

CORRELATION

Autocorrelation & Cross-correlation Applications

Cross-correlation and autocorrelation are commonly used for measuring the similarity of signals especially for "pattern recognition" and, for "signal detection".

Example: Autocorrelation used to extract radar signals to improve sensitivity. Makes use of radar signals being periodic so the signal is a pulse train (parameters: amplitude, pulse width and interval between pulses).

Example: - Cross-correlation used to establish symbol timing by comparing an incoming signal with a known bit sequence to identify a known bit pattern to reference to for system timing.

Example: - Correlation is used for analyzing fractal patterns.

AUTOCORRELATION FUNCTION:-

Consider a time-limited (or band-limited) signal $x(t)$,

$$x(t) = \begin{cases} x(t) & 0 \leq t \leq T \\ 0 & \text{otherwise} \end{cases}$$

its autocorrelation function is defined as

$$C_{xx}(t, t+T) = E[x(t)x(t+T)]$$

$$\approx \frac{1}{T} \int_0^T x(t)x(t+T) dt$$

where the definition equation in the first line is specified for random signals whereas the second line is more general and also applicable for deterministic signals. If the random signal $x(t)$ is drawn from an ergodic stochastic process, then the ensemble average can be approximated by the time average by allowing the duration T to approach infinity.

Some important concepts and properties related to the autocorrelation are summarized here:

Properties :-

* If $x(t)$ is drawn from a wide-sense stationary process, then its autocorrelation function is shift invariant, namely,

$$C_{xx}(t, t+\tau) = C_{xx}(\tau)$$

* The autocorrelation function is symmetric, namely,

$$C_{xx}(\tau) = C_{xx}(-\tau) \text{ and } C_{xx}(\tau) \leq C_{xx}(0) = \sigma_x^2,$$

where $\sigma_x^2 = \text{Var}[x(t)]$ denotes the variance of the $x(t)$.

* The normalized autocorrelation function is defined as

$$\bar{C}_{xx}(\tau) = \frac{C_{xx}(\tau)}{C_{xx}(0)}$$

* The decaying rate and the limit of the autocorrelation function can be characterized by

$$|C_{xx}(\tau)| \leq C_{xx}(0) \cos\left(\frac{\pi}{T} \tau\right) \quad (T < \tau)$$

* If $x(t)$ is wide-sense stationary, its autocorrelation function can be written in terms of spectral representations in light of the Wiener-Khinchin theorem.

$$C_{xx}(\tau) = \frac{1}{2\pi} \int_{-\infty}^{\infty} S_{xx}(\omega) e^{j\omega\tau} d\omega,$$

where $S_{xx}(\omega)$ denotes the power spectral density of $x(t)$.

* Let $x_1(t)$ denote the Hilbert transform of $x(t)$.

$$x_1(t) = \int_{-\infty}^{\infty} \frac{x(\tau)}{t-\tau} d\tau;$$

then it can be proved that the autocorrelation of $x_1(t)$ is equal to that of $x(t)$, namely,

$$C_{x_1 x_1}(\tau) = C_{xx}(\tau),$$

whereas $x_1(t)$ is orthogonal (or uncorrelated) to $x(t)$, namely, $E[x_1(t)x(t)] = 0$.

CROSS-CORRELATION FUNCTION

For two time-limited signals $x(t)$ and $y(t)$, the cross-correlation function may be defined as

$$C_{xy}(t, t+\tau) = E[x(t)y(t+\tau)]$$

$$\approx \frac{1}{T} \int_0^T x(t)y(t+\tau) dt$$

$$C_{yx}(t, t+\tau) = E[y(t)x(t+\tau)]$$

$$\approx \frac{1}{T} \int_0^T y(t)x(t+\tau) dt$$

$$x(t) * y(t) = \int_{-\infty}^{\infty} y(\tau)x(t-\tau) d\tau$$

$$C_{xy}(\tau) = \int x(t)y(t-\tau) dt$$

It is noted that the cross-correlation function is generally nonsymmetric, namely, $C_{xy}(t, t+\tau) \neq C_{xy}(t+\tau, t)$.

Properties of cross-correlation :-

* The cross-correlation function is bounded by the cross-correlation inequality

$$|C_{xy}(\tau)|^2 \leq C_{xx}(0)C_{yy}(0) = \sigma_x^2 \sigma_y^2, \quad \text{--- (1)}$$

Where $\sigma_x^2 = E[x^2(t)]$ and $\sigma_y^2 = E[y^2(t)]$ denote the power of $x(t)$ and $y(t)$, respectively.

* In terms of spectral representations, the cross-correlation function can be written as the inverse Fourier transforms

$$C_{xy}(\tau) = \frac{1}{2\pi} \int_{-\infty}^{\infty} S_{xy}(\omega) e^{j\omega\tau} d\omega$$

Where $S_{xy}(\omega)$ denotes the cross-spectrum density.

* The correlation coefficient (also called normalized cross-correlation) between two random signals $x(t)$ and $y(t)$ is defined as

$$\rho_{xy} = \frac{C_{xy}(0)}{\sqrt{\text{var}[x(t)] \text{var}[y(t)]}}$$

From eq (1), it follows that the correlation coefficient ρ_{xy} ranges between -1 and 1 . Positive / negative ρ_{xy} indicates $x(t)$ and $y(t)$ are positively / negatively correlated; $\rho_{xy} = 0$ indicates that they are uncorrelated.

In the frequency domain, let $X(\omega)$ and $Y(\omega)$ denote the Fourier transform of $x(t)$ and $y(t)$, respectively, then the cross-spectrum of $X(\omega)$ and $Y(\omega)$ is defined as

$$S_{xy}(\omega) = E [X(\omega) Y^*(\omega)],$$

Where the asterisk denotes the complex conjugate. In a similar vein, the normalized cross-spectrum is defined as

$$\bar{S}_{xy}(\omega) = \frac{S_{xy}(\omega)}{\sqrt{\text{Var}[X(\omega)] \text{Var}[Y(\omega)]}}$$

and its magnitude $|\bar{S}_{xy}(\omega)|$ is a real function between 0 and 1 that gives a measure of correlation between $x(t)$ and $y(t)$ at each frequency, ω . Observe that

$|\bar{S}_{xy}(\omega)|^2$ bears some similarity to ρ_{xy}^2 ; however, $|\bar{S}_{xy}(\omega)|^2$ takes into account out-of-phase relationships and can examine the variance of two signals in a selected frequency range.

PDF Auto correlation function and its properties

*The Mean of $x(t)$ is defined as

$$\begin{aligned} \mu_x(t) &= E [x(t)] \\ &= \int_{-\infty}^{\infty} x f_{x(t)}(x) dx \end{aligned}$$

The correlation coefficient (also normalized) between two random variables $x(t)$ and $y(t)$ is defined as

*The Autocorrelation Function of $x(t)$ is

$$\begin{aligned} R_x(t_1, t_2) &= E [x(t_1) x(t_2)] \\ &= \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} x_1 x_2 f_{x(t_1) x(t_2)}(x_1, x_2) dx_1 dx_2 \end{aligned}$$

$$= \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} x_1 x_2 f_{X(t_1), X(t_2)}(x_1, x_2) dx_1 dx_2$$

$$= R_X(t_2 - t_1)$$

for all t_1 and t_2

* The Autocovariance Function

$$C_X(t_1, t_2) = E[(X(t_1) - \mu_X)(X(t_2) - \mu_X)]$$

$$= R_X(t_2 - t_1) - \mu_X^2$$

Which is a function of time difference $(t_2 - t_1)$.
 We can determine $C_X(t_1, t_2)$ if μ_X and $R_X(t_2 - t_1)$ are known.

Properties of the autocorrelation function :-

For convenience of notation, we redefine

$$R_X(\tau) = E[X(t)X(t+\tau)], \text{ for all } t$$

Property 1 :- The mean-square value

$$R_X(0) = E[X^2(t)], \tau = 0$$

Property 2 :-

$R_X(\tau)$ is an even function of τ

$$R_X(\tau) = R_X(-\tau)$$

$$R(-\tau) = E[X(t)X(t-\tau)]$$

put $t - \tau = u$

$$\therefore R(-\tau) = E[X(u)X(u+\tau)]$$

Property 3 :-

The maximum value of $R_X(\tau)$ is attained at $\tau = 0$

$$|R_X(\tau)| \leq R_X(0)$$

Consider $E[(X(t+\tau) \pm X(t))^2] \geq 0$

$$\Rightarrow E[X^2(t+\tau)] \pm 2E[X(t+\tau)X(t)] + E[X^2(t)] \geq 0$$

$$\Rightarrow 2E[X^2(t)] \pm 2R_X(\tau) \geq 0$$

$$\Rightarrow 2R_X(0) \pm 2R_X(\tau) \geq 0$$

$$\Rightarrow -R_X(0) \leq R_X(\tau) \leq R_X(0)$$

$$\therefore |R_X(\tau)| \leq R_X(0)$$

Property - 4:-

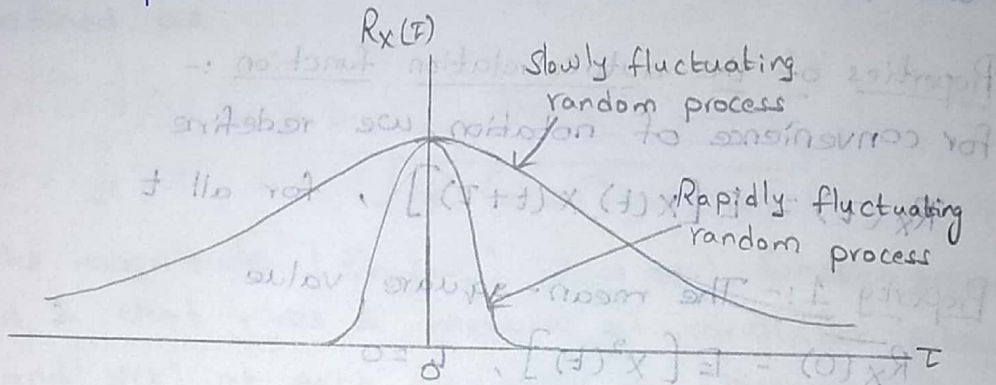
If a random process $\{X(t)\}$ has no periodic components and $E[X(t)] = \bar{X}$ then

$$\lim_{|\tau| \rightarrow \infty} R_{XX}(\tau) = (\bar{X})^2 = \mu_X^2$$

Property - 5:-

The auto correlation function of a random process cannot have an arbitrary shape.

The $R_X(t)$ provides the interdependence information of two random variables obtained from $X(t)$ at times t seconds apart.



Cross correlation and its properties:-

Let $\{X(t)\}$ and $\{Y(t)\}$ be two random processes.

Then the cross correlation between them is defined by

$$R_{XY}(t, t+\tau) = E[X(t)Y(t+\tau)] = R_{XY}(\tau)$$

$$R_{YX}(t, t+\tau) = E[Y(t)X(t+\tau)] = R_{YX}(\tau)$$

Note: Cross correlation function of two R.P is defined as a measure of the similarity between a signal and a time delayed version of a second signal.

Property - 1:-

Cross correlation function is not an even function

$$R_{XY}(\tau) \neq R_{YX}(\tau)$$

$$\text{But } R_{XY}(\tau) = R_{YX}(-\tau)$$

Proof:- Let $t-\tau = \mu$,

$$\Rightarrow R_{YX}(-\tau) = E[Y(t)X(t-\tau)]$$

$$R_{yx}(-\tau) = E[Y(\mu+\tau)X(\mu)] \\ = E[X(\mu)Y(\mu+\tau)] \\ = R_{xy}(\tau)$$

Property-2:-

If $\{X(t)\}$ and $\{Y(t)\}$ are 2 R.P's then

$$|R_{xy}(\tau)| \leq \sqrt{R_{xx}(0)R_{yy}(0)}$$

Proof:- We know that $E[(X(t) + KY(t+\tau))^2] \geq 0$ for any real K .

$$\Rightarrow E[X^2(t)] + 2KE[X(t)Y(t+\tau)] + K^2E[Y^2(t+\tau)] \geq 0$$

$$\Rightarrow R_{xx}(0) + 2R_{xy}(\tau) + K^2R_{yy}(0) \geq 0$$

If $a\lambda^2 + 2b\lambda + c \geq 0$ for all real λ

then $b^2 \leq ac$

$$\Rightarrow R_{xy}(\tau)^2 \leq R_{xx}(0)R_{yy}(0)$$

$$|R_{xy}(\tau)| \leq \sqrt{R_{xx}(0)R_{yy}(0)}$$

Property-3:-

If $\{X(t)\}$ and $\{Y(t)\}$ are 2 R.P's then

$$|R_{xy}(\tau)| \leq \frac{1}{2}[R_{xx}(0) + R_{yy}(0)]$$

Proof:- We know that the G.M of 2 positive numbers does not exceed their A.M. So we've

$$\sqrt{R_{xx}(0)R_{yy}(0)} \leq \frac{1}{2}[R_{xx}(0) + R_{yy}(0)]$$

So from Property 2 we've

$$|R_{xy}(\tau)| \leq \frac{1}{2}[R_{xx}(0) + R_{yy}(0)]$$

Property-4:- If the process $\{X(t)\}$ and $\{Y(t)\}$ are orthogonal then $R_{xy}(\tau) = 0$.

Property-5:- If the process $\{X(t)\}$ and $\{Y(t)\}$ are independent then $R_{xy}(\tau) = \mu_x \mu_y$.

Problem:-

1) A stationary process has an autocorrelation function given by $R_{xx}(\tau) = \frac{25\tau^2 + 36}{6.25\tau^2 + 4}$. Find the mean and variance of the process.

Sol:- $(\bar{x})^2 = \lim_{|\tau| \rightarrow \infty} R_{xx}(\tau) = \lim_{|\tau| \rightarrow \infty} \frac{\tau^2 (25 + \frac{36}{\tau^2})}{\tau^2 (6.25 + \frac{4}{\tau^2})}$

$= 4 \Rightarrow \text{Mean} = 2$

$E[x^2(t)] = R_{xx}(0) = 9$

Variance $= E[x^2(t)] - E[x(t)]^2 = 5$

2) If $\{x(t)\}$ is a WSS process with autocorrelation function $R_{xx}(\tau)$ and if $y(t) = x(t+a) - x(t-a)$, show that $R_{yy}(\tau) = 2R_{xx}(\tau) - R_{xx}(\tau+2a) - R_{xx}(\tau-2a)$

Sol:- $R_{yy}(\tau) = E[y(t) \cdot y(t+\tau)]$

$= E[(x(t+a) - x(t-a)) \cdot (x(t+\tau+a) - x(t+\tau-a))]$

Apply

$E[x(t+a) \cdot x(t+\tau+a)] = R_{xx}(\tau)$

So $R_{yy}(\tau) = 2R_{xx}(\tau) - R_{xx}(\tau+2a) - R_{xx}(\tau-2a)$

3) Two R.P's $\{x(t)\}$ and $\{y(t)\}$ are given by $x(t) = A \cos(\omega t + \theta)$ and $y(t) = A \sin(\omega t + \theta)$ where A and ω are constants and θ is uniformly distributed over $(0, 2\pi)$. Find the cross correlation.

Sol:- $R_{xy}(\tau) = E[x(t) \cdot y(t+\tau)]$
 $= E[A \cos(\omega t + \theta) \cdot A \sin(\omega t + \omega\tau + \theta)]$

$= \frac{A^2}{2} [\sin(2\omega t + \omega\tau + 2\theta) + \sin(\omega\tau)]$

Since θ is uniform in $(0, 2\pi)$, $f(\theta) = \frac{1}{2\pi}$

$R_{xy}(\tau) = \frac{A^2}{4\pi} \int_0^{2\pi} [\sin(2\omega t + \omega\tau + 2\theta) + \sin(\omega\tau)] d\theta$

$$= \frac{A^2}{4\pi} \left[\frac{\cos(2\omega t + \omega T + 2\theta)}{\theta} + \sin(\omega T)(\theta) \right]_0^{2\pi}$$

$$= \frac{A^2}{4\pi} [2\pi \sin(\omega T)]$$

$$= \frac{A^2}{2} \sin(\omega T)$$

Relation between convolution and correlation

Let $x_1(t)$ & $x_2(t)$

$$x_1(t) * x_2(t) = \int_{-\infty}^{\infty} x_1(\tau) x_2(t-\tau) d\tau$$

Now convolution of $x_1(t)$ & $x_2(-t)$

$$x_1(t) * x_2(-t) = \int_{-\infty}^{\infty} x_1(\tau) x_2(\tau - t) d\tau$$

Let $\tau = l$

$$= \int_{-\infty}^{\infty} x_1(l) x_2(l - t) dl$$

Now change t to τ

$$= \int_{-\infty}^{\infty} x_1(l) x_2(l - \tau) dl$$

$$x_1(t) * x_2(-t) = R_{x_1 x_2}(\tau)$$

$$x_1(t) * x_2(t) = R_{x_1 x_2}(\tau)$$

Suppose $x_2(t)$ is an even function

$$x_2(t) = x_2(-t)$$

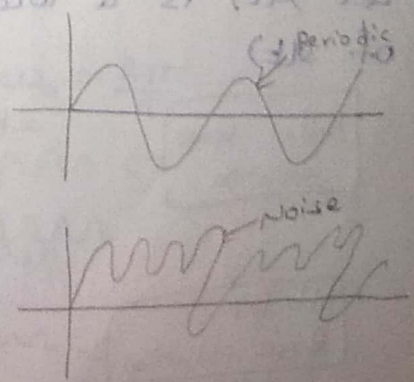
then convolution = correlation.

Detection of periodic signals in the presence of noise by correlation

If signal is $s(t)$
& noise is $n(t)$

Then the noise effected signal is

$$y(t) = s(t) + n(t) \quad \text{--- (1)}$$



Case-1:- Detection by Auto correlation

Let $R_{yy}(T)$, $R_{ss}(T)$ & $R_{nn}(T)$ are auto correlations of $y(t)$, $s(t)$ & $n(t)$ respectively.

Now autocorrelation of periodic signal $y(t)$ is

$$R_{yy}(T) = \lim_{T \rightarrow \infty} \frac{1}{T} \int_0^T y(t) y(t-T) dt \quad \text{--- (2)}$$

substitute (1) in (2)

$$\begin{aligned} &= \lim_{T \rightarrow \infty} \frac{1}{T} \int_0^T [s(t) + n(t)] [s(t-T) + n(t-T)] dt \\ &= \lim_{T \rightarrow \infty} \frac{1}{T} \int_0^T [s(t) \cdot s(t-T) + s(t)n(t-T) + n(t)s(t-T) + n(t)n(t-T)] dt \\ &= \lim_{T \rightarrow \infty} \frac{1}{T} \left[\int_0^T s(t) \cdot s(t-T) dt + \int_0^T s(t)n(t-T) dt + \int_0^T n(t)s(t-T) dt + \int_0^T n(t)n(t-T) dt \right] \end{aligned}$$

$$R_{yy}(T) = R_{ss}(T) + R_{sn}(T) + R_{ns}(T) + R_{nn}(T) \quad \text{--- (3)}$$

Since the periodic signal & noise are uncorrelated, then $R_{sn}(T) = R_{ns}(T) = 0$

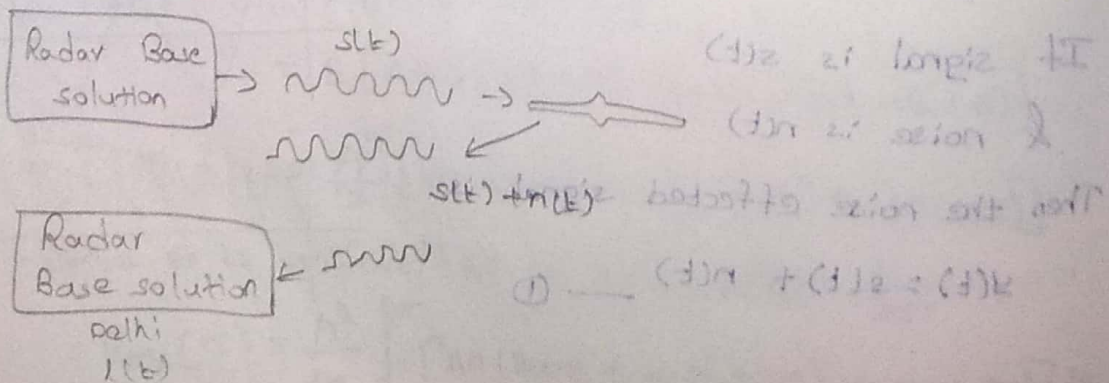
$$\text{(3) } \Rightarrow R_{yy}(T) = R_{ss}(T) + R_{nn}(T)$$

for larger values of T , the non periodic noise, $R_{nn}(T) = 0$ then $R_{yy}(T) = R_{ss}(T)$

Case-2:- Detection of periodic signal by cross correlation

$$\text{Let } y(t) = s(t) + n(t) \quad \text{--- (1)}$$

Let $\lambda(t)$ is a locally generated signal of same frequency of $s(t)$



Now the cross correlation of $y(t)$ & $l(t)$

$$R_{yl}(\tau) = \lim_{T \rightarrow \infty} \frac{1}{T} \int_0^T y(t) l(t-\tau) dt \quad \text{--- (2)}$$

Substitute (1) in (2)

$$\begin{aligned} R_{yl}(\tau) &= \lim_{T \rightarrow \infty} \frac{1}{T} \int_0^T [s(t) + n(t)] l(t-\tau) dt \\ &= R_{sl}(\tau) + R_{nl}(\tau) \end{aligned}$$

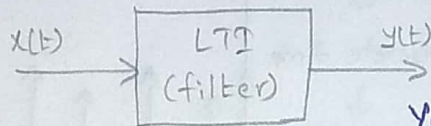
since n & l are uncorrelated

$$R_{nl}(\tau) = 0$$

$$\therefore R_{yl}(\tau) = R_{sl}(\tau)$$

since s & l are same frequency $R_{sl}(\tau)$.

Extraction of a signal from noise by filtering



$$Y(\omega) = X(\omega) H(\omega)$$

$$R_{yh}(\tau) = X(\omega) H(-\omega)$$

Let $s(t)$ = periodic signal

$n(t)$ = Noise signal

The received signal

$$y(t) = s(t) + n(t)$$

$s(t)$ can be detected in the presence of noise by cross correlating $y(t)$ with $l(t)$ which is same as $s(t)$

$$l(t) \xleftrightarrow{F.T} L(\omega)$$

$$l(-t) \xleftrightarrow{F.T} L(-\omega)$$

Since $l(t)$ is a periodic signal, then F.S of $l(t)$ is

$$l(t) = \sum_{k=-\infty}^{\infty} C_k e^{jk\omega_0 t}, \quad \omega_0 = \frac{2\pi}{T}$$

The F.T of $l(t)$ is

$$L(\omega) = 2\pi \sum_{k=-\infty}^{\infty} C_k \delta(\omega - k\omega_0)$$

$$L(-\omega) = L^*(\omega)$$

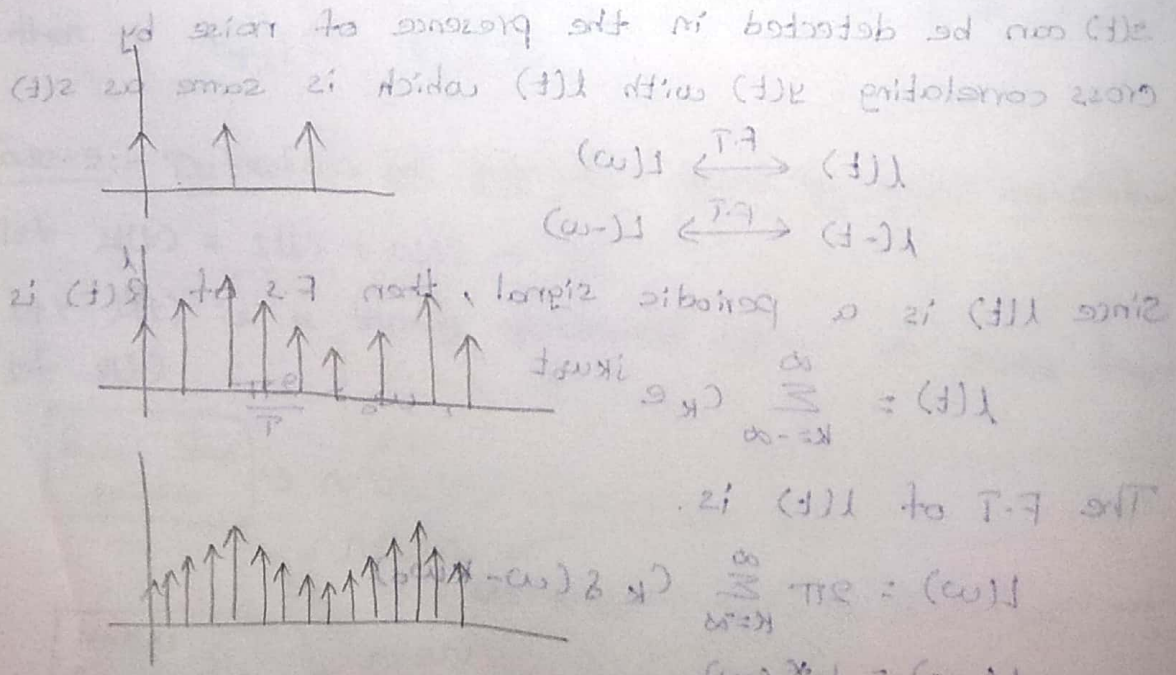
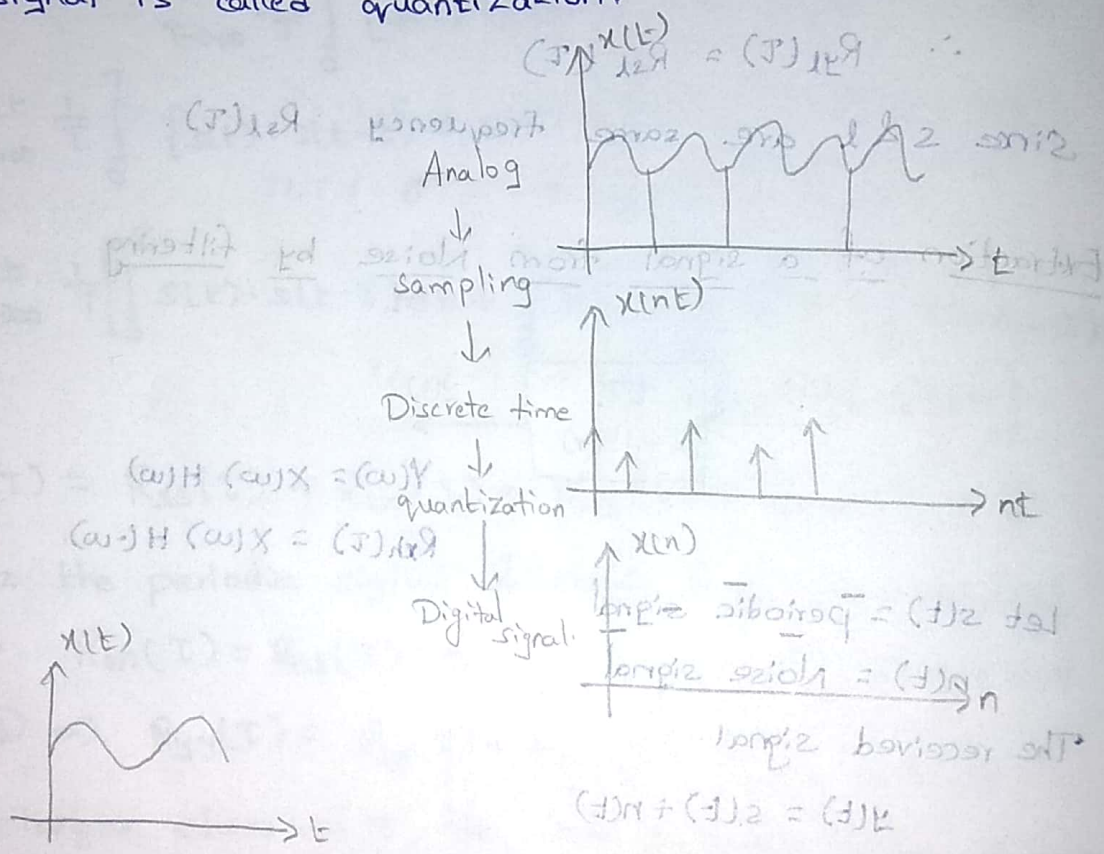
$$= 2\pi \sum_{k=-\infty}^{\infty} C_k^* \delta(\omega - k\omega_0)$$

The cross correlation in the time domain is equivalent to filtering in the frequency domain, which allows the frequencies of $s(t)$ and its harmonics.

Sampling:-

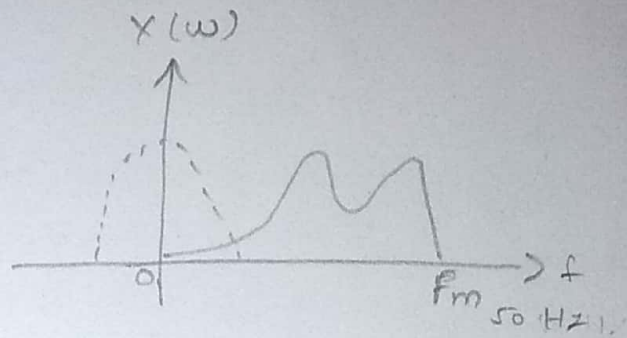
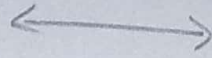
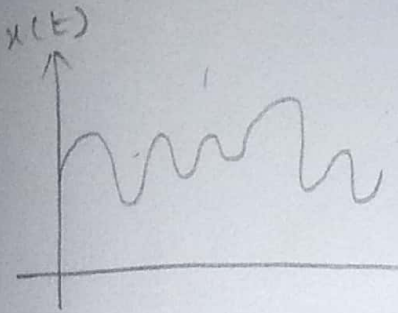
* The process of converting analog signal to discrete-time signal is called sampling

* The process of converting Discrete-time to Digital signal is called quantization.



Sampling Theorem :-

The sampling theorem states that any band limited signal $x(t)$ with maximum frequency f_m can be represented into its samples taken at the rate of $f_s \geq 2f_m$



$$f_s \geq 2f_m$$

$$\geq 2 \times 50$$

$$\geq 100$$

$$\frac{1}{T_s} = 0.01$$

$$T_s = 0.01$$