'(i, x) = \int_0^i \lambda(i, x) di$$

For a magnetically linear system, the above expression becomes

$$W_f'(i, x) = \frac{1}{2} i^2 L(x)$$

and the force acting on the plunger is then

$$F = \frac{\partial W_f'(i, x)}{\partial x} = \frac{1}{2} i^2 \frac{dL(x)}{dx}$$

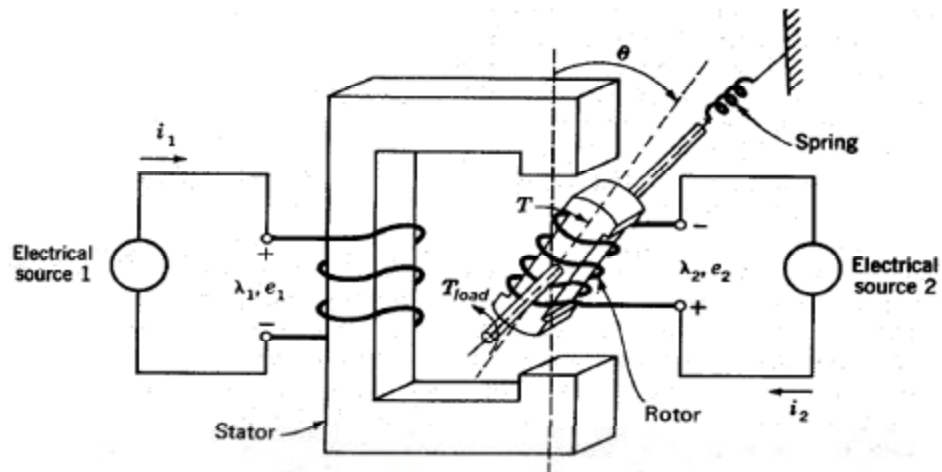
Doubly Excited Rotating Actuator

The general principle for force and torque calculation discussed above is equally applicable to multi-excited systems. Consider a doubly excited rotating actuator shown schematically in the diagram below as an example. The differential energy and coenergy functions can be derived as following:

$$dW_f = dW_e - dW_m$$

where

$$dW_e = e_1 i_1 dt + e_2 i_2 dt$$



A doubly excited actuator

$$e_1 = \frac{d\lambda_1}{dt}, \quad e_2 = \frac{d\lambda_2}{dt}$$

and

$$dW_m = T d\theta$$

Hence,

$$dW_f(\lambda_1, \lambda_2, \theta) = i_1 d\lambda_1 + i_2 d\lambda_2 - T d\theta$$

$$\begin{aligned}
&= \frac{\partial W_f(\lambda_1, \lambda_2, \theta)}{\partial \lambda_1} d\lambda_1 + \frac{\partial W_f(\lambda_1, \lambda_2, \theta)}{\partial \lambda_2} d\lambda_2 \\
&+ \frac{\partial W_f(\lambda_1, \lambda_2, \theta)}{\partial \theta} d\theta
\end{aligned}$$

and

$$\begin{aligned}
dW_f'(i_1, i_2, \theta) &= d[i_1 \lambda_1 + i_2 \lambda_2 - W_f(\lambda_1, \lambda_2, \theta)] \\
&= \lambda_1 di_1 + \lambda_2 di_2 + T d\theta \\
&= \frac{\partial W_f'(i_1, i_2, \theta)}{\partial i_1} di_1 + \frac{\partial W_f'(i_1, i_2, \theta)}{\partial i_2} di_2 \\
&+ \frac{\partial W_f'(i_1, i_2, \theta)}{\partial \theta} d\theta
\end{aligned}$$

Therefore, comparing the corresponding differential terms, we obtain

$$\begin{aligned}
T &= -\frac{\partial W_f(\lambda_1, \lambda_2, \theta)}{\partial \theta} \\
T &= \frac{\partial W_f'(i_1, i_2, \theta)}{\partial \theta}
\end{aligned}$$

DC MACHINES

* Its common name given two types:-

- (1) DC gen^r (2) DC motor.

① DC gen^r → * A m/c which is designed to take the advantage of electromagnetic indⁿ, in order to convert mech. movement into electricity (dc vol.)

* Faraday's law of electromagnetic indⁿ → * Whenever a cond^r cuts magnetic flux a dynamicaly induced emf is produced in the cond^r. The induced emf is directly proportional to rate of change of flux linkage.

Flux → The amount of magnetic field around the magnet represented by lines of force. In generally it is indicated by ϕ & unit is wb.

Flux linkage → The extent of interaction between the flux & cond^r & cond^r & flux.

* It depends on the nature of flux time varying or time in varying.

* If the flux is time in varying in nature it require a relative motion between the flux & cond^r for the flux linkage.

* If the flux is time varying it automatically links with stationary cond^r.

* According to Faraday there are 3 modes of flux linkage

$$\lambda = N \times \phi$$

$$e \propto \frac{d\lambda}{dt} \propto \frac{d(N\phi)}{dt} = N \frac{d\phi}{dt} \text{ v}$$

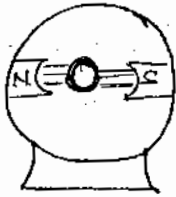
$$e = N \frac{d\phi}{dt} \times \frac{di}{dt} \text{ volts}$$

$$e = N \frac{d\phi}{dt} \times \frac{di}{dt}$$

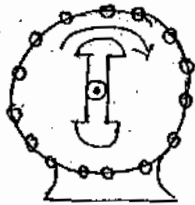
$$e = L \frac{di}{dt}$$

3 modes →

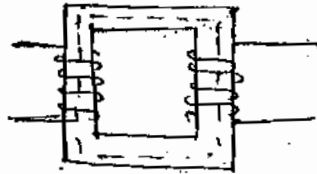
- (1) Conductors (Rotating) Flux (stationary)
 - (2) Conductors (stationary) Flux (rotating)
 - (3) Conductors (stationary) Flux (stationary)
- (Time in varying)



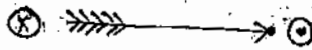
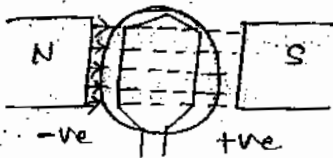
DC m/c mode (1)



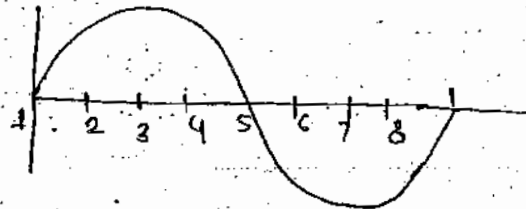
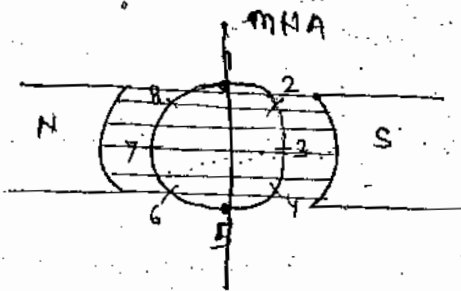
syn. m/c mode (2)



T/F mode (3)



1880 Thomas Alva Edison
1886 (AC) Nicola Tesla



$$e = N \frac{d\phi}{dt} \text{ volts}$$

$$e = B l v \sin \theta \text{ volts}$$

FRR
Fore - Dirn of flux
Thumb - Dirn of rot/motion
Middle - Dirn of emf/current

θ = Angle b/w cond^r rotation & flux line

- * Consider a simple coil rotating clockwise between 2 poles N & S. In one complete rotation it will produce 1 +ve half cycle & -ve half cycle which is periodic in nature known as alternating vol or current.
- * In positions 1 & 5 the cond^r movement is exactly parallel to the flux lines ($\theta = 0$)

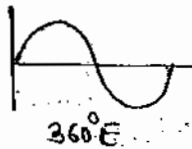
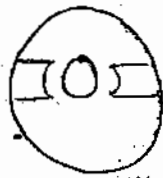
* Due to this which there is no rate of change of flux linkage & the induced emf is 0. The axis along 1 & 5 where the emf induced is 0 is called as Magnetic Neutral axis.

* At positions 3 & 7, $\theta = 90^\circ$ & the cond^r movement is exactly \perp to the flux lines.

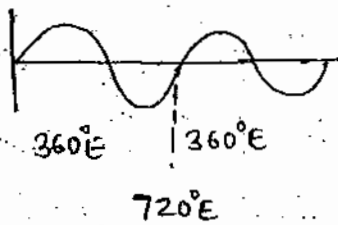
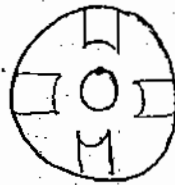
* Consequently max^m induced emf.

* MNA will be always 90° with the flux lines.

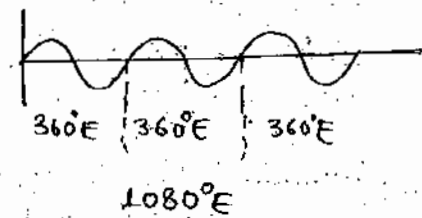
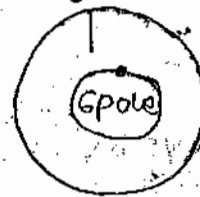
1 rot = 360° mech



1 Rot = 360° mech



1 rot: 360° m



* Under a pole there are always 180 ele. degree.

$$\theta_m = \theta_e / (P/2)$$

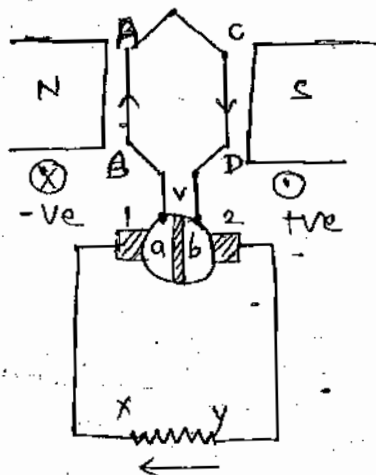
$$\theta_e = \theta_m \cdot P/2$$

* In AC or DC gen^r there will be alternating vol. induced in the cond^r (AC)

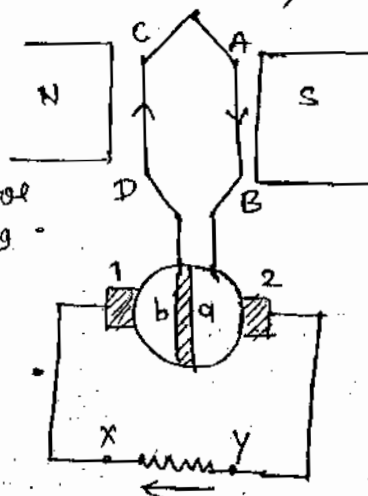
* In AC gen^r AC is directly taken but in a DC gen^r it will be converted into DC using a rotating commutator.

Action of Rotating Commutator →

1 → 5, 0 → 180°
BA CD b 2 Y X 1 a B

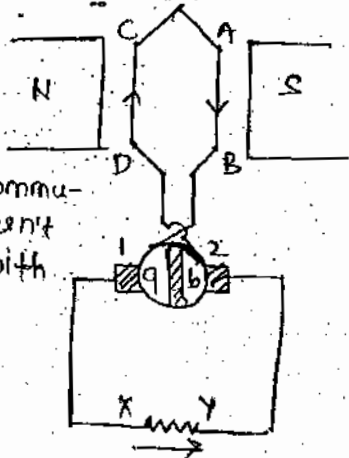


5 → 1, 180 → 360°
DCABQ 2 Y X 1 b D

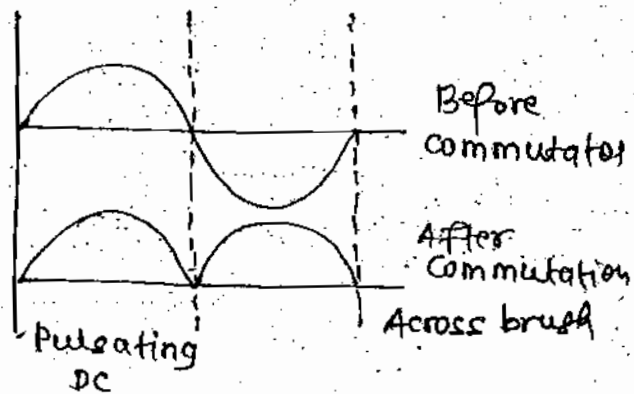


Commutator is rotating with coil

DCABQ 1 X Y 2 b D



when commutator doesn't rotate with coil



* By virtue of its rotation the commutator converts bidirectional induced emf into unidirectional (AC into DC) therefore it is called as mech. rectifier.

* In AC gen^r or DC gen^r the induced emf is AC.

* In DC gen^r the construction will be always rotating cond^r with the stationary field in order to make commutator action possible.

* When a coil rotates across the brushes there is pulsating DC which is not used for commercial DC operation.

In order to improve the shape of waveform there are many no. of coils

connected in series uniformly distributed known as q.m. wdg.

Constructional Details →

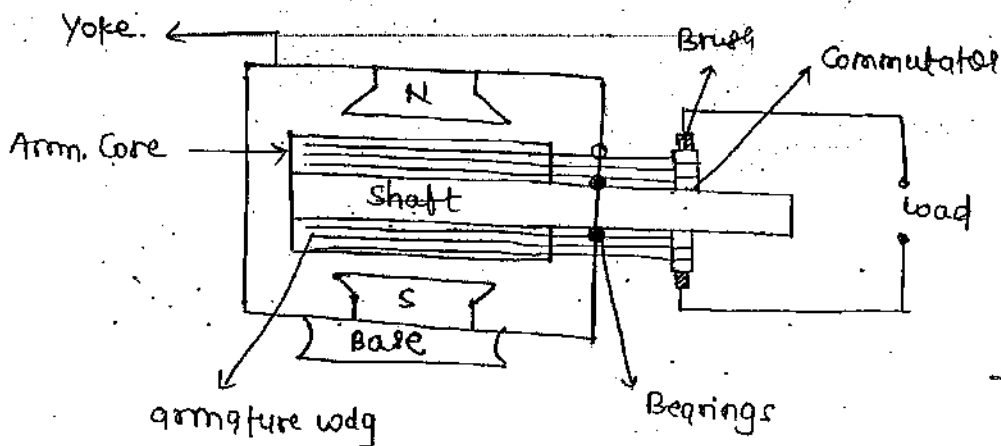
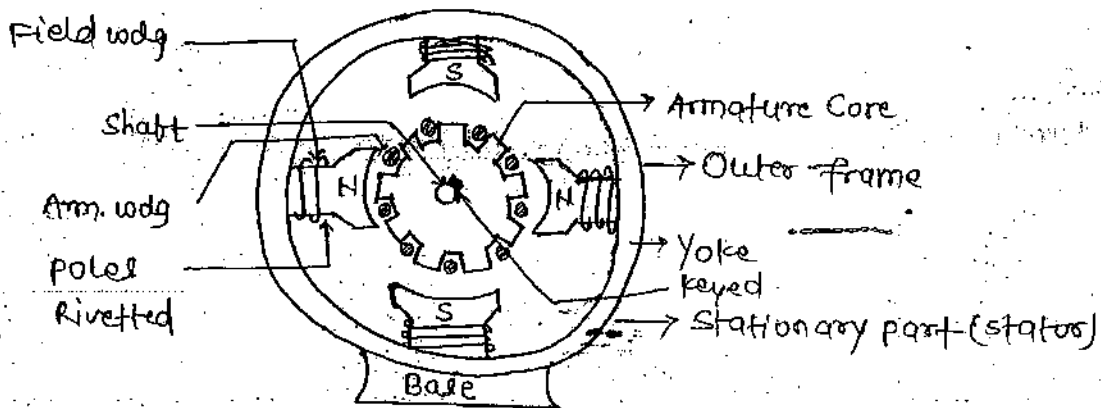
(1) Common features for all rotating electric →

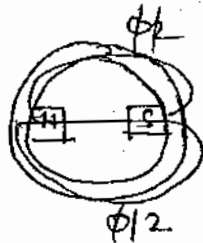
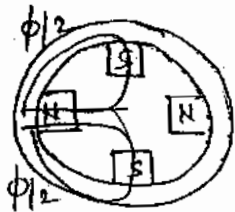
(i) The poles contains heteropolar st. (alternate North & South poles of even no.)

(ii) Excitation should be essentially in dc.

(iii) There will be stator (stationary part) & rotor (rotating part) with a least possible air gap between them.

Air gap → 0.5-2mm





Yoke → * It acts as protective covering to the entire m/c.

* It supports the poles as the poles are directly rivetted to it.

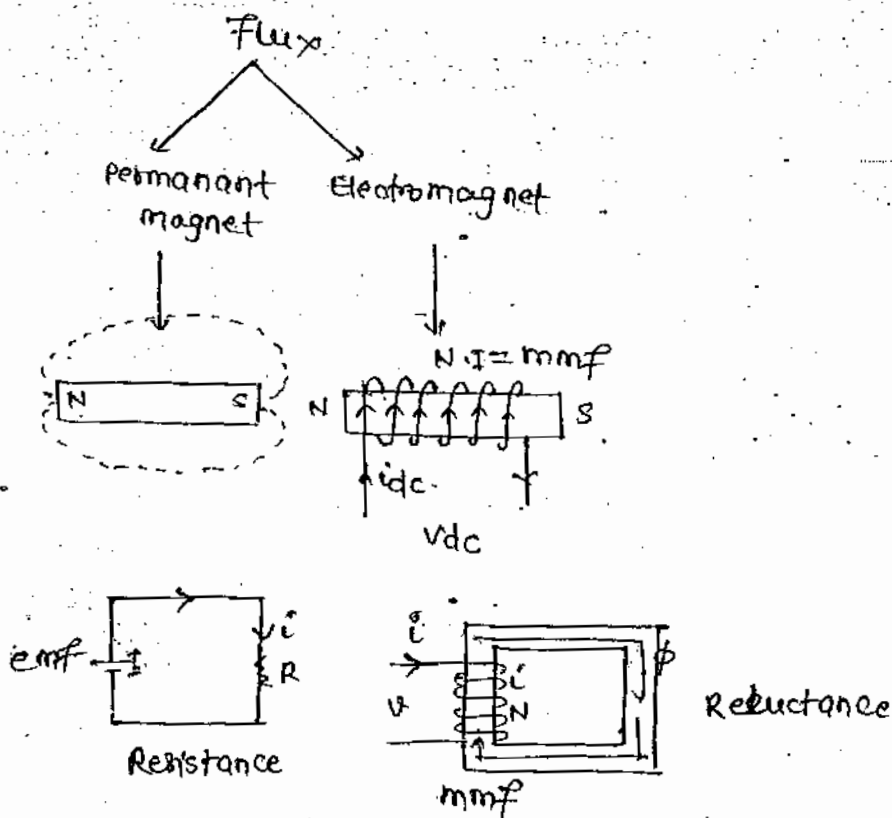
* It offers flux path completion of $\phi/2$ (if ϕ is flux/pole flux passing through yoke is $\phi/2$). Therefore yoke should be good magnetic material.

* For small m/c cast iron is used, large m/c - Fabricated ~~ste~~ steel.

* If the dc-m/c are operating across power electronic converter laminated yokes are preferred. (to reduce eddy current loss)

Pole → * It has to produce working flux in the m/c.

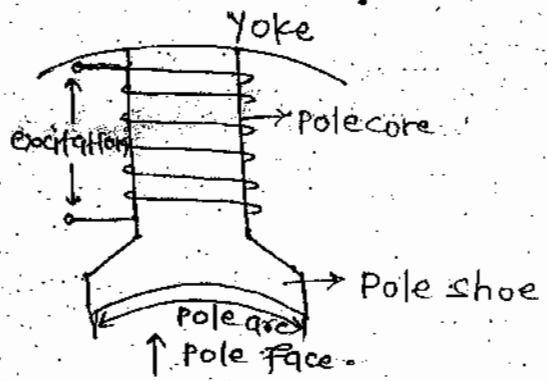
* The basic's



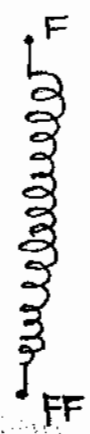
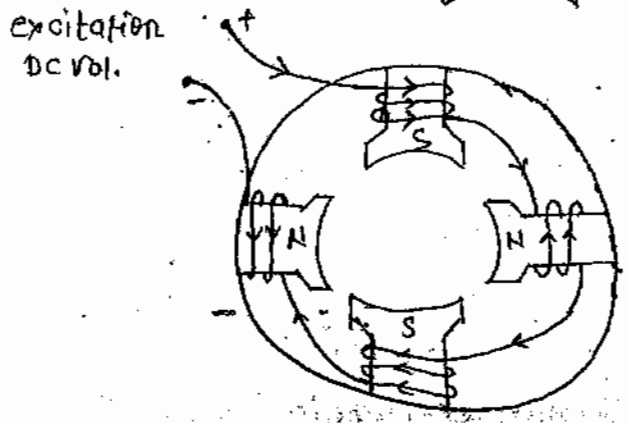
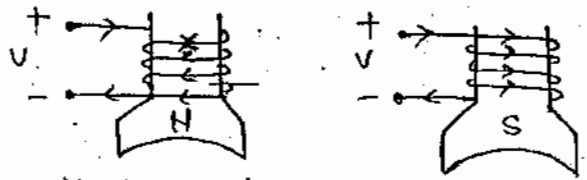
permeability → permit to flow the flux.

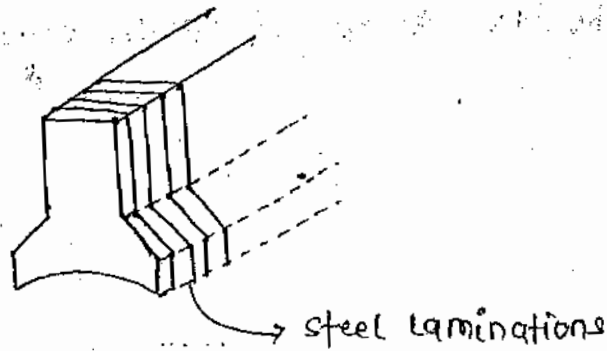
Excitation should be always dc for the field wdg.

- * The funⁿ of the pole is to produce the working flux in the m/c.
- * The basic source of flux is permanent magnet which is uncontrolable.
- * In order to control the m/c flux need to be controllable. Therefore electromagnets are preferred which require wdg & a dc vol. across it called as excitation.
- * Excitation is essentially dc because it produce fixed poles polarity.
- * The pole is spread out as pole shoe.
- * In order to reduce the reluctance in the air gap & to spread flux uniformly on the arm. condⁿ



* The polarity of a pole depends on the polarity of excitation & orientation or sense of wdg.





* Poles are also made up of steel laminations.

* In order to reduce eddy current loss when the flux is not ideally dc.

3) Armature Core → * Arm. core is a cylindrical drum like st punched into slot on the peripheral.

* The arm. wdg is placed in this slot with the suitable insulation.

* It is mounted & keyed to the shaft. It should be superior magnetic material.

* As all the electromech. energy conversion happens in this rotating part of m/c (motor or gen), generat

* Generally si is used as it has superior magnetic property.

* Due to its high conductivity it will also produce eddy currents.

* Therefore solid cores are not preferred but cores are laminated with thin laminations 0.4-1mm thickness.

* Each lamination act as individual core to form single core & the eddy current in each lamination will be considerably reduced.

si + steel

↓
- low hysteresis coefficient -

Reduce the conductivity of steel

$\alpha = 1.5 - 2.5$

Reduce eddy current & hysteresis loss

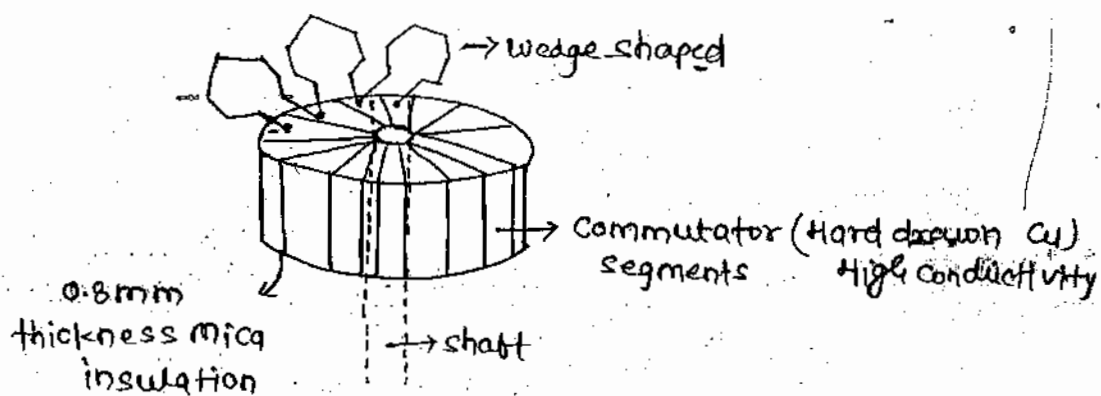
Also called electrostatic technic

* 3.5-4% Si is added to steel which reduce the losses occurring at core known as iron loss.

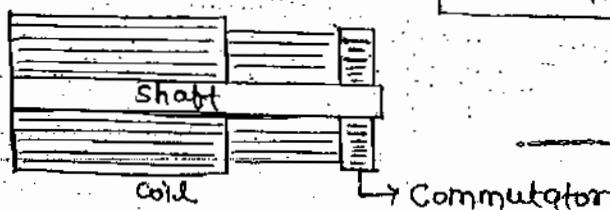
* Adding Si reduce the conductivity of steel without destroying its magnetic property as well as it has low hysteresis coefficient as 1.6 to reduce its hysteresis loss.

* If more Si is added it will destroy mech. property of steel.

* Commutator →



Commutator segments
= No. of coils



* Commutator is split ring which is segmented through 0.8mm thickness mica insulation.

* The segments are made up of hard drawn Cu.

* It converts by directional emf or current inside the coil into unidirectional.

* It is the ~~qr~~ image of arm. wdg inside.

* The no of commutator segments are equal to no. of coil.

* Commutation plays the vital role in the operⁿ of dc m/c

Brushes → * Brushes offer ele. connection b/w rotating commutator & stationary load.

* They collect current from the wdg placed on the commutator through brush holders & spring.

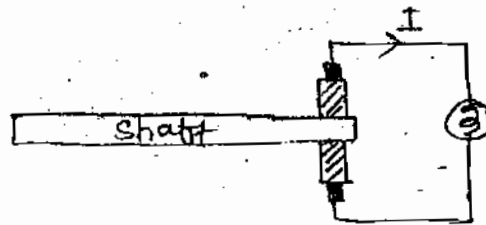
* These are stationary sliding contacts.

* If the brushes collect current without any sparking then the commutation is successful.

* If there exist any spark while collecting current then the commutation is not successful.

* Due to high peripheral speed any spark will spread into two or 3 segments & so the coil inside it & produce large current in them.

* In order to insure successful commutation mech. as well as ele. condⁿ should be proper.

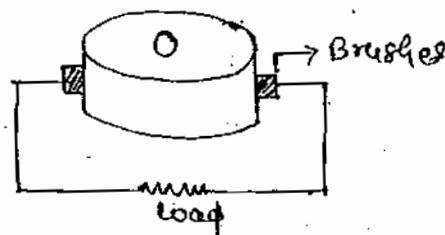


* In order to insure good mech. condⁿ the brush is placed in a brush holder & placed on the commutator through spring.

* In order to insure good ele. condⁿ & successful commutation the brushes should be always placed on MNA (Neutral zone).

* The brush materials used are Cu, C & electrographite.

* C brushes are used generally to improve commutation. (refer commutation topic).



small dc - c
All dc m/c - Electro-graphite
LVHC dc - Cu graphite

Shaft & bearing → * The purpose of a shaft is to provide mechs. o/p.

* When the m/c operate as a gen^r & to collect mechs o/p then the same m/c operate as a motor.

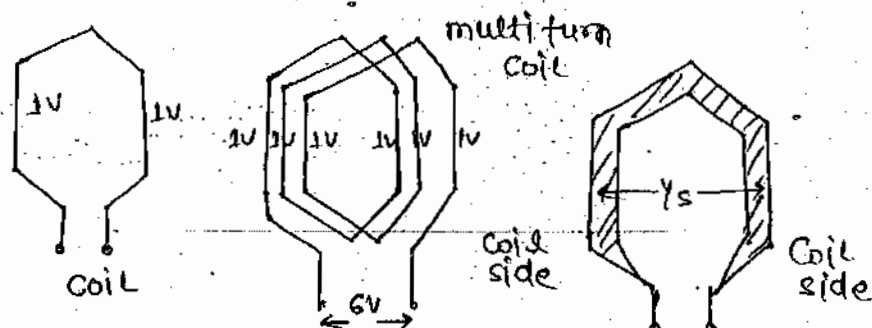
* It is hold through bearings which offer rotation.

For large m/c → Rolar bearing
Small → Ball bearing

Armature wdq → * Cond^r may be made into turns & multiturns to form coils which are connected in series & distributed uniformly throughout the entire peripheral of the arm.

Cond^r → (z) The length of the wire lying in magnetic field where emf is induced.

Turn → Two cond^r made 1 turn; if there are z cond^r there will be z/2 turns.



* If a coil consist of 1 turn it is known as 1 turn coil.

* If there are two or more turns then it will be multiturn coil.

* Practically multiturn coil are used, as they provide more voltage.

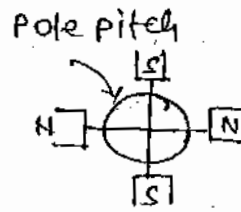
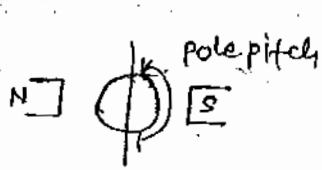
* A multiturn coil consist of two coil sides which are placed in the slots with a coil span.

Coil span → (y_s) Distance b/w 2 coil sides of a coil

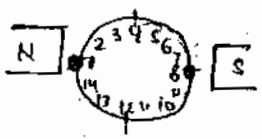
Pole pitch → The peripheral distance b/w 2 adjacent poles expressed in no. of slots or cond^r.

(y_p)

slots/pole or z/p

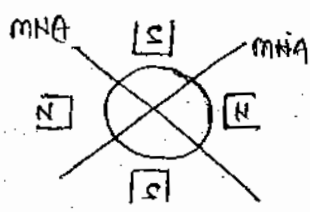


1 pole pitch = 180° ele degree



* For obtaining max^m vol. the coil side must kept at point 1 & 2nd will be at 8.

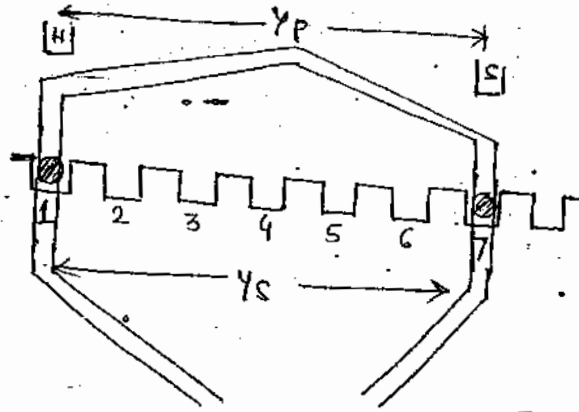
slots/pole = $\frac{14}{2} = 7$ (hence put a 8)
(maintain 7 slots gap)



Between the two adjacent pole there is one MNA

- * If the coil span is exactly equal to pole pitch then it is known as full pitch coil & wdg known as full pitch wdg.
- * Do m/c arm. wdg are full pitch wdg only
- * In order to get max^m induced emf

Coil span < pole pitch

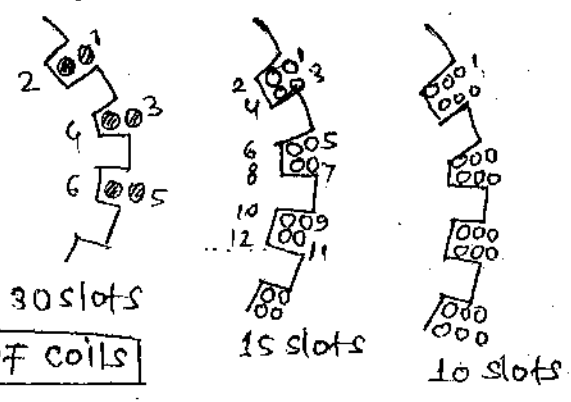


Single layer wdg → One coil side placed in one slot.



60 coil side (small m/c)
30 coil
60 slots

Double layer → Two coil side placed on one slot.
 or
 Two layer large m/c

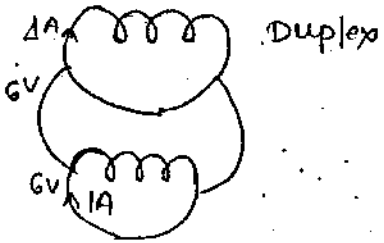
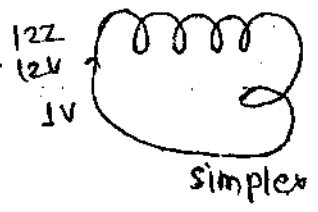


No. of slots = No. of coils

Multiplex wdg →

multiplicity 'm'

- * If dc wdg are closed wdg if there is one completely closed set of wdg it is known as simplex wdg with multiplicity factor $m=1$.
- * If there are 2 such completely closed sets of wdg connected in parallel it is known as duplex, $m=2$
- * Similarly 3 sets triplex, $m=3$
- * 4 sets = Quadruplex, $m=4$



- * Multiplex wdg increase the current rating or loading capability of m/c.
- * This are more advantageous in wave wdg than lap wdg.
- * For a given no. of cond^r multiplicity is increase the current rating while decreasing its vol. rating.

Back pitch \rightarrow The no. of cond^r spanned by one coil at the back end of arm.
(γ_b)

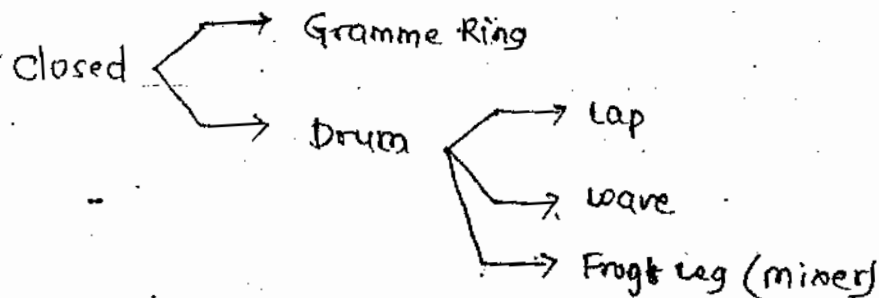
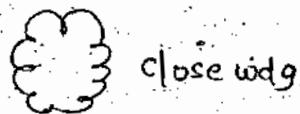
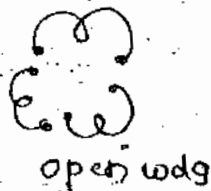
Front pitch \rightarrow The no. of cond^r spanned by one coil at the front end of arm. (front end is commutator end)
(γ_f)

Resultant pitch \rightarrow The beginning of one coil & its next successive coil (distance between them).
(γ_R)

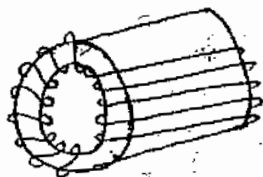
Commutator pitch \rightarrow NO. of commutator segments connected to 2 successive coil.
(γ_c)

Types of arm wdg \rightarrow The basic classification is done depending on the closure.

Closure: $\left\{ \begin{array}{l} \text{Open (ac gen^r) / AC @ arm. wdg} \\ \text{close (dc gen^r)} \end{array} \right.$



* DC gen^r wdg should be closed type as there is a commutator
Gramme Ring \rightarrow



* This is the 1st form of arm. wdg which has been totally replaced with drum type due to the following disad:-

(1) Half of the turn is wasted which is lying inside the arm.

Core.

(2) Insulating maintenance, repairs & design is costly.

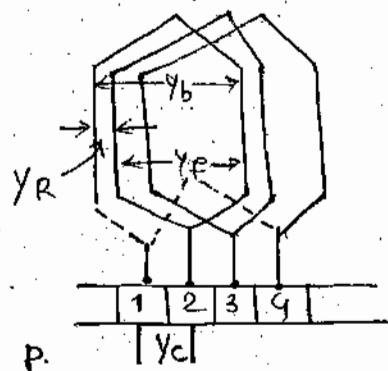
Drum →



* The arm. core is a cylindrical drum which is slotted on the periphery where the wdg is placed in that.

* No turns are wasted, designed maintenance & repairs is comparatively easy.

Lap wdg →



$$Y_R = Y_b - Y_f$$

- (1) Calc Y_b, Y_f
- (2) Develop wdg dia table
- (3) Wdg dia
- (4) Polarity mark brush positions



$$Y_s = Y_p \approx Y_b \approx Y_f$$

$$Y_A = \frac{Y_b + Y_f}{2} = \frac{Z}{P} \text{ avg. value}$$

* In order to support lap wdg format Y_f, Y_b should be opposite signs.

* Therefore $Y_b \neq Y_f$

* Both Y_b, Y_f should be odd no. in order to have symmetric double layer wdg.

* Therefore

$$Y_b - Y_f = \pm 2m$$

$$Y_c = \pm m$$

$$Y_b > Y_f \rightarrow +2m$$

* It also called RHS / progressive wdg.

$$Y_b < Y_f \rightarrow -2m$$

* So it called as Retrogressive / LHw.

Eg. \rightarrow Design simplex progressive lap wdg for 24 Cond^r 4 Poles,

solⁿ \rightarrow

$$\frac{Y_b + Y_f}{2} = \frac{Z}{P} = \frac{24}{4} = 6$$

$$Y_b + Y_f = 12,$$

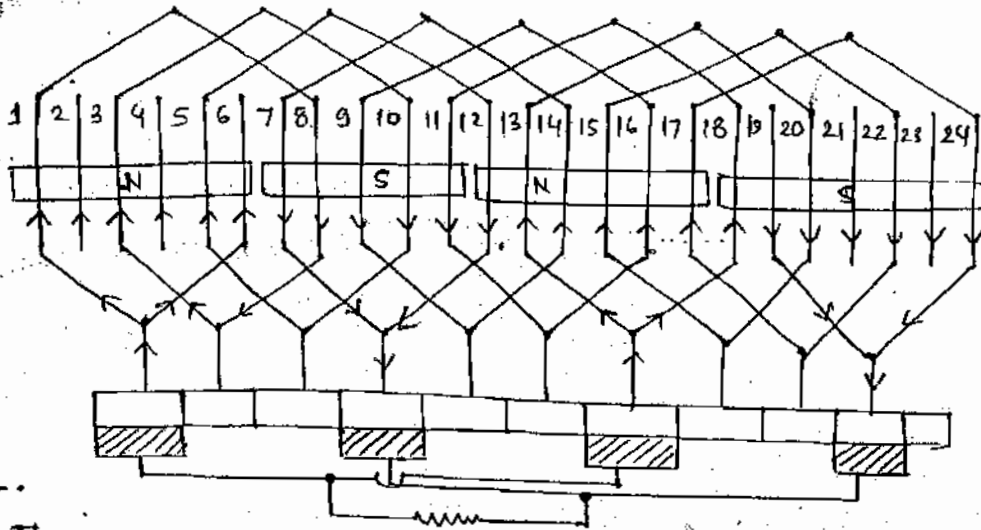
$$Y_b - Y_f = 2$$

$$Y_b = 7, Y_f = 5$$

wdg table \rightarrow

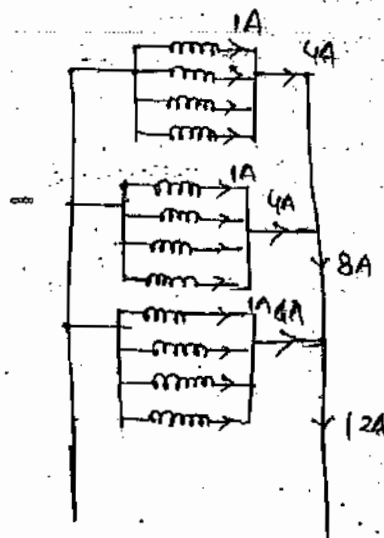
	$Y_b = 7$	$Y_f = 5$
	$1+7=8$	$8-5=3$
	$3+7=10$	$10-8=5$
	$5+7=12$	$12-5=7$
	$7+7=14$	$14-5=9$
	$9+7=16$	$16-5=11$
	$11+7=18$	$18-5=13$
	$13+7=20$	$20-5=15$
	$15+7=22$	$22-5=17$

wdg. diagram →



- * The no. of parallel paths are always equal to the no. of poles.
- * Multiplicity also increases the no. of parallel path.

Therefore $A = p m$



$$A = p m$$

$$= 4$$

$$A = p m$$

$$= 4 \cdot 2$$

$$= 8$$

$$A = p m$$

$$= 4 \cdot 3$$

$$= 12$$

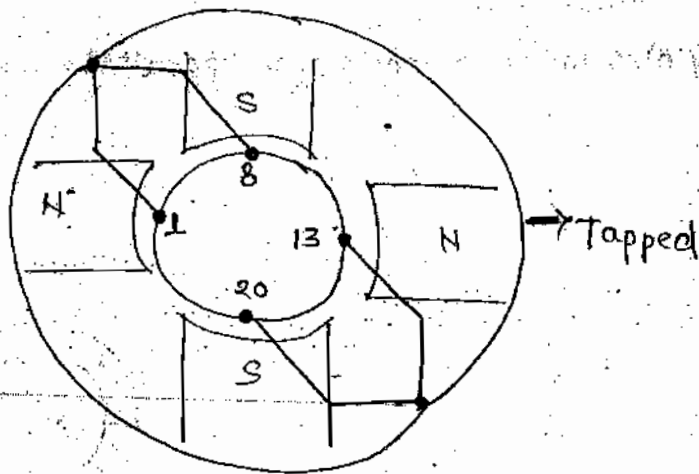
- * It is also known as parallel wdg as there are more parallel path in it.
 - * Consequently no. of cond^r in series per parallel path is less.
- Therefore it is employed ^{for} high currents low vol. ratings.

Equalizer rings → These are thick cu cond^s located at the back end of arm, which has low resistance.

* Equipotential coils are tapped individually to respective equalizer rings.

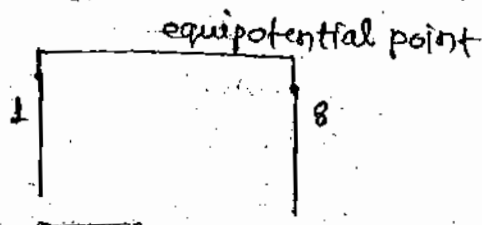
* Any circulating current will be bypassed & circulates within the equalizer rings at the back end & doesn't enter into the commutator at front end.

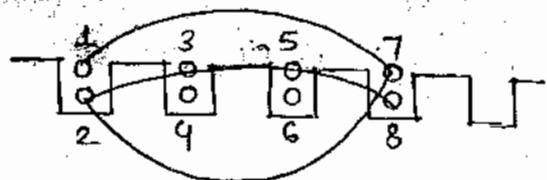
Equipotential coil → * These are exactly located twice the pole pitch distance or $360^\circ E$ apart.



* Any unbalance in the induced emfs in the parallel paths will create a circulating current which interfere with commutation when it flows through the bus.

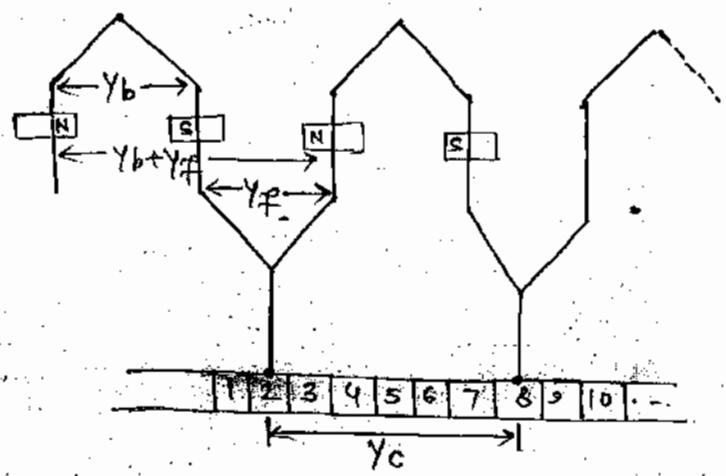
* Therefore any circulating current will be bypassed through the equalizer ring at the back end & doesn't enter into the commutator at front end.





→ Unsymmetric wdg.

Wave winding →



- * The tech. Cons. diff. be/w lap & wave is in the commutator pitch.
- * In wave it is twice the pole pitch.

$$y_A = \frac{y_f + y_b}{2} = \frac{z \pm 2}{p}$$

* y_A should be integer

* y_b, y_f have same sign to support wave format

$$y_b = y_f$$

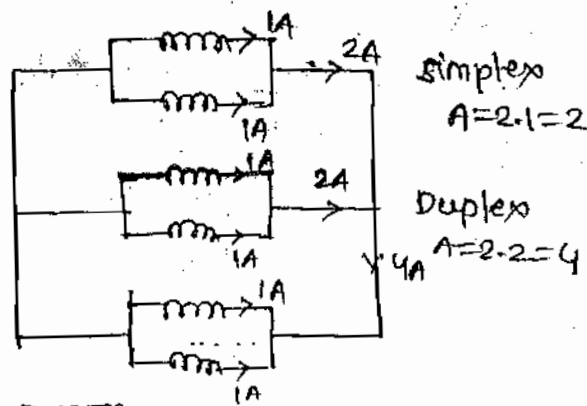
(OR)

$$y_b - y_f = \pm 2m$$

$$A = 2m$$

* The no. of parallel paths are independant of no. of poles & always = 2 as the multiple wdg increase the no. of parallel path.

$$A = 2m$$



Poles	Lap	wave
2	2	2
4	4	2
8	8	2
...
10	10	2
...
20	20	2

* Multiplex wdg's are advantageous in wave than lap.

* As there are less no. of parallel path it is employed for high voltage low current ratings.

* Dummy Coil →

Q → Design wave wdg's for 60 cond^r 15 slots 4 pole simplex.

Solⁿ →

$$Y_A = \frac{Z \pm 2}{p} = \frac{60 \pm 2}{4} \quad 2 \rightarrow \text{missed}$$

$$Y_A = \frac{62}{4} \text{ or } \frac{58}{4} \quad \frac{60}{4} \text{ or } \frac{56}{4}$$

$$14 \times 4 = 56$$

But in 15th slot = ?

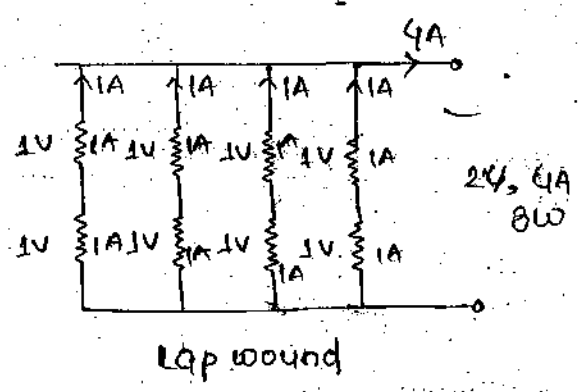
* For some set of data Y_A will not be integer in order to make it integer value the nearest possible cond^r will be considered.

* due to which some cond^r are missing in any one of the slots

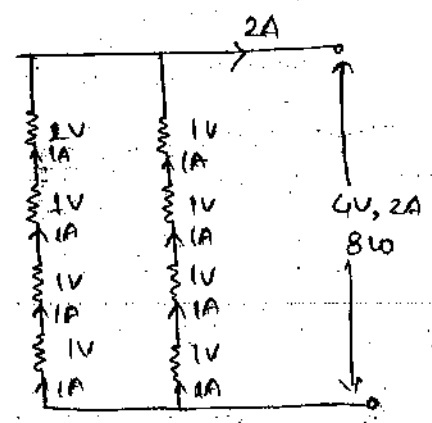
In order to maintain mech. balance the missing cond^r are well insulated & placed in the missing slot as dummy.

- * It is not connected to the rest of the lodg.
- * As there are only 2 parallel path any unbalance will become a balanced condⁿ.
- * Therefore no need of equalizer rings.

eg → Consider a 4p simplex lap wound gen^r arm, with 8 cond^r is reconnected, as wave.



A_L, V_L, I_L, P_L, R_L



A_W, V_W, I_W, P_W, R_W

$$I \propto A \rightarrow \frac{I_L}{I_W} = \frac{A_L}{A_W}$$

$$V \propto \frac{1}{A} \rightarrow \frac{V_L}{V_W} = \frac{A_W}{A_L}$$

$$P_L = P_W \quad I_L^2 R_L = I_W^2 R_W$$

eg → A 4p dc gen^r with lap wound arm, is reconnected as a wave what will be the change in V, I & P?

solⁿ →

$$\frac{I_L}{I_W} = \frac{A_L}{A_W} = \frac{4}{2} \quad \therefore \frac{I_L}{I_W} = 2 \quad \boxed{\frac{I_L}{2} = I_W}$$

$$e = \frac{P\phi}{60/N} = \frac{\phi PN}{60} \text{ voltz}$$

emf induced per parallel path $\frac{\phi PN}{60} \times \frac{Z}{A}$

$$e = \frac{\phi PN}{60} \times \frac{Z}{A}$$

$$e = \frac{\phi ZNP}{60A}$$

$$E_g = \frac{\phi ZNP}{60A} \text{ volts}$$

Gen/induced emf

$$E_g \propto \phi N$$

$$\frac{E_{g1}}{E_{g2}} = \frac{\phi_1 N_1}{\phi_2 N_2}$$

(OR)

$$\frac{E_{g2}}{E_{g1}} = \frac{\phi_2 N_2}{\phi_1 N_1}$$

If N is constant

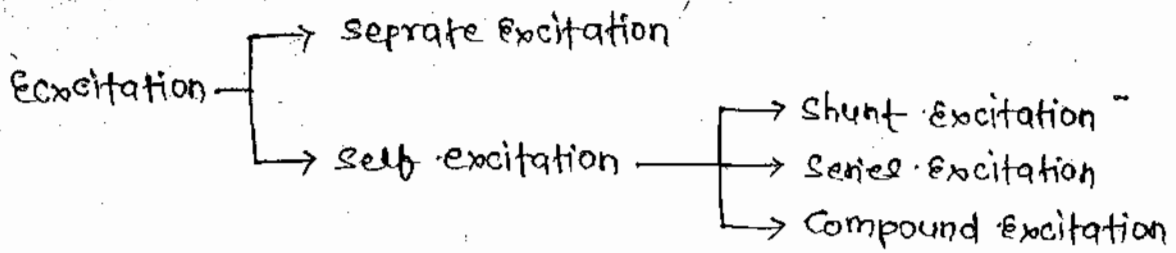
$$\frac{E_{g2}}{E_{g1}} = \frac{\phi_2}{\phi_1}$$

If ϕ is constant

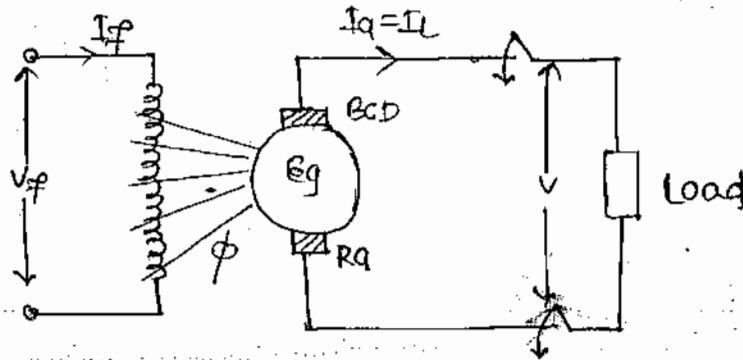
$$\frac{E_{g2}}{E_{g1}} = \frac{N_2}{N_1}$$

Classification of dc gen^r → * In order to control the m/c electro-magnets are preferred as poles which requires a dc vol. excitation across it

The classification of gen^r is according to method of excitation:-



* Seprate Excitation →



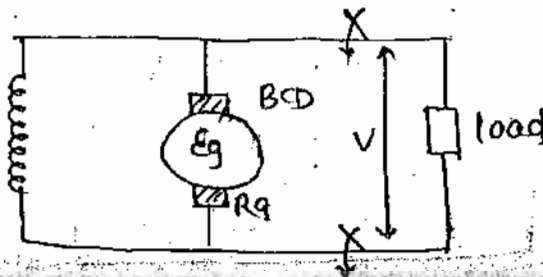
- * It requires a seprate dc source for its excitation.
- * Its field & arm. wdg are isolated electricaly.
- * The terminal voltage (V) across the load doesn't affect its excitation.
- * Due to additional voltage source requirement it is rarely used gen^r around 5% do gen^r are self excited only.

gen/Ind emf

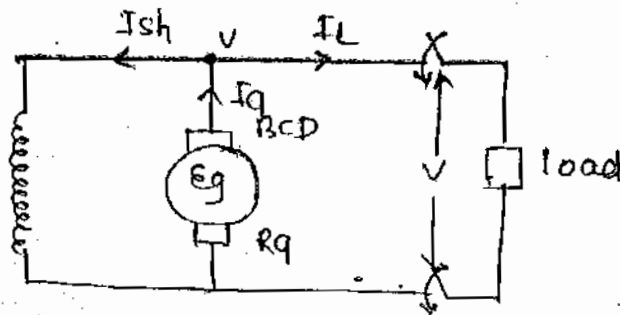
$$E_g = I_a R_a + BCD = V \text{ Terminal } V$$

$$I_a = I_L$$

* self excitation → * The field wdg will be excited by its own arm. which requires some essential condⁿ starts with residual flux.



shunt excitation →



$$E_g = I_a R_a - BCD = V$$

$$I_a = I_L + I_{sh}$$

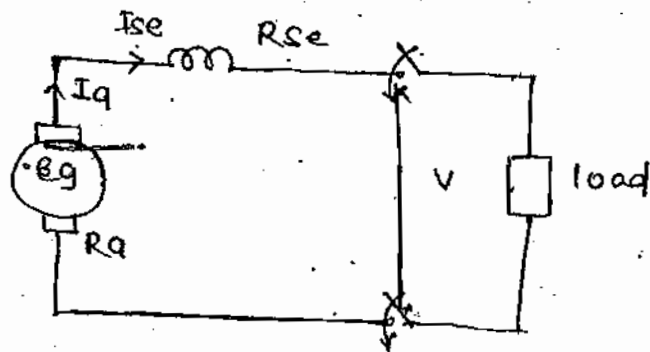
$$I_{sh} = \frac{V}{R_{sh}}$$

- * The field wdg is connected in parallel with arm. through brushes.
 - * The terminal vol. itself act as excitation.
 - * Shunt field turns are very large in no. due to its resistance is high with thin cond^r.
- Range 50-250Ω

Loading any m/c = Reducing the resistance

- * The field turn current remains approx. same from NL to rated value.
- * Consequently flux is approx. same in the operating region.
- * In this mode it is known as shunt gen^r & vol. operated field.

Series excitation →

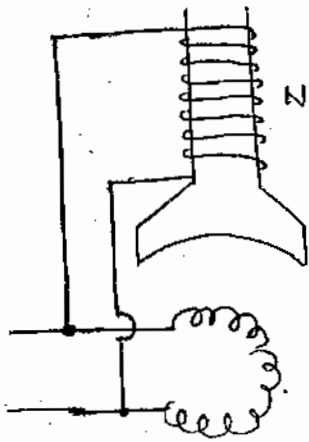


- * In this mode it is known as series gen^r & current operated field.

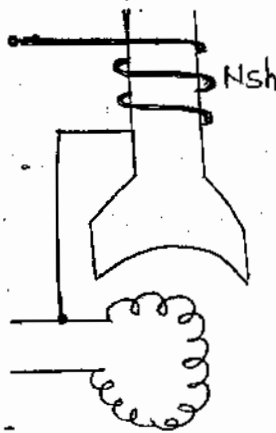
$$E_g = I_a (R_a + R_{se}) - BCD = V$$

$$I_a = I_L = I_{sh}$$

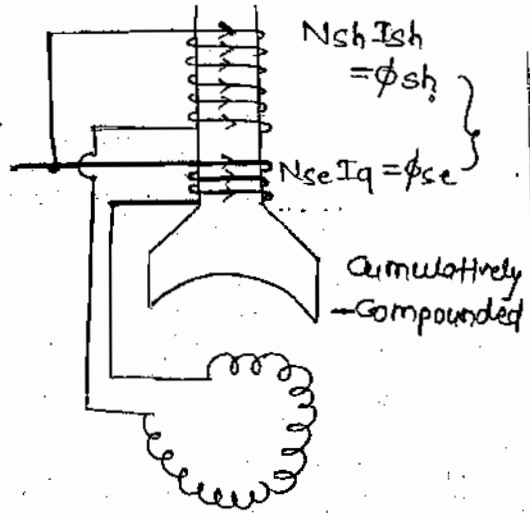
* The series field wdg contain less no. of turns with thick cond^r



Shunt



Series



Compound

$$N_{sh} I_{sh} = \phi_{sh}$$

$$N_{se} I_a = \phi_{se}$$

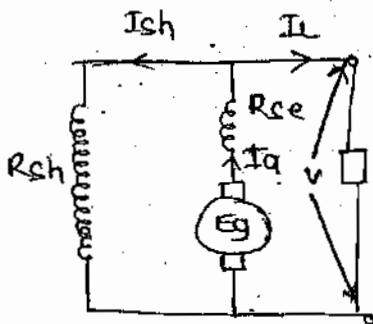
Cumulatively compounded

* Shunt flux always dominate series flux

* If series field wdg is reversed than $\phi_{sh} - \phi_{se}$: Differently

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Long shunt →

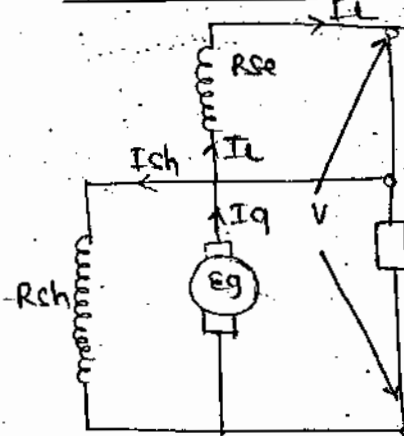


$$E_g = I_a(R_a + R_{se}) - BCD = V$$

$$I_a = I_L + I_{sh}$$

$$I_{sh} = V / R_{sh}$$

Short shunt →

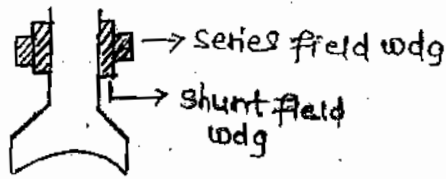


$$E_g - I_a R_a - I_L R_{se} - BCD = V$$

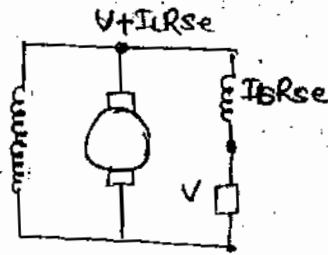
$$I_a = I_{sh} + I_L$$

$$I_{sh} = \frac{V + I_L R_{se}}{R_{sh}}$$

* There is no



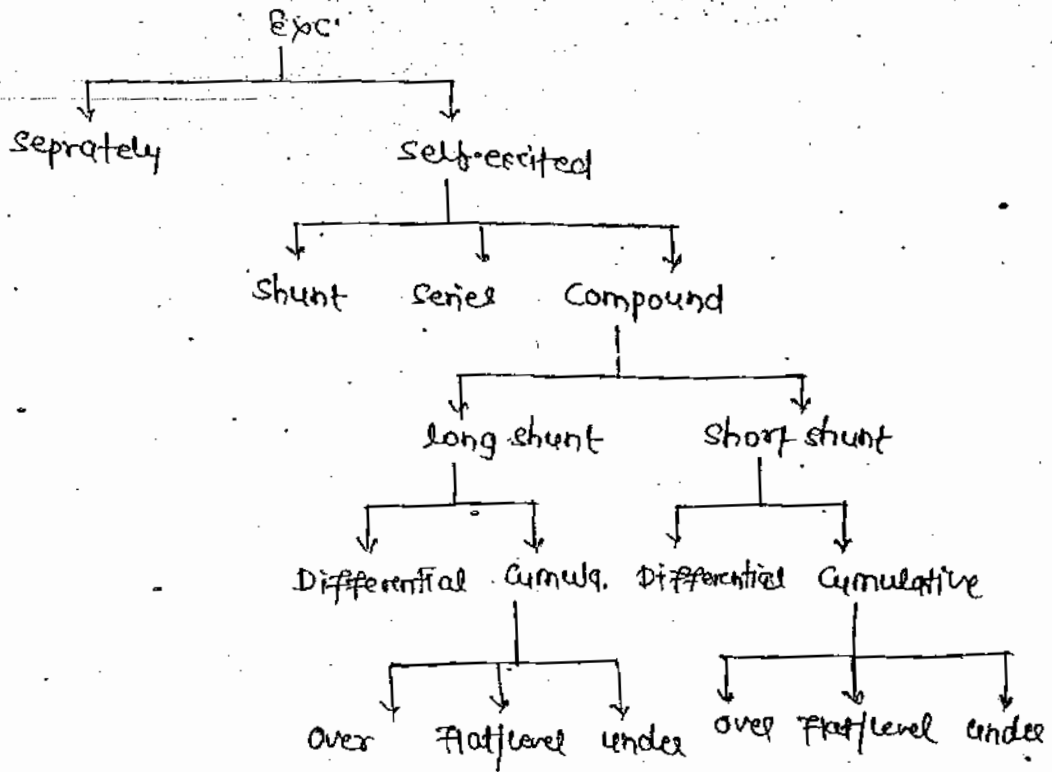
Short shunt



- * There is no distinctive diff. b/w long shunt & short shunt gen^r.
- * The induced emf in long shunt is slightly greater than that of short shunt ($I_a > I_f$)
- * If the series field wdg. is reversed than the cumulative mode becomes differential.

Cumulative currents in both field wdg same dir on the pole core. Therefore both fluxes add each other, and the net flux increases with load.

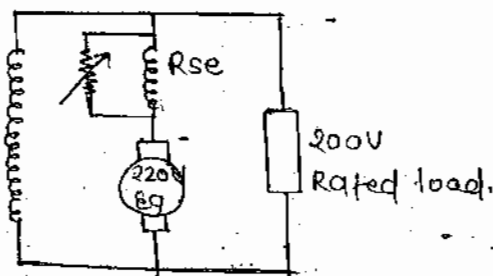
In differential the series flux oppose the shunt flux & the net flux decreased with load.



Compounding → * Adjusting the terminal voltage of cumulative compound gen^r by varying its series field ampere turns by connecting a diverter across its series field wdg.

* Depending on the degree of compounding a cumulative compound gen^r can act as under, flat, over compound gen^r.

Compounding is done at rated load. to adjust the terminal vol. below rated, exactly rated & above rated respectively.



* By varying the diverter resistance from 0 to max^m value a cumulative compound gen^r can be acted in 4 other modes.

- (i) $R_d = 0$ (shunt)
- (ii) $R_d \uparrow$ Basic cumulative
- (iii) $R_d \uparrow \uparrow$ Under
- (iv) $R_d \uparrow \uparrow \uparrow$ Flat/level
- (v) $R_d \uparrow \uparrow \uparrow \uparrow$ Over

* Compounding can't be done with differential compound gen^r as its series flux oppose shunt flux.

* Induced emf of flat compound gen^r is exactly equal to rated voltage at rated load.

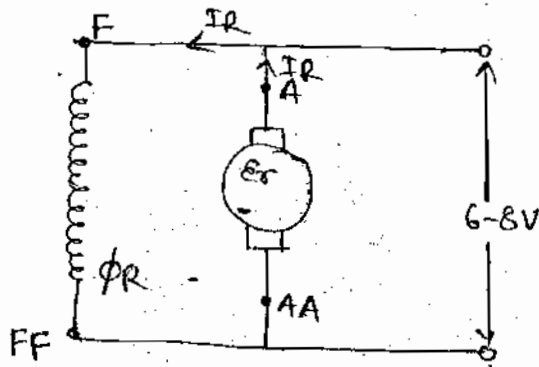
Since series flux is adjusted to compensate all the drops at rated load.

Voltage build up → [In self excited gen^r]

* In a separately excited gen^r there is a separate source to excite the field & produce flux.

* In a self excited gen^r there is no such source to excite the field wdg. Therefore in order to build up voltage in self excited gen^r the poles should contain the residual flux.

Shunt gen^r →

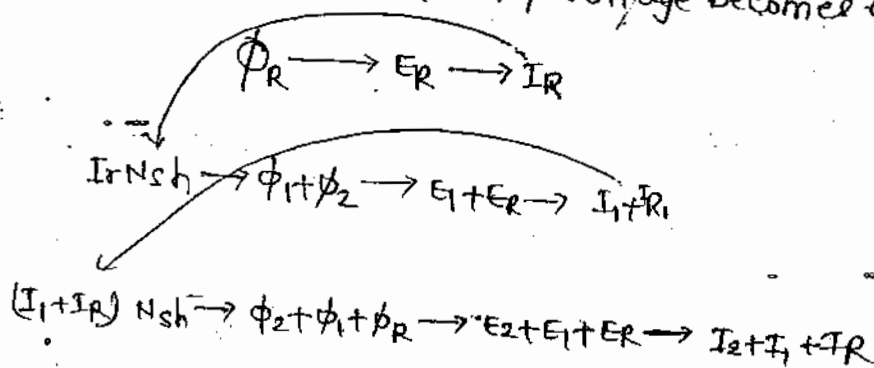


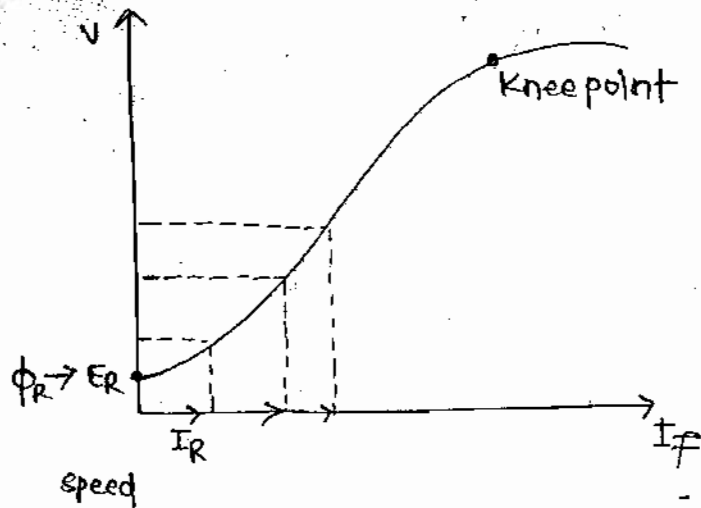
1) * Consider a shunt gen^r made to run at rated speed with residual flux in its poles small voltage is induced in the arm. (6-8V) which produces small current.

* If the gen^r is on NL, all the initial current will flow into field wdg. to produce initial mmf with field wdg turns.

2) * The field should be properly connected to the arm. in order to make initial mmf at the residual flux.

* It is a cumulative process as the current increases the flux out of the pole increased & the induced voltage also increases after saturation of poles even though the current increases. Flux doesn't increase. Consequently voltage becomes constant.



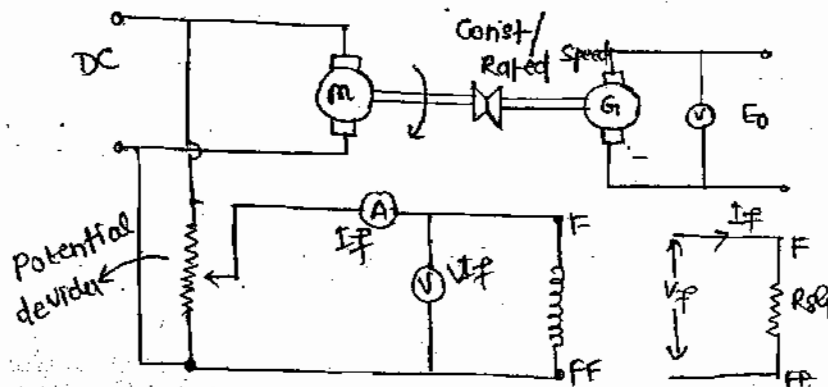


- (3) The operating ~~condⁿ~~ is of the gen^r should be greater than critical speed.
- (4) The resistance of field wdg should be less than critical resistance value.
- (5) The best condⁿ for a shunt gen^r to build up voltage successfully is it should be on NL.
- (6) In spite of all the above 5 condⁿ satisfied if the vol. doesn't build up it may be due to improper contact across brushes & commutator.

critical
Determination of critical resistance & speed →

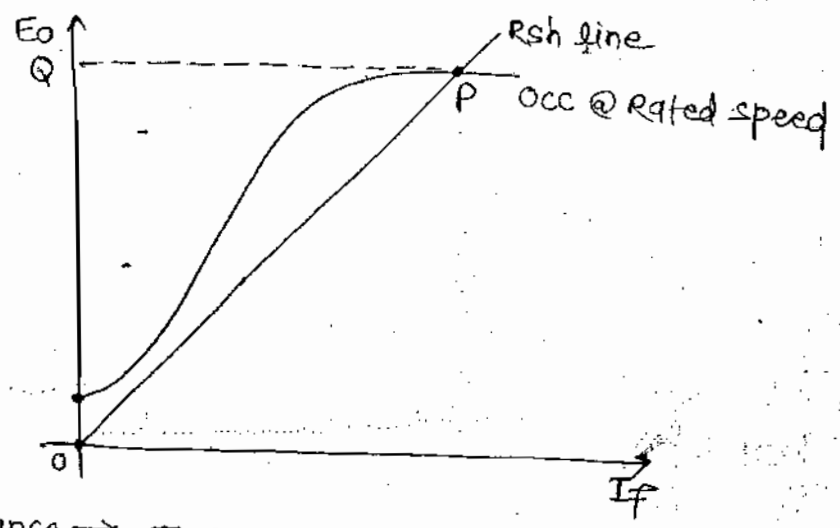
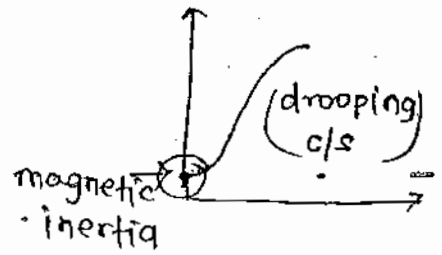
OCC (open circuit c/s)
 NLS (No load saturation)
 Magnetisation c/s } E_o vs I_f

* In order to plot OCC to determine R_c & N_c of a self exc. shunt gen^r it requires to be separately excited.



S.N.	V_f	I_f	E_o
1	0	0	6.8
2	10	0.1	30
3	20	0.2	60
4	30	0.3	80

Linear

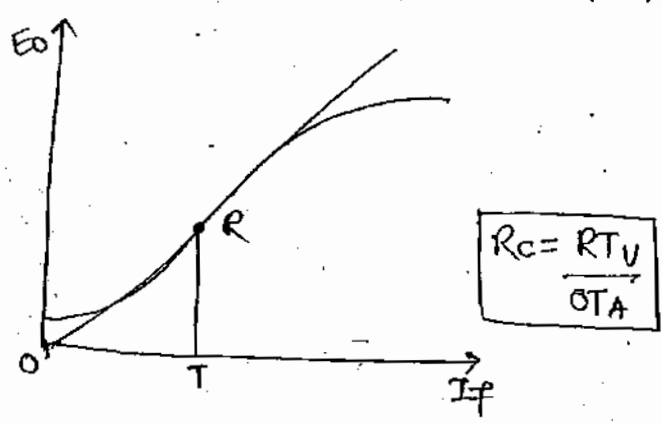


Critical Resistance → The resistance of the field wdg. about which the gen^r doesn't build up voltage ~~this is~~

Critical speed → The speed of gen^r below which the gen^r doesn't build up voltage. s

Determination of critical resistance →

- * The field resistance line intercepts OCC at a point P.
- * Length of OQ in volts is the max^m emf induced in the gen^r.



Steps: (1) Draw the OCC line.

Sagar Sen

887145356

(2) Plot a tangent through OCC.

Critical speed Determination →

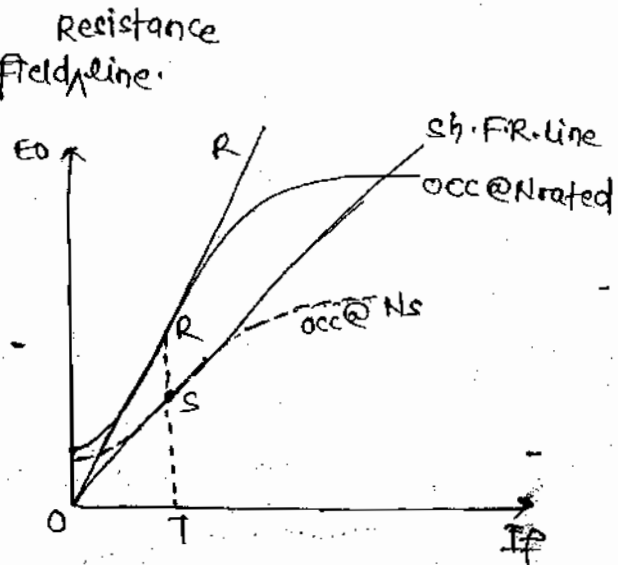
Steps → (a) Draw original shunt field line.

length of (ST) $\propto N_c$

length of (RT) $\propto N_{rated}$

$$\frac{ST}{RT} = \frac{N_c}{N}$$

$$N_c = \frac{ST}{RT} N$$



* The critical resistance of a given gen^r depends on the operating speed. From critical speed & above.

* It varies proportionally with speed.

* When the m/c is running at critical speed its field resistance value itself is critical resistance.

Eg. → A shunt gen^r building up vol. normally. If the field wdg is reversed & operated that

(a) Build up vol. normally

(b) Build up vol. with -ve polarity

(c) No build up of vol.

(d) GV across the arm.

Ans. → (c)

Eg. → Same above que. Direcⁿ of rotation of gen^r is reversed. options same.

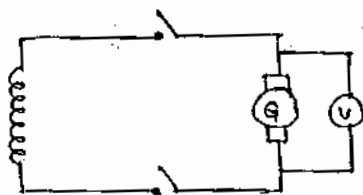
Ans. → (c) $-E_g \propto \phi(-N)$

eg. → If both direcⁿ of rotation as well as field wdg. is reversed,

$$E_g \propto (\phi)(N) \text{ is } E_g (+ve)$$

ans. (a)

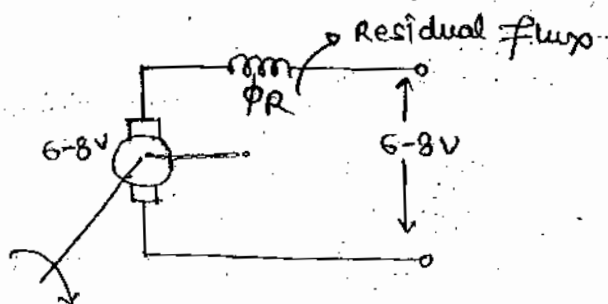
2/1



switch

	Open	Close	
A.	8V	8V	1. Critical Resistance
B.	8V	12V	2. load resistance
C.	0V	0V	3. Field polarity

* Voltage build up in series gen^r →



- (1.) It requires residual flux.
- (2.) The terminals should be closed with some load. As the load current flows into the field wdg the initial mmf is considerably good value & no need of large series field wdg terms.
- (3.) The total resistance ($R_a + R_{se} + R_L$) should be less than its critical resistance.

$(R_a + R_{se} + R_L) < R_c$
- (4.) Its speed should be greater than critical speed.

Cumulative Compound gen^r →

* A compound gen^r is eq. to a shunt gen^r on NL. Therefore its vol. build up is eq. to shunt gen^r.

Armature Reaction →

- * On NL the arm. current is negligible. Consequently there is only main flux in the air gap distributed uniformly in flat top^{ped} nature.
- * As the arm. is loaded load current flows in the arm. conductors which produce arm. mmf & arm. flux which is also distributed uniformly throughout the arm. peripheral in the air gap.
- * The arm. flux will take an action on main flux distribution which is called as arm. reaction.

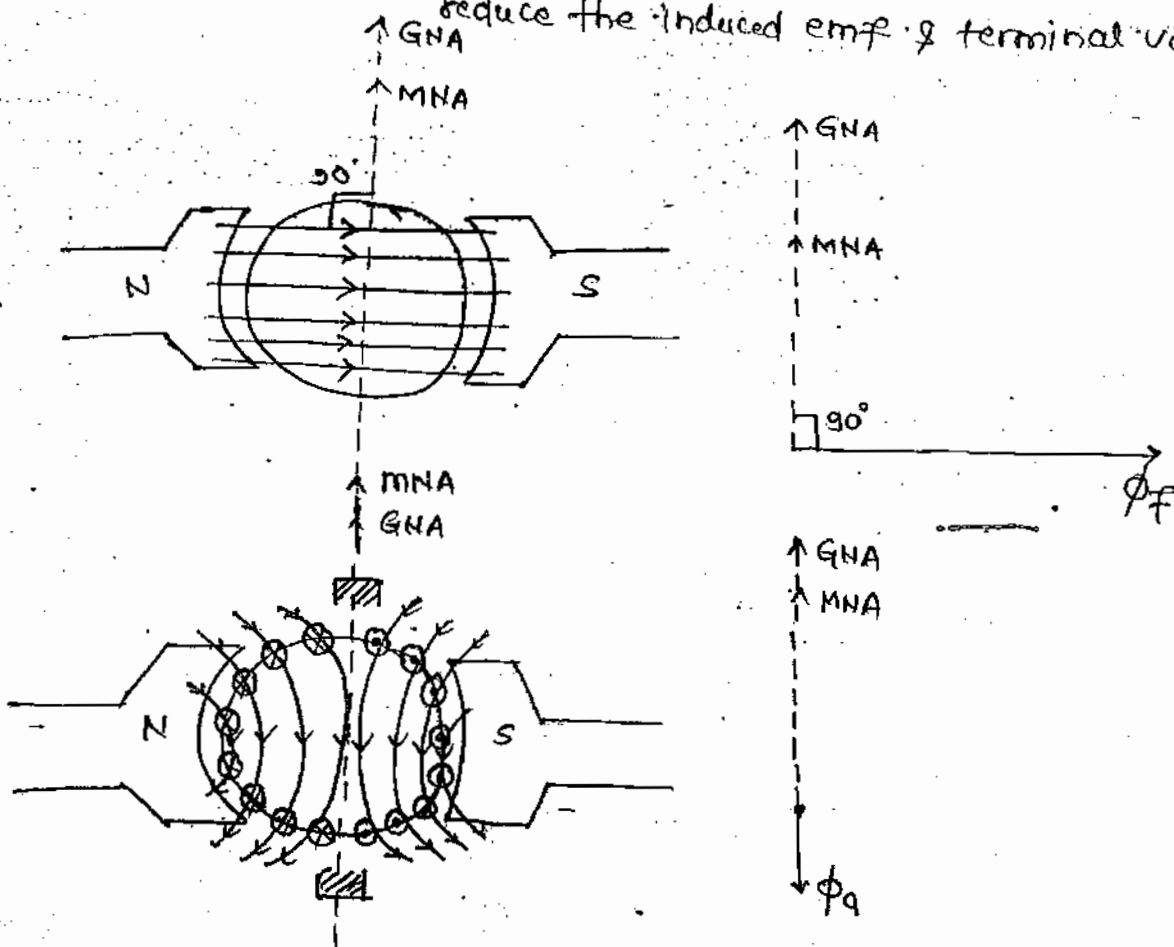
* The effect of the arm. flux on main flux produce :-

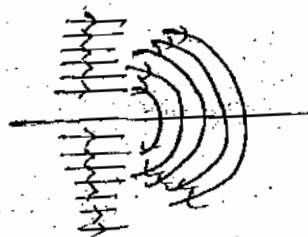
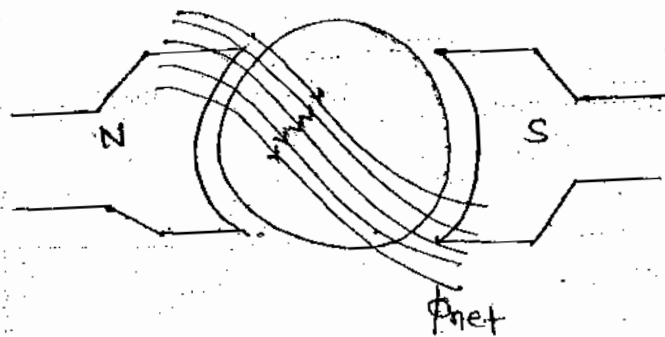
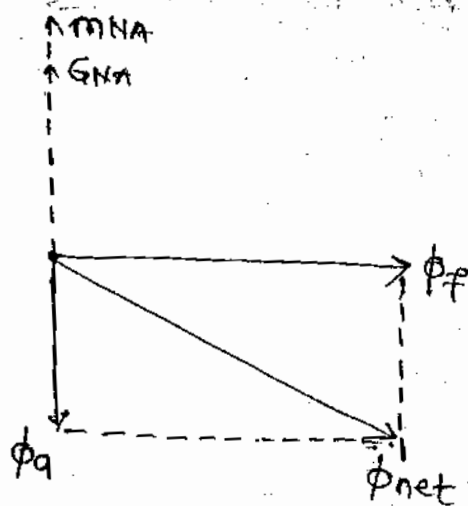
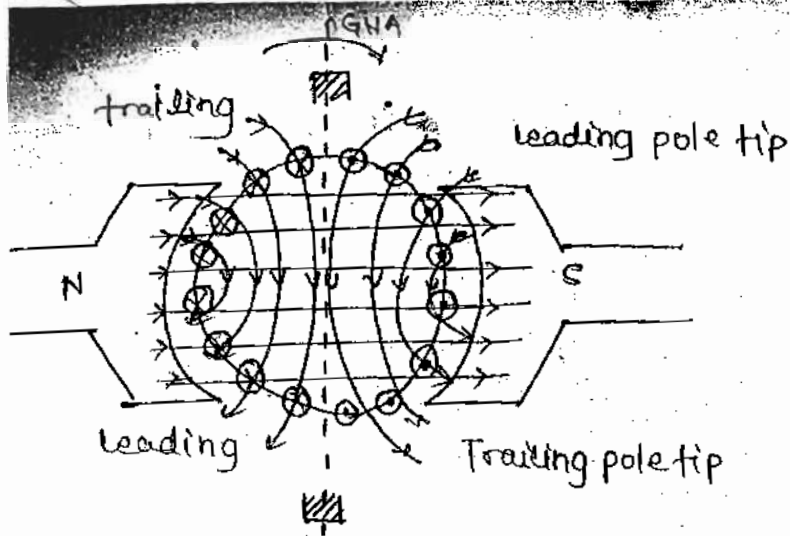
(1.) Cross magnetisation.

(2.) Demagnetisation.

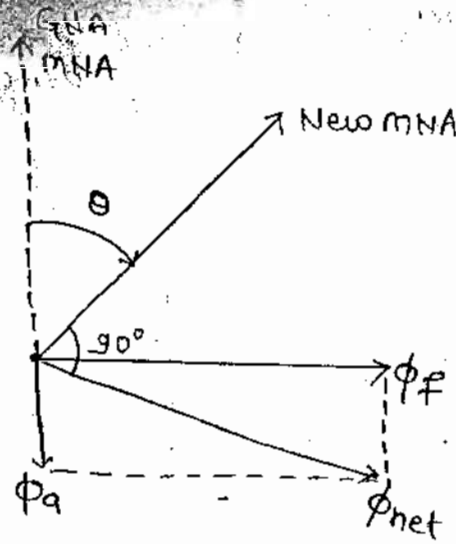
Cross magnetisation → It is distortion of main flux due to which mna is shifted. & the commutation will not be successful due to sparking.

Demagnetisation → It is reduction in the main flux which reduce the induced emf & terminal voltage



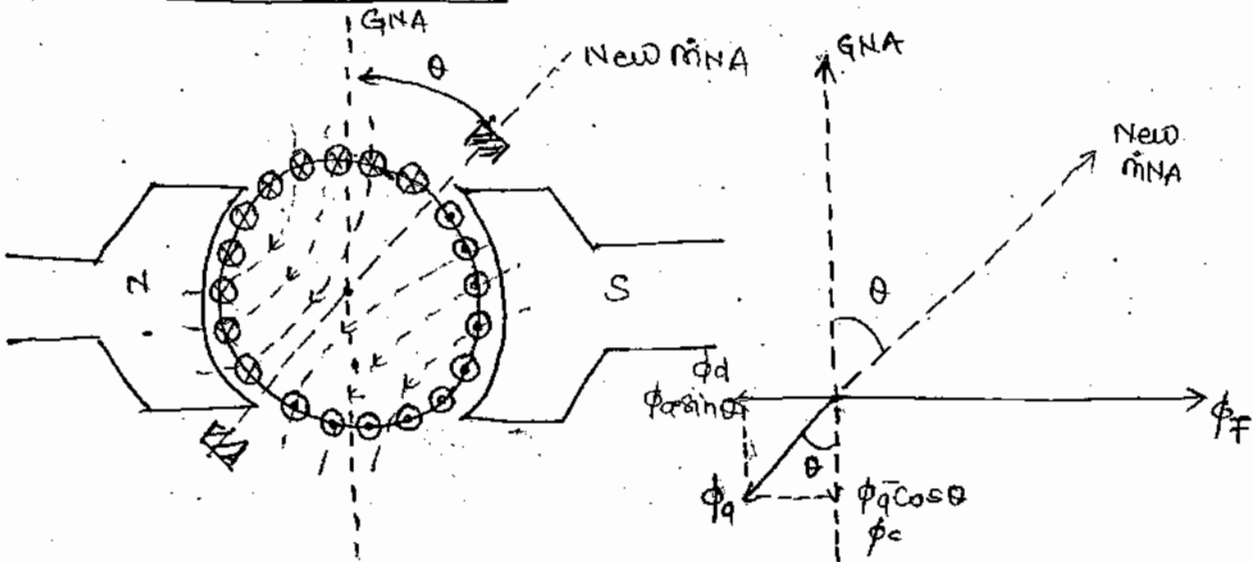


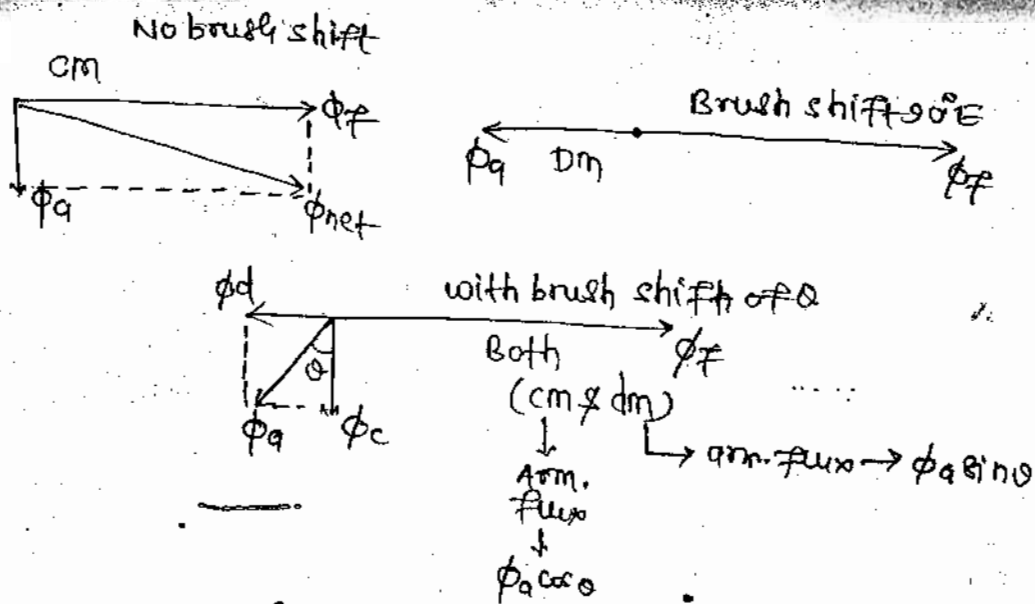
- * Depending on the dirⁿ of rotation the pole tips are named as leading & trailing.
- * The arm. flux under the trailing pole tips of gen^r will increase the flux density as it is in the same dirⁿ with the main flux.
- * The arm. flux under the leading pole tips will demagnetise the main flux as it is in the opposite dirⁿ.
- * If the amount of magnetisation & amt of Dm are equal then there is no net reduction in main flux but only distortion.
- * Under practical condⁿ the poles get magnetically saturated & consequently the increase in flux density under trailing pole tips is comparatively less than that of decrease in the flux under leading pole tips.



- * Due to arm. flux which is also called as cross flux the main flux is distorted. known as cross magnetised due to which mNA is shifted in the dirⁿ of rotation of gen^r.
- * In order to improve commutation the brushes also need to be shifted in the dirⁿ of rotation of gen^r.
- * Due to the affect of brush shift to an angle θ there exist additional demagnetisation.
- * Brush shift is not done generally. It has been replaced with interpole as it is not reliable method & also produce additional demagnetisation.

Affect of brush shift →





ϕ_c or $\phi_a \cos \theta$ is cross magnetisation of arm. flux

$\phi_d = \phi_a \sin \theta$ is demagnetisation of arm. flux

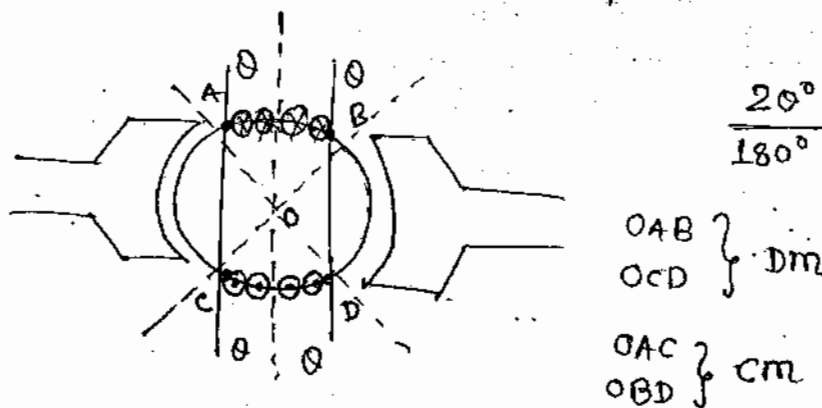
when $\theta = 0$

$\phi_d = 0, \phi_c = \phi_a$ (arm. flux is cm in nature)

when $\theta = 90^\circ$

$\phi_d = \phi_a, \phi_c = 0$ (arm. flux is demagnetisation in nature)

De-magnetising Amp-turns/pole $AT_d/p \rightarrow$



$$\frac{2\theta}{180^\circ} \times \frac{I_a}{A} \times \frac{Z}{2P}$$

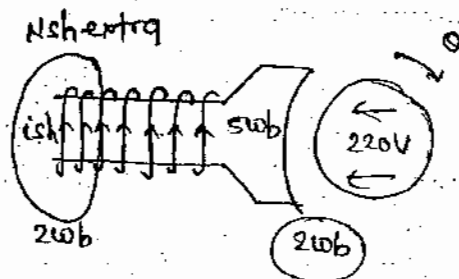
Gross magnetising amp-turns/pole \rightarrow

$$AT_c/p = \frac{180-2\theta}{180^\circ} \times \frac{I_a}{A} \times \frac{Z}{2p}$$

$$\text{Total arm. mmf} = \frac{I_a}{A} \times \frac{Z}{2p}$$

* AT_d/p is representing the arm. flux which produce additional demagnetisation.

* In order to compensate this additional demagnetisation extra amp-turns need to be provided on each pole.



$$N_{sh \text{ extra}} \times I_{sh} = AT_d/p$$

$$N_{sh \text{ extra}} = \frac{AT_d/p}{I_{sh}}$$

$$\frac{AT}{A} = T$$

* The no. of extra turns to be added on each pole of shunt gen^r in order to compensate additional demagnetisation produced by the brush shift is equal to $AT_d/p / I_{sh}$.

* Similarly in a series gen^r

$$N_{se \text{ extra}} = \frac{-AT_d/p}{I_a \text{ or } I_{se}}$$

DATE - 07/07/19

* Upper effects of arm. reaction →

(1) Decrease in efficiency due to increased iron loss →

The increased flux density under the pole tips will increase iron loss in the core as iron loss is directly proportional to flux density.

(2) Increased maintenance & repair →

Due to CM Commutation is not successful & there will be unacceptable sparking which damage the brush surface.

(3) Increased design cost →

* Methods to reduce armature reaction & its effects →

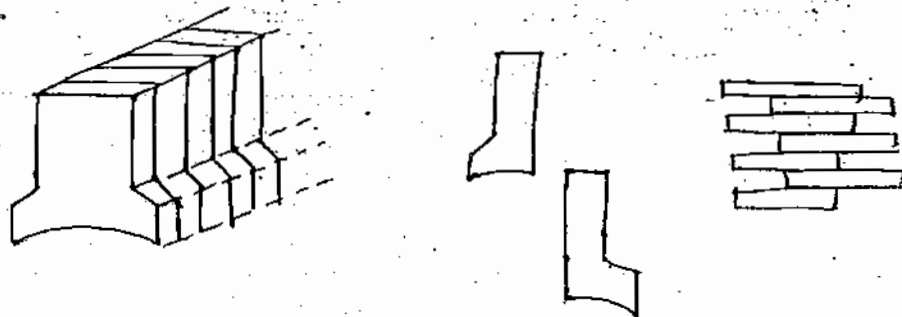
(1) Pole stacking

(2) Pole chamfering

(3) Pole core slotting

(4) ^{**} Compensating wdg.

(1) Pole stacking →



* The pole laminations are alternately stacked to introduced air gap under the pole tips.

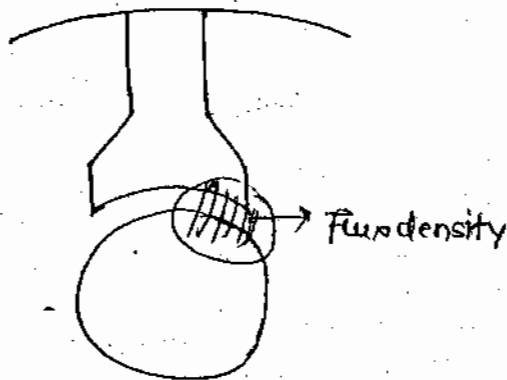
* The increased reluctance will reduce the flux density & reduce the iron loss but the net reluctance in the m/c increased which demands more mmf which increase the size & cost of m/c.

-stacking → ↑ Reluctance

② Pole chamfering →

min^m reluctance at the center & increased reluctance towards the pole tips.

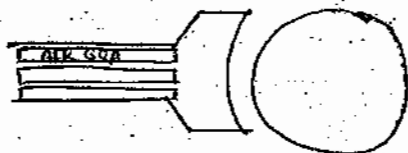
Concept is same as pole stacking.



③ Pole core slotting →

* The pole core contains rectangular slots to introduce air gap to some part of the flux, & reduces it to some extent.

$$\text{Flux} = \frac{\text{MMF}}{\text{Reluctance}}$$



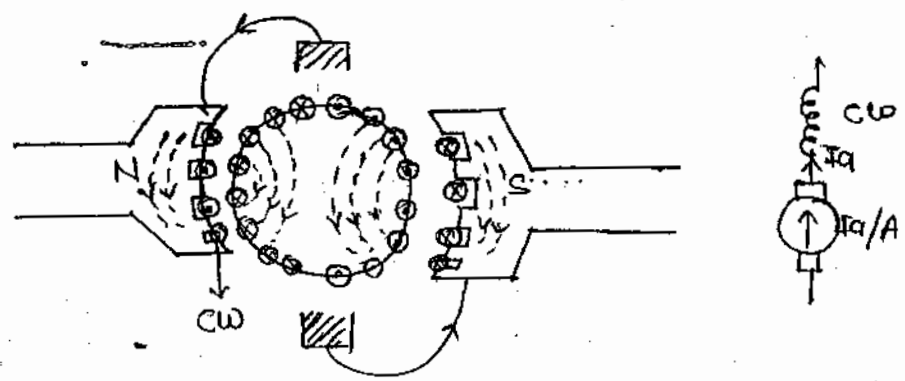
④ Compensating winding →

* In large rating dc. m/c operating at varying load condⁿ running at high speeds compensating wdg are essential in order to reduce flashover on the commutator.

* As the load vary arm. current vary & produce a varying flux on the arm. links with arm. condⁿ. & produce statically induced emf, which results in circulating currents which interfere with commutation & produce sparking.

* Due to high speeds if the spark spreade it will become a flash over to damage the wdg. Compensating wdg is provided in the pole shoe or pole face by cutting into teeth or slots.

* It is always connected in series with arm wdg. through brushes
 * In order to automatically neutralise the arm. flux under the pole.



* The current flowing in the CW under any pole should exactly opposite dirn to the arm cond^r current dirn under the pole. in order to cancel out the arm. flux

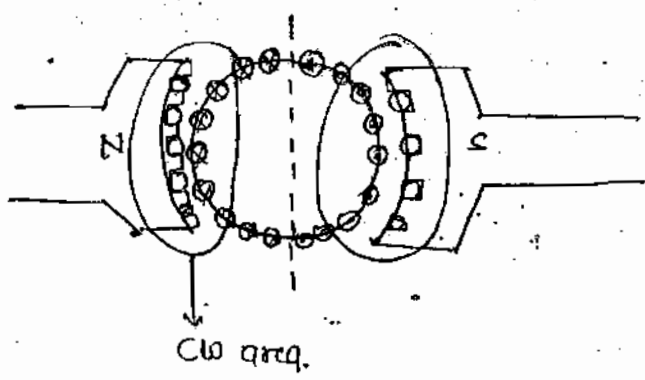
Let z_c be CW cond^r, z be AW cond^r

$$I_a \cdot \frac{z_c}{2} = \frac{I_a}{A} \cdot \frac{z}{2}$$

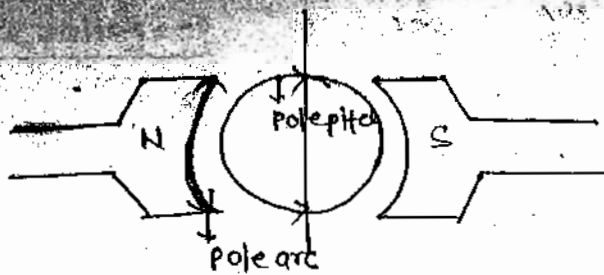
$$z_c = z/A$$

No. of compensating wdg. under each pole

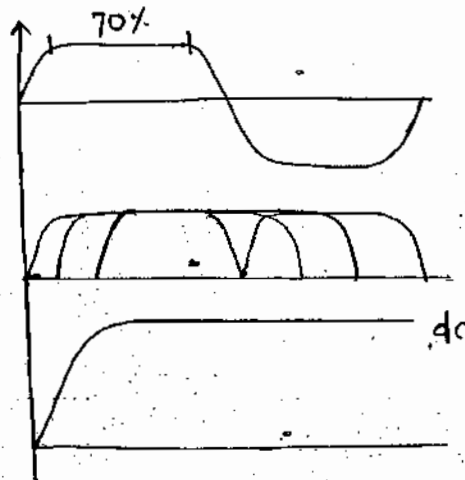
$$z_c = z/A_p$$



Pole arc	factor = 0.7 or 70%
Pole pitch	



* Flux distribution under the pole in dc m/c is Flat topped nature.



$$Z_c = \frac{Z}{A_p} \times \frac{\text{Pole Arc}}{\text{pole pitch}}$$

Because Z_c is placed on pole arc region that's why multiplying with pole arc/pole pitch.

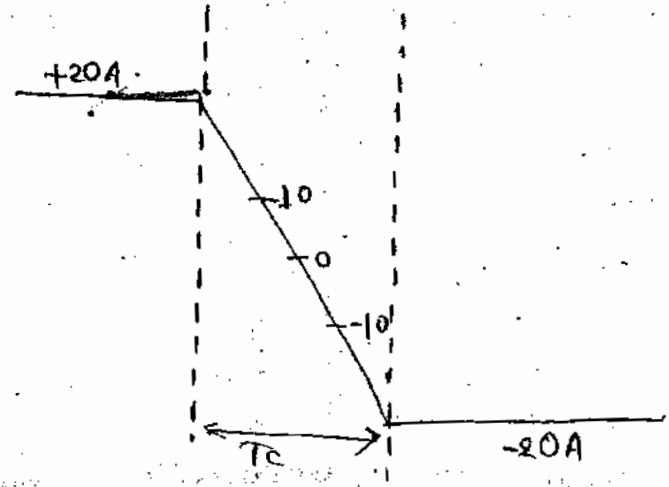
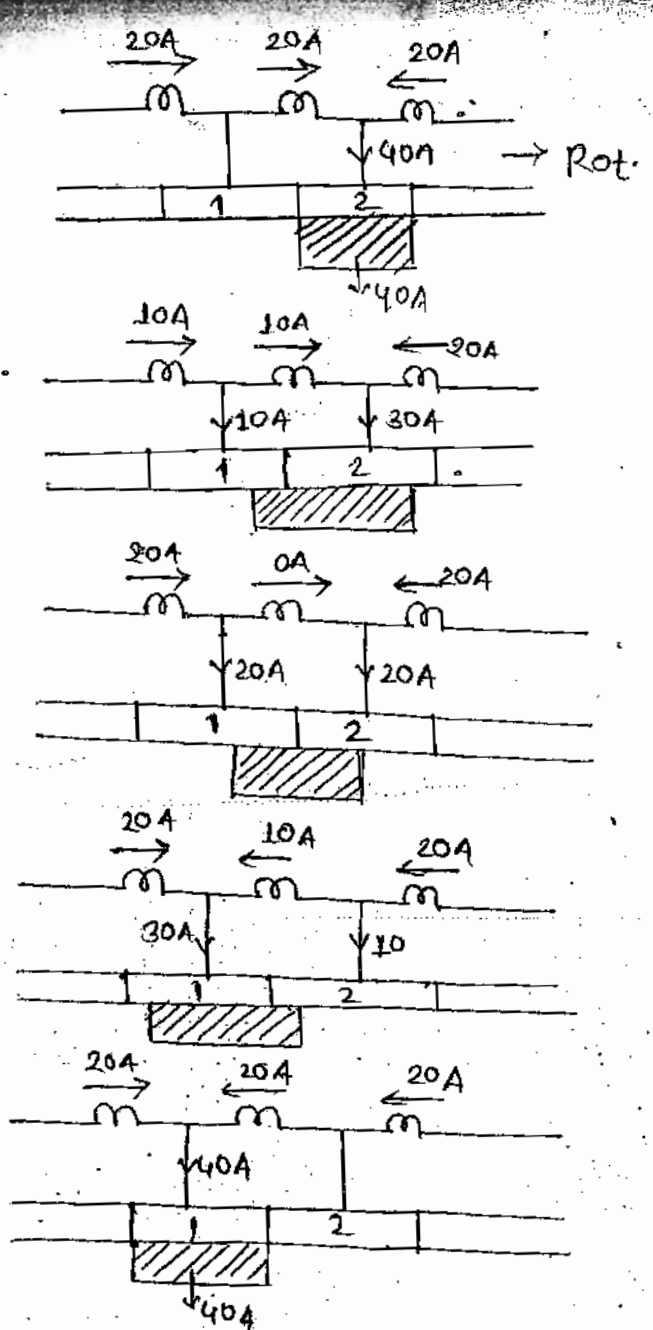
* COMMUTATION → The process of current reversal in the coil when it passes through a brush is known as commutation.

* The time taken for the brush to span from 1 segment to the other is known as commutating time.

* If the current reverses completely within the commutation time in the coil undergoing commutation then commutation is successful also called as linear, ideal or straight line commutation.

* There will be no sparking at the brush.

* If the current doesn't reverse completely within the commutating time in the coil undergoing commutation there will be sparking at the brush & the commutation is unsuccessful known as delayed commutation or non-linear.



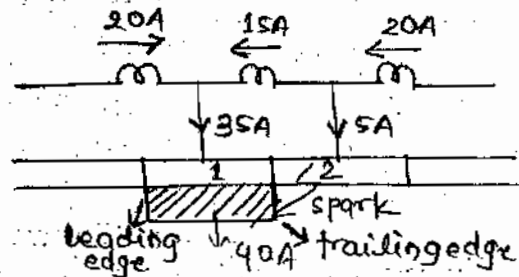
Reactance voltage →

$$e_r = N \frac{d\phi}{dt} = L \frac{di}{dt}$$

$\nearrow 2I$
 $\rightarrow T_c$
 $\rightarrow L_{self}$

* In the coil undergoing commutation there is a total change of current from I_a/A to $-I_a/A$ within commutating time T_c which produce the reactance voltage due to self inductance property of the coil.

* According to Lenz law it will oppose the cause of change in current. Therefore by the end of commutating time T_c the current will not be reversed completely.



* Any unchanged current by the end of commutating time will jump into the brush through spark at the trailing edge of brush.

Methods to improve commutation →

(1) Resistance Commutation

(2) Voltage Commutation
 $\begin{cases} \rightarrow \text{Brush shift} \\ \rightarrow * \text{Interpole} \end{cases}$

① Resistance Commutation → * Replacing low resistance Cu brush

with high resistance C brush to improve

commutation by reducing the chance of sparking to some extent.

* C brushes have high resistance compare to Cu.

* Due to its high resistance C brush doesn't encourage sparking at trailing edge & improve commutation.

* The added adv. of c brush are:-

- (1) It is not hard material as Cu.
- (2) It is self lubricating (polishing) in nature which offers good mech. condⁿ with the comm. surface.
- (3) If any spark occurs it will get less damaged than Cu.

disad. →

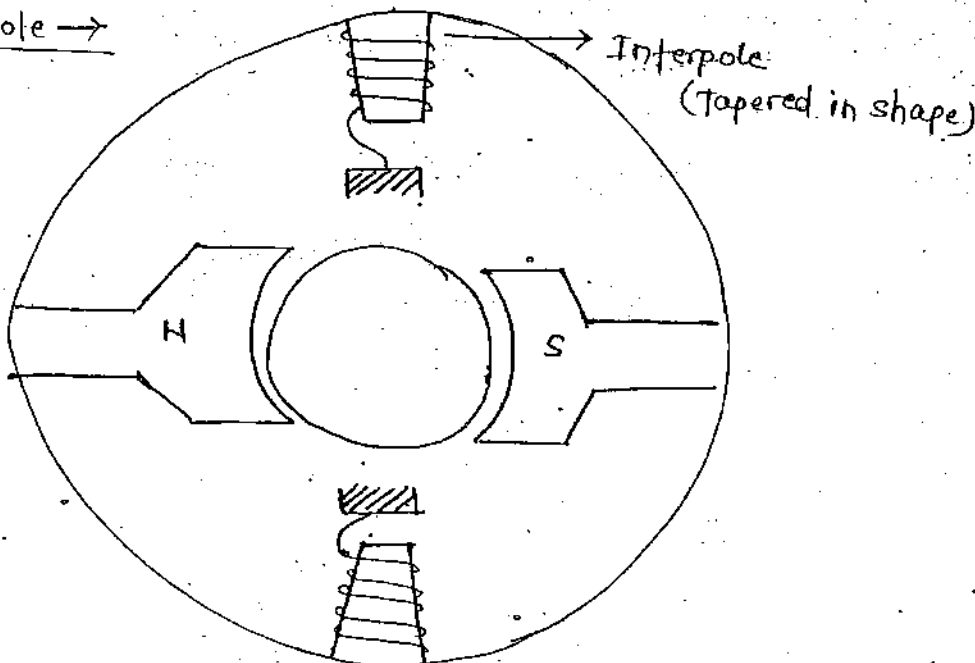
- ① more brush contact drop.
- ② Low current density, requires larger brush.

* Brush shift → * It is the 1st method of improving comm. which is complicated & not reliable because $m \neq 1$ changed with loads continuously & after the design brushes can't be changed.

* Due to brush shifting there is additional D_m .

* It is not ~~turn~~ done after the invention of interpole.

* Interpole →



* Interpoles are small poles compare to main poles placed in the interpolar region between the main poles on the yoke.

* These are also electromagnets with interpole wdg which is connected in series with arm. wdg through brushes in order to have automatic neutralisation of arm. flux in the interpolar region.

* It performs 2 funⁿ:-

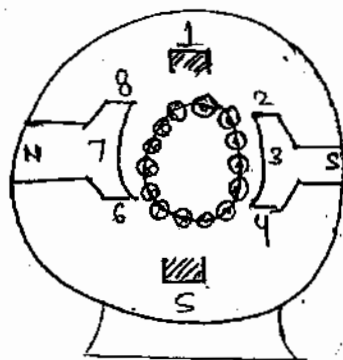
(1) Produce a counter flux on the coil undergoing commutation to nullify the reactance voltage (2) It also remove the inequality in the flux densities on the top & bottom region of arm.

* The interpoles are tapered in shape with comparatively more air gap in order to avoid easy saturation as the load current flows in the interpole wdg. The no. of turns on the interpoles are calculated acc to $\text{cm} \cdot \text{amp turns} / \text{pole}$.

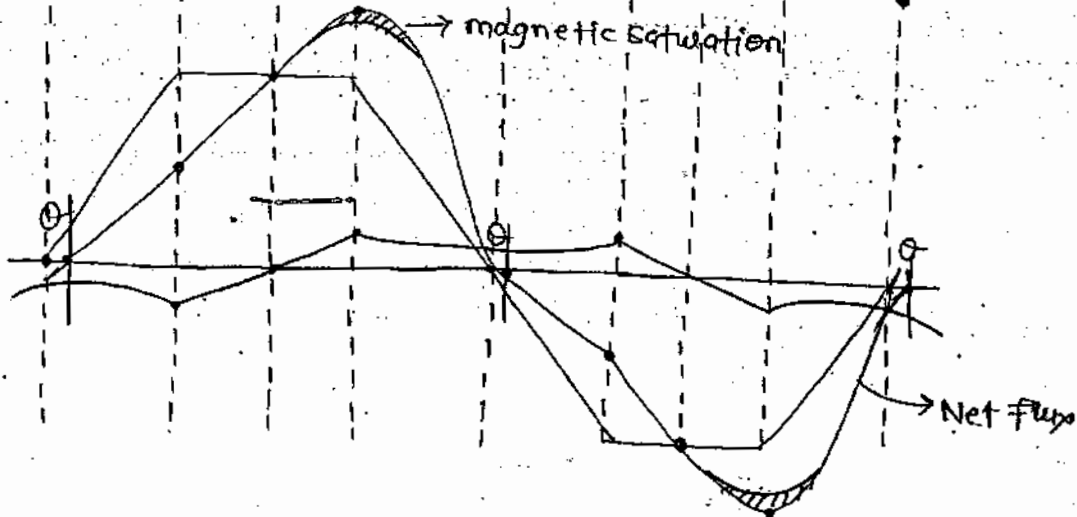
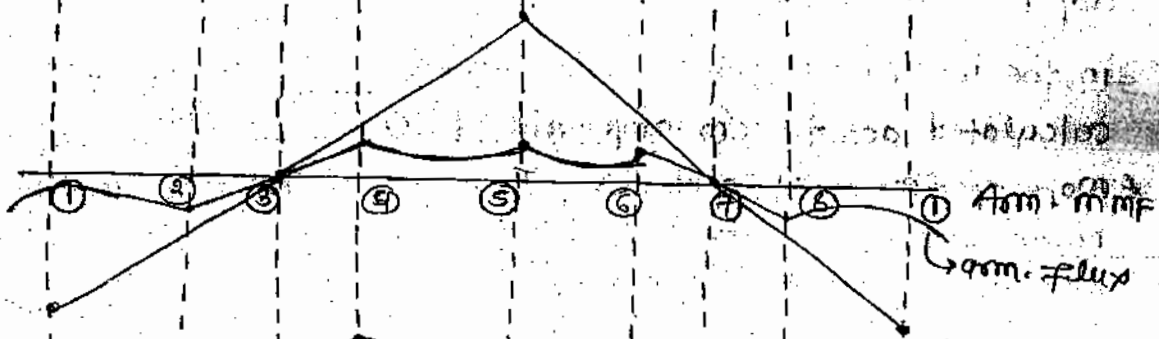
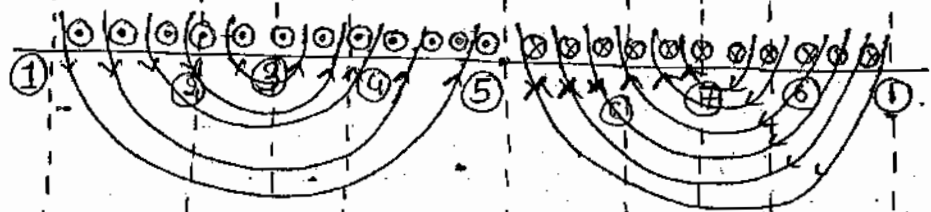
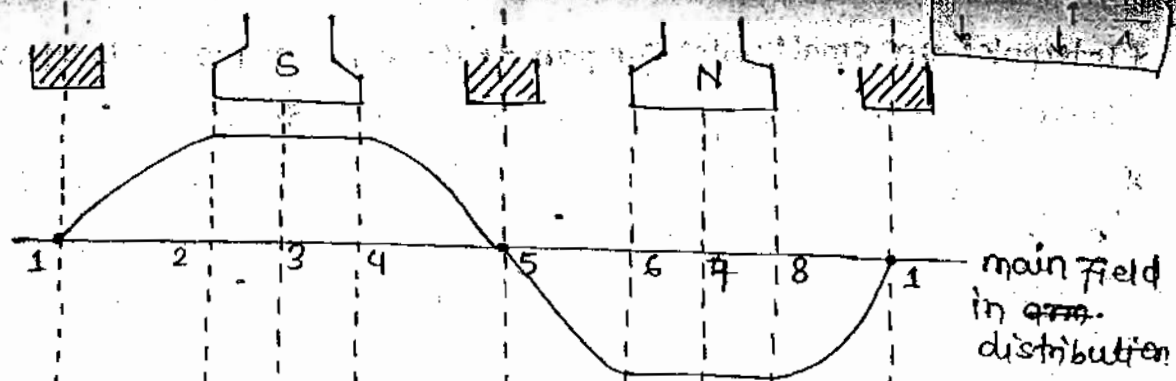
* More specifically it is the amp turns in $\frac{1}{2}$ of arm. in the interpolar region. & some additional turns are also required in order to produce reactance voltage.

CW	→	Reduces arm. reaction
IPW	→	Improve Commutation

Interpole are directly improve comm. & CW indirectly improve comm.



	turns (or) ampF
1-3	↓ -ve
3-5	↑ +ve
5-7	↓ +ve
7-1	↑ -ve
3-1	↑ -ve



(Reluctance)

* In the m/c if air gap is uniform then the arm. mmf & arm. flux follow the same shape but magnitude may vary.

$$\text{Armature Flux} = \frac{\text{Arm. mmf}}{\text{Reluctance (Air gap)}}$$

	Reluctance	
1-2	↓	-ve
2-3	↓ min ^m	-ve
3-4	↑	+ve
4-5	↑ max ^m	+ve
5-6	↓	+ve
6-7	↓ min ^m	+ve
7-8	↑	-ve
8-1	↑ max ^m	-ve

* For drawing arm. flux ^{mmf} arm. flux is taken as ref.

(1) Main field flux distribution is having flat topped shape, or trapezoidal.

(2) Arm. mmf is triangular in shape increasing or directed towards the brush axis.

(3) Net. flux due to arm. reaction is having peaky (If not given peaky then Δ)

(4) Arm. flux is saddle shape (If not given then prefer Δ).

*** The main flux & arm. flux are always 90° E wrt each other (quadrature or ⊥) / orthogonal.

* Without arm. flux the main flux is exactly 0 along GNA.

* With arm. flux the neutral (MNA) has been shifted in the dirn of rotation of gen^r ~~to~~ ~~away~~.

* To improve comm. the brushes need to be given a forward shift.

* The arm. flux stationary wrot poles.

$$\frac{62}{12}$$

$$I_a = \frac{50}{6} \times \frac{720}{2 \times 6}$$

AT/p

$$T = \frac{720}{2}$$

$$P = 6$$

$$AT/p = \frac{50 \times 720}{6 \times 2 \times 6} = 500 \text{ AT}$$

* Factors affecting terminal voltage of dc generator →

* When the gen^r is on NL the NL induced emf is E_0 . when it is loaded there is voltage drop due to arm. reaction demagnetisation.

$$E_0 - E_g = \text{Arm. reaction drop.}$$

* Due to arm. resistance there will be arm. resistance drop $I_a R_a$.

* These drops are proportional to load.

* In a separately excited generator there are only above two drops affecting the terminal voltage.

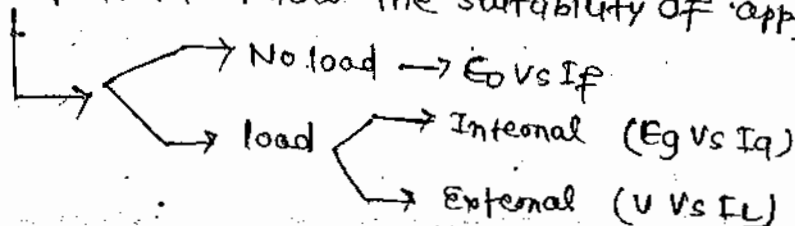
* In a shunt gen^r its excitation is the terminal vol.

Therefore reduction in V in turn reduces the terminal vol. itself.

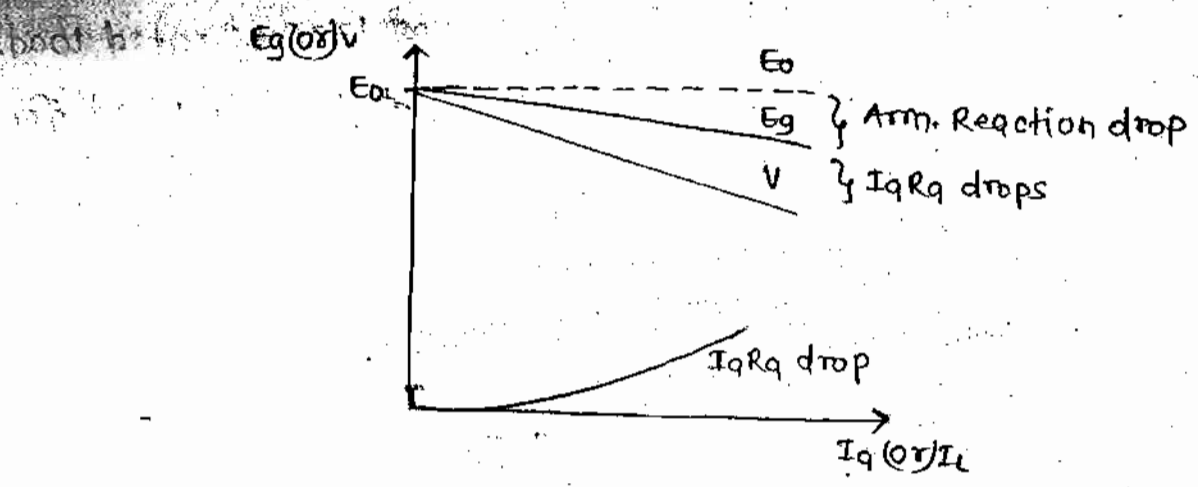
* This will be a cumulative effect when the gen^r is loaded beyond its rated value known as break down point.

* Characteristics →

* These are the graphical representation of the key parameters which are plotted to know the suitability of appliⁿ.

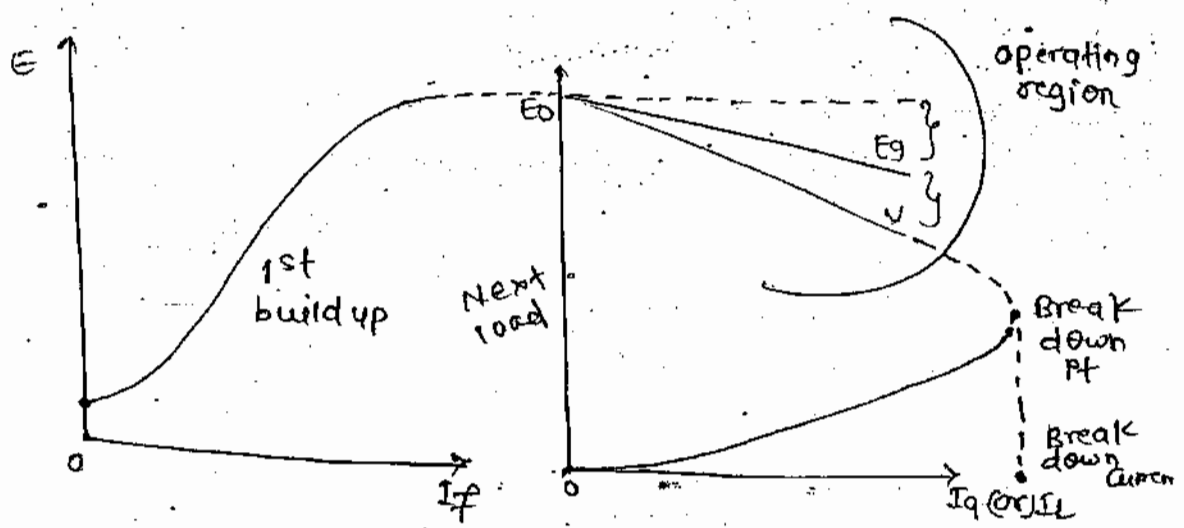


* Separately excited gen^s →



- * This are not commercially used for normal power supply because it requires additional dc. Vol. source.
- * It was used in excitation sys. of power plant gen^s.
- * Dc supplies in air crafts / ships.
- * Used in a speed control method known as Ward-Leonard method.

* Shunt gen^s →

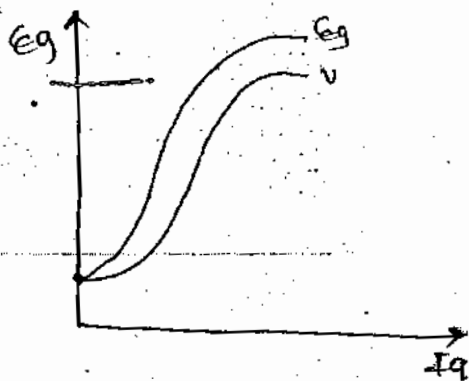


* In its operating region its flux remains approx. same, but beyond break-down value a cumulative reduction of vol. happen reduce the terminal vol. 0 drastically.

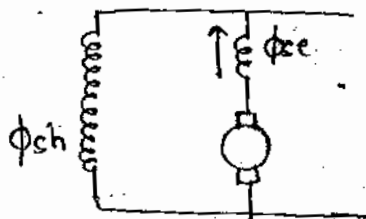
- * Generally the operating region will be around 125% of rated load.
- * It is used as small dc power supply & exclusively used for battery charging.
- * It was also used in excitation sys. of power plant gen^r along with separately excited gen^r.

* Series Gen^r →

* As the field wdg is in series with arm. & load as the load increases flux increases to in turn increase the voltage. Therefore in its operating region it has rising vol. c/s also called as variable vol. gen^r. not suitable for ordinary power supplies but used as boosters in long feeders, particularly in dc distribution sys.



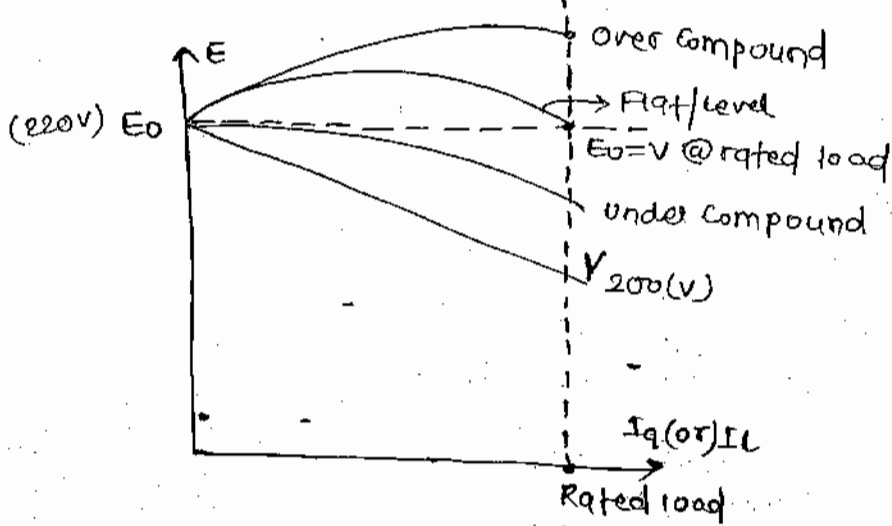
* Cumulative →



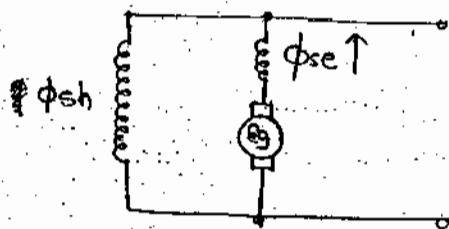
* As the series flux add its shunt flux the net flux increases with load. Consequently it has better vol. c/s. than separately & shunt gen^r.

* It can be compounded to adjust its terminal voltage.

Therefore it is widely manufactured dc gen^r due to its flexible c/s which can be used in large rating dc power supplies.



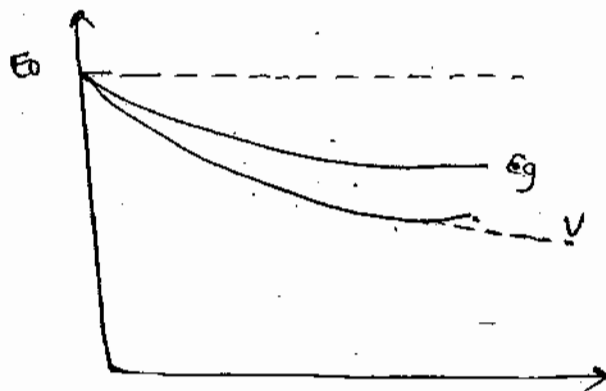
Differentially → -

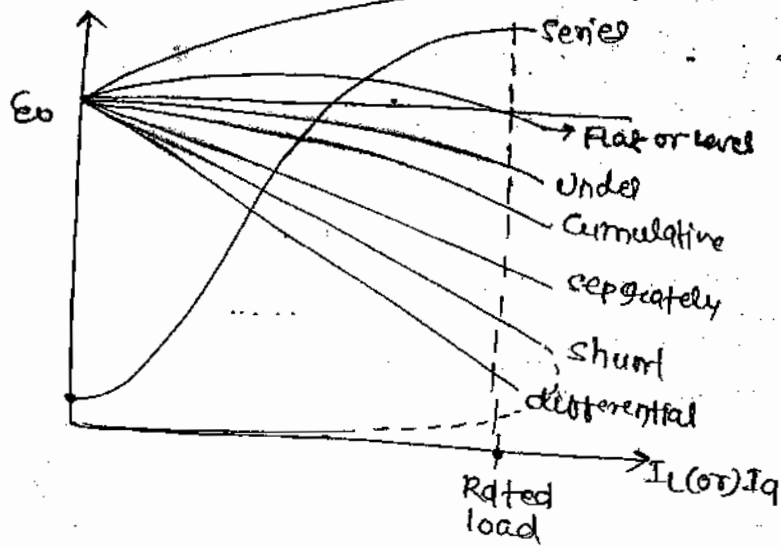


* Arc weldings - -

* The net flux decreases with load to decrease the terminal vol.

This is not use for ordinary power supply but specifically used in welding purposes to limit the welding currents.





Voltage Regulation → * It is the change in terminal vol. when the full or rated load across the terminal is disconnected. keeping the flux & speed constant.

$$\% VR = \frac{E - V}{V}$$

E → NL induced vol.

V → Rated terminal vol. @ rated load

* VR is % drop in the m/c.

$$VR = \frac{V_{drop}}{V}$$

$V_{drop} \downarrow$ as possible

* It should be as min^m as possible, best/ideal value is 0. which happens for only flat/level compound gen^r.

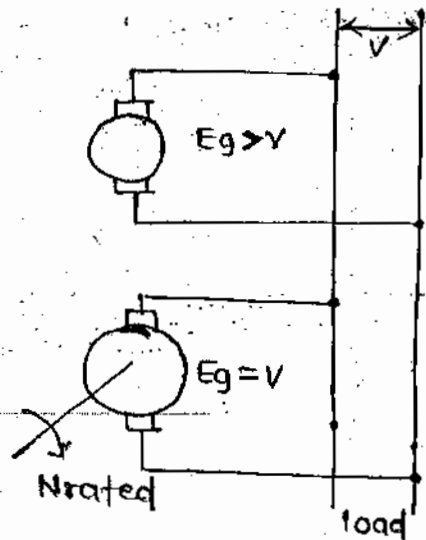
* For series & over compound VR becomes -ve. not suitable for ordinary load purpose.

* PARALLEL OPERATION →

Operating gen^r in parallel across the common terminals known as bus bars provide the following ad. :-

- (1) High ele inertia across the busbars (Constant voltage sys.)
- (2) High reliability.
- (3) Efficiency. (✓)
- (4) Future expansion.
- (5) Continuity of supply during maintenance & repairs.

Due to this reasons universally the gen^r in the power plants & all the power plants are operating in parallel to form a grid st.

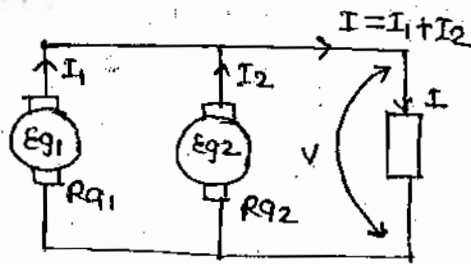


$E_g > V$	Generating mode
$E_g = V$	Floating mode
$E_g < V$	motoring mode

* In order to connect 2 dc gen^r in parallel it requires 2 essential condⁿ :-

- (1) Terminal vol. should be same.
- (2) Polarity should be matched.

* Consider 2 dc gen^r operating in parallel at a common terminal vol. V & sharing a common load with induced emfs E_{g1} , E_{g2} respectively.



$$E_{g1} - I_1 R_{a1} = V$$

$$E_{g2} - I_2 R_{a2} = V$$

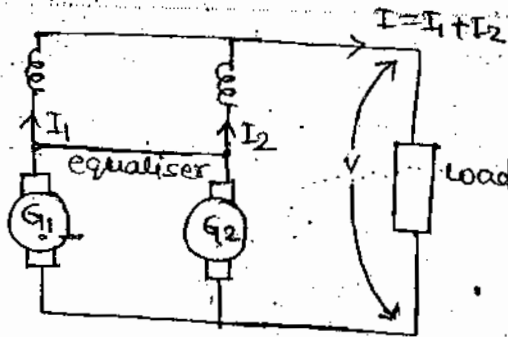
$$I_1 = \frac{E_{g1} - V}{R_{a1}}$$

$$I_2 = \frac{E_{g2} - V}{R_{a2}}$$

*** The load sharing of gen^r operating in parallel is significantly depending on induced emf (directionally proportional)

* In order to have stable or proper parallel operation the voltage c/s should be slightly drooping in nature but not rising.

* series gen^r in parallel →



$I_2 \uparrow \phi_{s2} \uparrow E_{g2} \uparrow$ (gain load sharing capability)

$I_1 \downarrow E_{g1} \downarrow$ (It goes in motoring mode)

* If we connect an equaliser with the both series gen^r then only it operate in parallel.

* If any one gen^r share more load the increase in its current will increased its flux & induced emf.

* Consequently its load sharing capability will increase in a cumulative manner & that gen^r gets overloaded leaving the other.

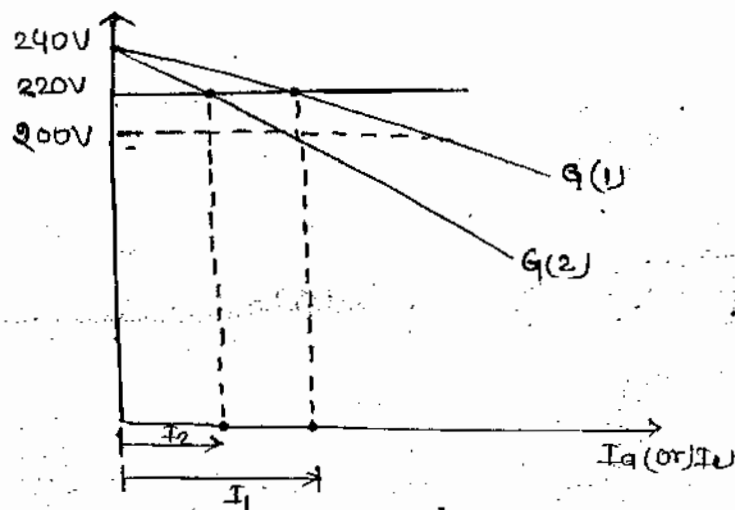
* This is due to the rising c/s of series gen^r.

* In order to make them in parallel an equaliser is required.

* The increased current will bypass into both field wdg to increase there Flux ϕ & induced emf equally.

* Equalisers are required for cumulative compound gen^r also.

* Shunt gen^r in parallel →



* Shunt gen^r are best suitable for parallel operation due to there drooping c/s.

* The gen^r which is more drooping will share less load (viceversa)

* By adjusting the c/s the gen^r can be loaded according to there ratings.