

3. Feed Back Amplifiers

Feed back principle and concept, types of feed back, feed back topologies, characteristics of -ve feed back amplifiers, classification of amplifiers, generalized analysis of feed back amplifiers, performance comparison of feed back amplifiers, Method of analysis of feed back amplifiers.

↑ The concept of Feed back (Need of Feed back in amplifiers)
The practical amplifier has a gain of nearly 1 million i.e., its output is 1 million times of the input and consequently even small disturbance is added at the ip of the amplifier, it will appear as amplified form at the op of the amplifier. There is a strong possibility in amplifiers to introduce noise due to sudden changes in temperature, electric and magnetic fields. Therefore, every high gain amplifier tends to give noise along with original signal at its op. This amplified noise at the op of an amplifier is undesirable and must be kept as small as possible. The noise level in amplifiers can be reduced by the use of -ve feed back.

* Feed back :-

The process of injecting a fraction (or) some of the op energy of an amplifier ^(either voltage/current) back to the ip is called as feed back.

An amplifier which uses the ^(principle) concept of feed back is called feed back amplifier.

Types of feed back :-

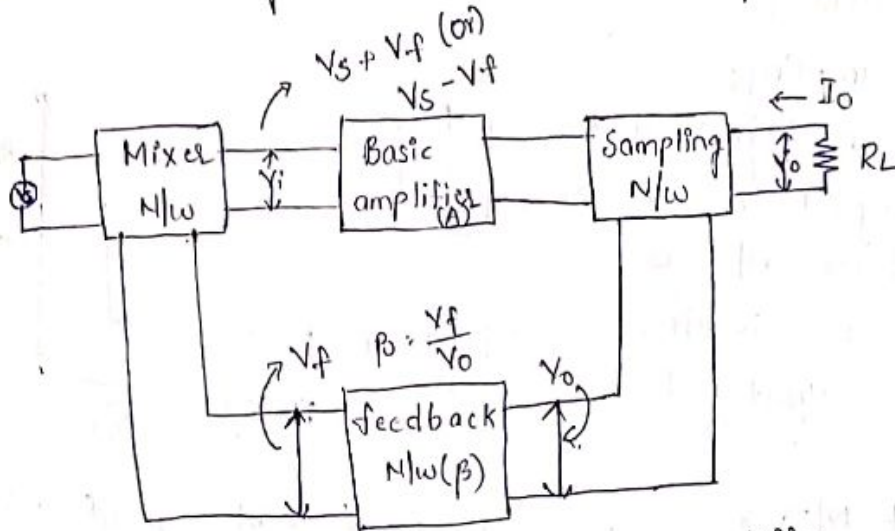
Depending upon either the feedback signal adds or) opposes ^{with} the ip signal these are 2 types of feed back techniques.

1) +ve feed back amplifier

2) -ve feed back "

If the feed back signal adds with the ip signal (when both are ^{same} in phase) is called +ve feed back

If the feed back signal opposes with the i/p signal (when both are in out of phase) is called -ve feed back
 * General Block diagram of feed back amplifier:



Block diagram of feed back amplifier

The above diagram shows the general block diagram for feed back amplifier. As shown in figure, the basic amplifier provides amplification. The sampling N/w is used to take the part of the o/p signal and it is back to the i/p through a feed back N/w. Again at the i/p of an amplifier the feed back signal is mixed with the original i/p signal by the mixed N/w and it provides summation signal ($V_s + V_f$) (or) difference signal ($V_s - V_f$)

Basic amplifier:

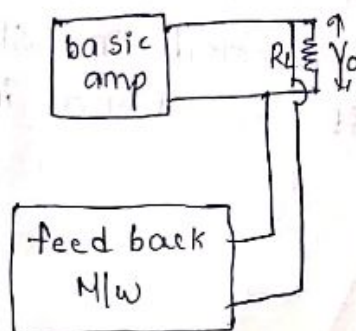
Generally, CE amplifier is used for providing amplification. and also it introduces 180° phase shift b/w i/p and o/p.

Sampling N/w:

There are 2 methods of sampling the o/p signal.

1. voltage sampling
2. Current sampling

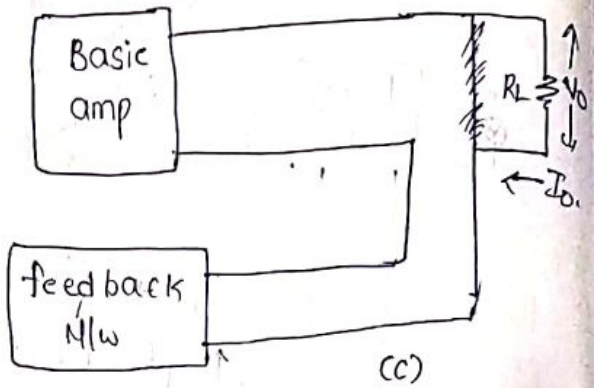
Voltage Sampling:



As shown in fig (b) if the sampling signal is voltage the feed back N/w is connected in shunt with the output terminals.

Current Sampling:

As shown in fig (c) if the sampling signal is current, the feed back N/w is connected in series with the output terminals.



Feed back N/w :

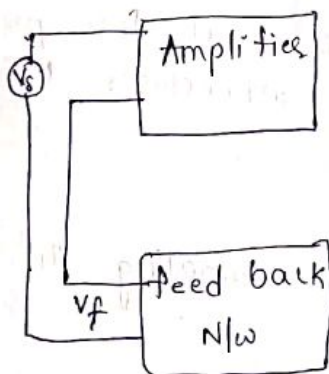
It is a two port N/w and it is made up of resistors, capacitors and inductors. But most advantages method, it is made up of with ^{only} resistors. It is designed to provide either 0° phase shift (or) 180° phase shift.

Mixer Network :

There are 2 methods for mixing the feed back signal with the i/p signal.

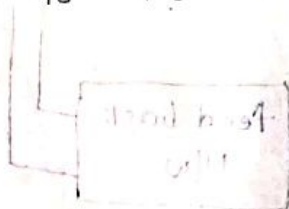
Series Mixing :

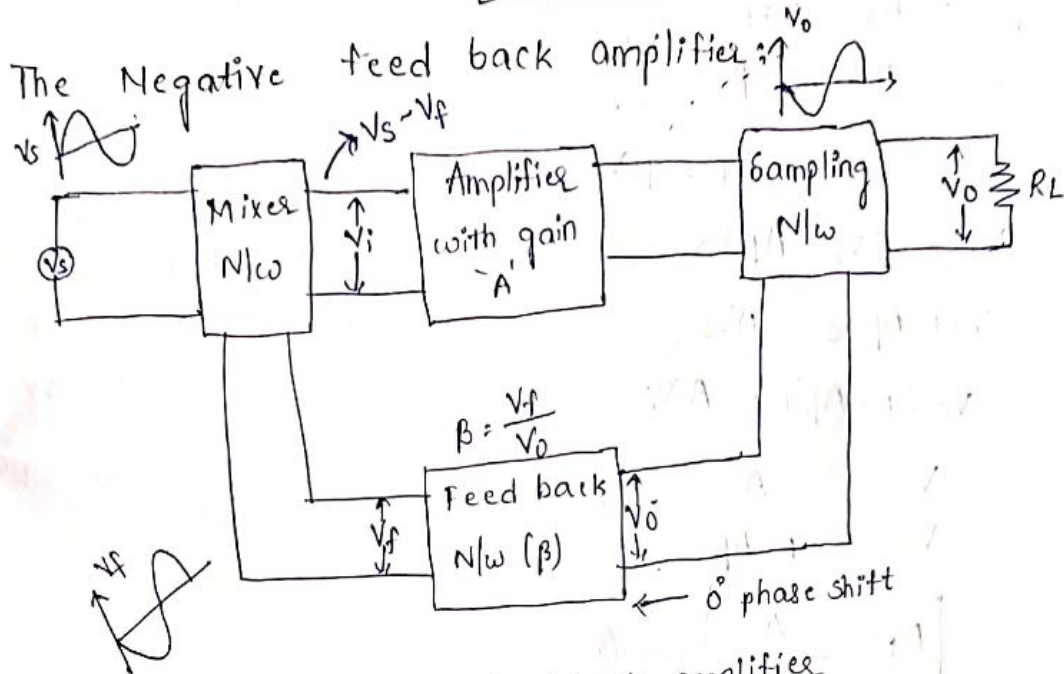
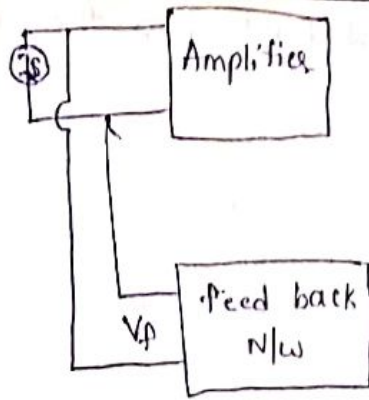
If the i/p signal is voltage source the feed back N/w is connected in series with the i/p terminals and it is shown in below fig.



Shunt Mixing :

If the i/p signal is current source the feed back N/w is connected in shunt with the i/p terminals and it is shown in the figure





B.D of -ve feed back amplifier

The above fig. shows the block diagram of -ve feed back amplifier. It consists of amplifier with gain A , sampling N/w , the feed back N/w and the mixer N/w .

The basic amplifier amplifies the weak i/p signal and it introduces 180° phase shift b/w o/p & i/p. The amplified o/p is sampled by using sampling N/w and it is fed back to the mixer N/w through a feedback N/w .

The feedback N/w is so designed to introduce 0° phase shift. Now, the feedback signal is out of phase with the i/p signal. The mixer N/w produces, the difference signal $V_s - V_f$ and it is properly given to the i/p of basic amplifier.

Expression for gain with -ve feed back (A_f)

The gain of an amplifier without feed back is given by $A = \frac{V_o}{V_s}$ — (1)

By introducing, -ve feed back into the circuit, V_s changes to $V_s - V_f$.

$$\therefore A = \frac{V_o}{V_s - V_f}$$

$$V_o = A(V_s - V_f)$$

$$V_o = AV_s - AV_f$$

we know that, $\beta = \frac{V_f}{V_o}$

$$V_f = \beta V_o$$

$$V_o = AV_s - A\beta V_o$$

$$V_o + A\beta V_o = AV_s$$

$$V_o(1 + A\beta) = AV_s$$

$$\frac{V_o}{V_s} = \frac{A}{1 + A\beta} \quad \text{--- (2)}$$

$$\boxed{A_f = \frac{A}{1 + A\beta}} \quad \text{--- (3)}$$

From eq (3), we can say that, without feedback the gain of an amplifier is 'A' and by providing -ve feed back, the gain of an amplifier 'A' is decreased by the factor of $1 + A\beta$.

Advantages:

1. Stability of Gain is increased.
2. Reduction in distortion
3. Reduction in Noise
4. i/p impedance is increased.
5. o/p impedance is decreased
6. Band width increases

→ The only disadvantage is the gain of the amplifier is decreased

Applications:

1. This concept used in almost all electronic amplifiers.
2. This concept used in regulated Power Supplies.
3. Used in amplifiers having large band width.

*The characteristics of negative feed back amplifier

1. The stabilization of gain is improved.
2. The frequency response (or) band width of an amplifier is increased.
3. The distortion and noise of an amplifier decreased.
4. input impedance is increased.
5. Output impedance is decreased.

Stabilization of gain is improved:-

As we know that, the gain of an amplifier with -ve feed back is given by $A_f = \frac{A}{1+A\beta}$ — (1)

The eq (1) is differentiated w.r.t to 'A'

$$\frac{dA_f}{dA} = \frac{(1+A\beta)(1) - A(\beta)}{(1+A\beta)^2}$$

$$\frac{dA_f}{dA} = \frac{1}{(1+A\beta)^2} \text{ — (2)}$$

The eq (2) is multiplied and divided with 'A'

$$\frac{dA_f}{dA} = \frac{A}{1+A\beta} \cdot \frac{1}{A(1+A\beta)}$$

$$\frac{dA_f}{dA} = \frac{A_f}{A(1+A\beta)}$$

$$\frac{dA_f}{dA} \cdot \frac{A_f}{A_f} = \frac{dA}{A} \cdot \frac{1}{(1+A\beta)} \text{ — (3)}$$

In eq (3), $\frac{dA_f}{A_f}$ is fractional change in voltage gain with -ve feedback.

$\frac{dA}{A}$ is fractional change in voltage gain without feedback.

Now from eq (3), we conclude that the fractional change in voltage gain with -ve feed back is given by the fractional change in gain without feedback is decreased by a factor of $(1+A\beta)$. Thus, the stability of gain is improved.

1. An amplifier has a open loop gain of 1000 with feed back ratio of 0.04. If the open loop gain changes by 10% due to temp. find the % change of gain of an amplifier with feedback.

Given, $A = 1000$

$\beta = 0.04$

$\frac{dA}{A} = 10\% = 0.1$

$\frac{dA_f}{A_f} = ?$

we know that,

$$\frac{dA_f}{A_f} = \frac{dA}{A} \cdot \frac{1}{(1+A\beta)}$$

$$= \frac{0.1}{1+1000(0.04)}$$

$$= 0.00243$$

$$\frac{dA_f}{A_f} = 0.24\%$$

2. An amplifier has a gain with feed back is 100. If the gain of an amplifier without feedback changes by 20% and with feedback changes by 2% then determine the values of open loop gain and feed back ratio.

Given, $A_f = 100$

$A = ?$

$\frac{dA}{A} = 20\%$

$\beta = ?$

$\frac{dA_f}{A_f} = 2\%$

we know that,

$$\frac{dA_f}{A_f} = \frac{dA}{A} \cdot \frac{1}{1+A\beta}$$

$$\Rightarrow 0.02 = \frac{0.2}{1+A\beta}$$

$$0.1 = \frac{1}{1+A\beta}$$

$$1+A\beta = 10$$

$$\omega_{KT} \quad A_f = \frac{A}{1+A\beta}$$

$$100 = \frac{A}{10}$$

$$A = 1000$$

$$A\beta = 9 \Rightarrow \beta = 1/10$$

$$\beta = 0.009$$

Frequency Response (or) Band width is increased.
The band width of an amplifier is the difference b/w upper cutoff frequency and lower cutoff frequency.
without feed back, it is given by $B.W = f_2 - f_1$ — (1)
By providing -ve feedback, it is given by

$$B.W_f = f_{2f} - f_{1f} \text{ — (2)}$$

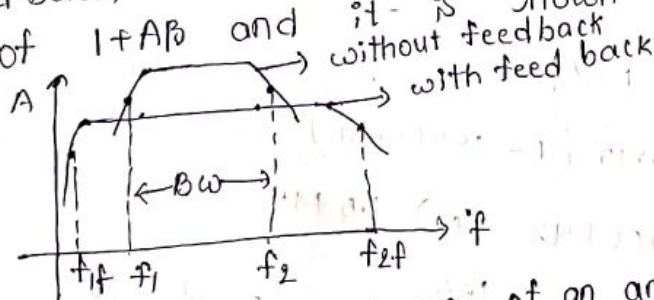
→ The gain and bandwidth remains the same and it is given by

$$B.W_f \times A_f = B.W \times A$$

$$B.W_f \times \frac{A}{1+A\beta} = B.W \times A$$

$$B.W_f = B.W (1+A\beta) \text{ — (3)}$$

From eq (3), we can say that without feed back the band width of an amplifier is B.W. By providing -ve feed back, the band width is increased by a factor of $1+A\beta$ and it is shown in below fig.



a) frequency response of an amplifier with and without feed back

Distortion and Noise of an amplifier is decreased without feed back, the distortion in amplifier is denoted as 'D' and with negative feed back the distortion 'D' is decreased by a factor of $(1+A\beta)$.

It is given by $D_f = \frac{D}{1+A\beta}$

Similarly,
$$M_f = \frac{M}{1 + A\beta}$$

Input Impedance increases:

Without feedback, the input impedance of an amplifier is R_i and by providing -ve feedback it is increased by a factor of $(1 + A\beta)$ and it is given by $R_{if} = R_i(1 + A\beta)$

Output Impedance decreases:

Without feedback, the output impedance of an amplifier is R_o and by providing -ve feedback it is decreased by a factor of $(1 + A\beta)$ and it is given by

$$R_{of} = \frac{R_o}{1 + A\beta}$$

An amplifier has a mid band gain without feedback is 125 and band width without feedback is 250KHz.

i) If 4% of -ve feedback is introduced in the ckt find the new band width and gain.

ii) If bandwidth is restricted to 1MHz, find out the feedback ratio and gain of the amplifier with feedback.

Given,

$$A = 125$$

$$BW = 250 \text{ KHz}$$

$$\beta = 4\% = 0.04$$

$$BW_f = ?$$

$$BW_f = BW(1 + A\beta)$$

$$= 250 \times 10^3 (1 + 125(0.04))$$

$$BW_f = 1500 \text{ KHz. (or) } 1.5 \text{ MHz}$$

$$A_f = \frac{A}{1 + A\beta} = \frac{125}{1 + 125(0.04)} = 20.833$$

ii) $BW_f = 1 \text{ MHz}$

$$\beta = ? ; A_f = ?$$

$$\text{WKT } BW_f = BW(1 + A\beta)$$

$$1 \times 10^6 = 250 \times 10^3 (1 + 125\beta)$$

$$\Rightarrow 125\beta = 3$$

$$\beta = \frac{3}{125} = 0.024$$

$$A_f = \frac{A}{1+A\beta} = \frac{125}{1+125(0.024)} = 31.25$$

Classification of -ve feed back amplifiers:-

Based on method of sampling the output signal and method of mixing the feed back signal with the input signal, there are 4 types of -ve feed back amplifiers.

1. Voltage - Series feed back amplifier
2. Voltage - shunt feed back amplifier
3. Current - Series feed back amplifier
4. Current - shunt feed back amplifier

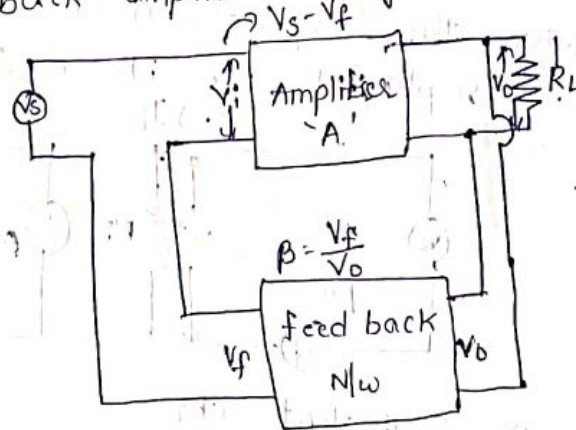
Each type of feed back amplifier is a combination of 2 words, first word refers to sampling and second word is for mixing.

Voltage Series Feed back amplifier (Voltage amplifier):

Fig 1a) shows the block diagram of voltage series feed back amplifier.

As shown in figure, for sampling the output voltage the feed back network is connected in shunt with the output terminals.

and the feed back signal is connected in series with the input terminals. It is also called shunt-series feed back amplifier.



Expression for gain with feed back (A_f)

The gain of an amplifier without feed back $A = \frac{V_o}{V_i}$

$$A = \frac{V_o}{V_i} \text{ (or) } \frac{V_o}{V_s} \quad \text{--- (1)}$$

By introducing -ve feed back into the circuit V_s changes to $V_s - V_f$.

$$A = \frac{V_o}{V_s - V_f}$$

$$\Rightarrow A(V_s - V_f) = V_o$$

$$AV_s - AV_f = V_o$$

wkt $\beta = \frac{V_f}{V_o}$

$$V_f = \beta V_o$$

$$AV_s - A\beta V_o = V_o$$

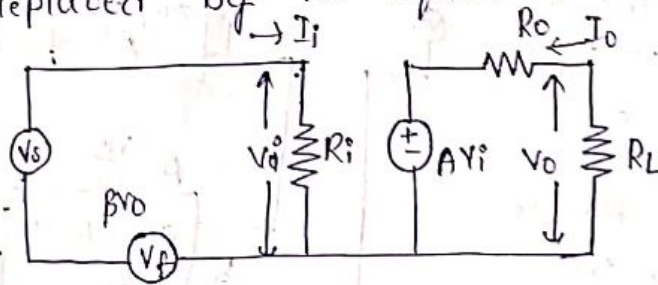
$$AV_s = V_o(1 + A\beta)$$

$$\frac{V_o}{V_s} = \frac{A}{1 + A\beta} \quad \text{--- (2)}$$

where $\frac{V_o}{V_s}$ is gain with -ve fb.

$$A_f = \frac{A}{1 + A\beta} \quad \text{--- (3)}$$

Input resistance with feedback (R_{if}) :-
To derive the expression for input resistance, fig(a) is replaced by its equivalent circuit



b) Equivalent ckt
Without feed back, input resistance of an amplifier

is $R_i = \frac{V_i}{I_i} \quad \text{--- (4)}$

By introducing -ve fb, V_i changes to $V_s - V_f$

$$R_i = \frac{V_s - V_f}{I_i}$$

$$= \frac{V_s - \beta V_o}{I_i}$$

$$R_i = \frac{V_s - \beta A Y_i}{I_i}$$

$$R_i I_i = V_s - \beta A Y_i$$

$$R_i I_i = V_s - \beta A I_i R_i$$

$$\therefore V_i = I_i R_i$$

$$R_i I_i + A \beta R_i I_i = V_s$$

$$R_i I_i (1 + A \beta) = V_s$$

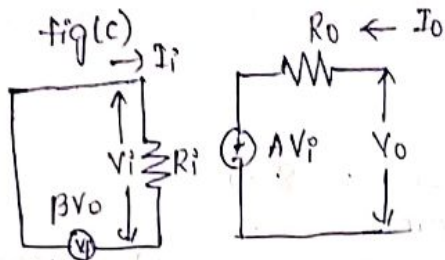
$$\boxed{\frac{V_s}{I_i} = R_i (1 + A \beta)} \quad (5)$$

$$\boxed{R_{if} = R_i (1 + A \beta)} \quad (6)$$

With series mixing, the input resistance is increased by a factor of $(1 + A \beta)$.

Output Resistance with feedback (R_{of}) :-

To measure the output resistance $V_s = 0$ & $R_L = \infty$.
 with these two assumptions fig (b) is modified as shown in fig (c)



Apply KVL to the output loop:

$$\begin{aligned} V_o &= I_o R_o + A V_i \\ &= I_o R_o + A (V_s - V_f) \\ &= I_o R_o + A (0 - V_f) \end{aligned}$$

$$V_o = I_o R_o - A V_f$$

$$V_o = I_o R_o - A \beta V_o$$

$$V_o + A \beta V_o = I_o R_o$$

$$V_o (1 + A \beta) = I_o R_o$$

$$\boxed{\frac{V_o}{I_o} = \frac{R_o}{1 + A \beta}} \quad (7)$$

$$\boxed{R_{of} = \frac{R_o}{1 + A \beta}} \quad (8)$$

With series mixing, the output resistance is decreased by a factor of $(1 + A \beta)$.

1) A Voltage Series -Ve feed back amplifier has a voltage gain without feed back is 500, input resistance is $3k\Omega$, output resistance is $20k\Omega$ and feed back ratio is 0.01. Calculate the voltage gain, input resistance and output resistance with feed back.

Given

$$A = 500$$

$$R_i = 3k\Omega = 3000\Omega$$

$$R_o = 20k\Omega = 20000\Omega$$

$$\beta = 0.01$$

$$A_f = ?$$

$$R_{if} = ?$$

$$R_{of} = ?$$

$$A_f = \frac{A}{1+A\beta} = \frac{500}{1+500(0.01)} = 83.3$$

$$R_{if} = R_i(1+A\beta) = 3000(1+500(0.01)) = 18000\Omega$$

$$R_{of} = \frac{R_o}{1+A\beta} = \frac{20000}{6} = 3.3 \times 10^3\Omega$$

2) An amplifier has a voltage gain is 4000, i/p resistance is $2k\Omega$ and o/p resistance is $16k\Omega$. Calculate the voltage gain, i/p resistance and o/p resistance of the circuit when 5% of feedback is fed in the form of series.

$$A = 4000$$

$$R_i = 2k\Omega$$

$$R_o = 16k\Omega$$

$$\beta = 5\% = \frac{5}{100} = 0.05$$

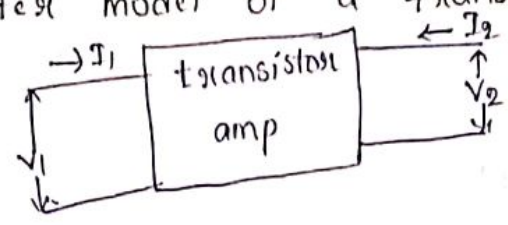
$$A_f = \frac{A}{1+A\beta} = \frac{4000}{1+4000(0.05)} = 19.9$$

$$R_{if} = R_i(1+A\beta) = 2 \times 10^3(1+4000(0.05)) = 402k\Omega$$

$$R_{of} = \frac{R_o}{1+A\beta} = \frac{16 \times 10^3}{1+200} = 0.07k\Omega$$

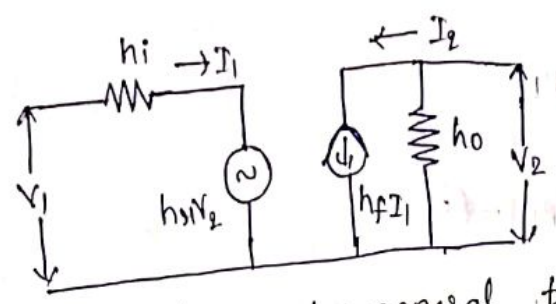
2) A practical ckt diagram of voltage series feedback amplifier

h-parameter model of a transistor amplifier :-

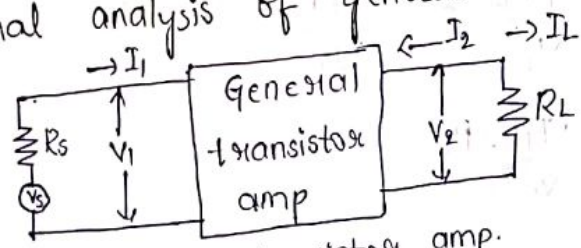


$$V_1 = h_{11}I_1 + h_{12}V_2$$

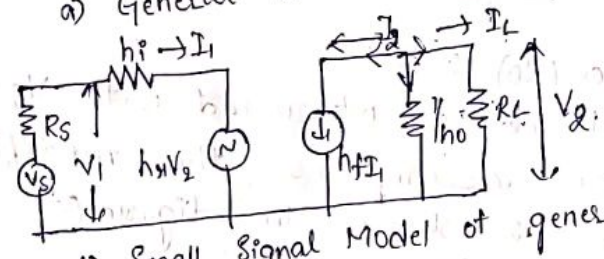
$$I_2 = h_{21}I_1 + h_{22}V_2$$



Small Signal analysis of general transistor amplifier :-



a) General transistor amp.



b) Small Signal Model of general transistor amp.

i) $A_I = \frac{I_L}{I_1} = \frac{-I_2}{I_1}$ — (1)

$$I_2 = h_f I_1 + h_o V_2$$

$$= h_f I_1 + h_o I_L R_L$$

$$I_2 = h_f I_1 + h_o (-I_2) R_L$$

$$I_2 + h_o I_2 R_L = h_f I_1$$

$$I_2 (1 + h_o R_L) = h_f I_1$$

$$\frac{I_2}{I_1} = \frac{h_f}{1 + h_o R_L}$$

$$A_I = \frac{-h_f}{1 + h_o R_L}$$
 — (2)



$V_2 = I_2 R_L$
 $I_2 = h_f I_1 + h_o V_2$
 $I_2 = h_f I_1 + h_o I_2 R_L$
 $I_2 - h_o I_2 R_L = h_f I_1$
 $I_2 (1 - h_o R_L) = h_f I_1$
 $I_2 = \frac{h_f I_1}{1 - h_o R_L}$

$$\begin{aligned}
 2) R_I &= \frac{V_1}{I_1} \\
 &= \frac{h_{i1}I_1 + h_{o1}V_2}{I_1} \\
 &= \frac{h_{i1}I_1 + h_{o1}(-I_2 R_L)}{I_1} \\
 &= h_{i1} - \frac{h_{o1}I_2 R_L}{I_1}
 \end{aligned}$$

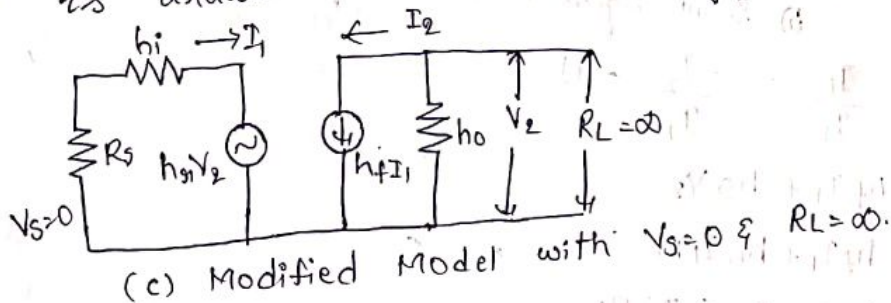
$$R_I = h_{i1} + h_{o1} A_I R_L \quad \text{--- (3)}$$

$$3) A_V = \frac{V_2}{V_1} = \frac{I_2 R_L}{V_1} = \frac{-I_2 R_L}{V_1}$$

$$A_V = \frac{-I_2}{I_1} \cdot \frac{I_1}{V_1} \cdot R_L$$

$$A_V = \frac{A_I R_L}{R_{i1}} \quad \text{--- (4)}$$

4) Output Resistance (R_o):
 The output admittance is measured with $V_s = 0$ and $R_L = \infty$, with these assumptions the modified model is drawn as shown in figure (c)



$$\begin{aligned}
 Y_o &= \frac{I_2}{V_2} \\
 &= \frac{h_f I_1 + h_o V_2}{V_2}
 \end{aligned}$$

$$Y_o = h_f \frac{I_1}{V_2} + h_o \quad \text{--- (5)}$$

The relation b/w I_1 & V_2 is obtained by applying KVL to the i/p node.

$$(R_s + h_{i1}) I_1 + h_{o1} V_2 = 0$$

$$(R_s + h_{i1}) I_1 = -h_{o1} V_2$$

$$\frac{I_1}{V_2} = \frac{-h_{o1}}{R_s + h_{i1}} \quad \text{--- (6)}$$

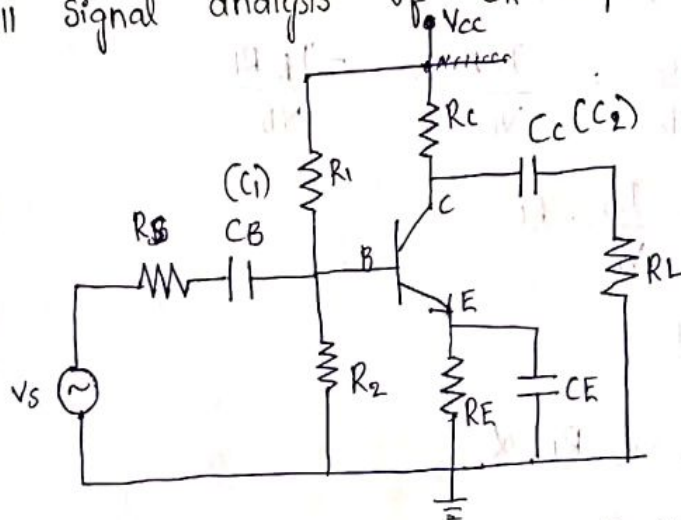
Sub. eq 6 in eq (5)

$$y_o = h_f \left(\frac{-h_{se}}{R_s + h_i} \right) + h_o$$

$$y_o = \frac{-h_f h_{se}}{R_s + h_i} + h_o \quad \text{--- (7)}$$

$$R_o = \frac{1}{y_o} = \frac{1}{h_o - \frac{h_f h_{se}}{R_s + h_i}} \quad \text{--- (8)}$$

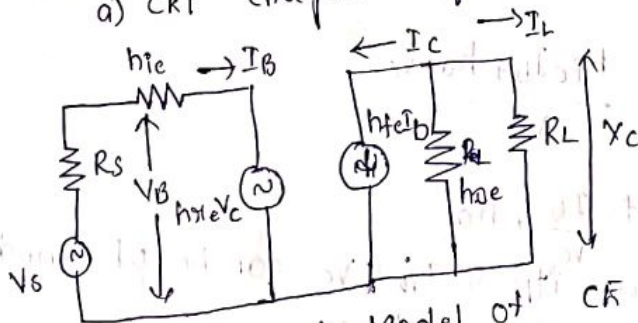
Small signal analysis of CE amplifier (Exact analysis) :-



C_1, C_2 - Coupling capacitors

C_E - Bypass capacitor

a) ckt - diagram of CE amplifier



b) Small signal Model of CE amp

$$1) A_I = \frac{I_L}{I_i} = \frac{-I_c}{I_b}$$

$$I_c = h_{fe} I_b + h_{oe} V_c$$

$$I_c = h_{fe} I_b + h_{oe} I_L R_L$$

$$I_c = h_{fe} I_b + h_{oe} (-I_c) R_L$$

$$I_c + h_{oe} I_c R_L = h_{fe} I_b$$

$$(1 + h_{oe} R_L) I_c = h_{fe} I_b$$

$$\frac{I_c}{I_b} = \frac{h_{fe}}{1 + h_{oe} R_L}$$

$$A_I = \frac{-I_c}{I_b} = \frac{-h_{fe}}{1 + h_{oe} R_L}$$

$$2) R_i = \frac{V_i}{I_i} = \frac{V_b}{I_b}$$

$$= \frac{h_{ie} I_b + h_{ie} V_c}{I_b}$$

$$= \frac{h_{ie} I_b + h_{ie} (-I_c R_L)}{I_b}$$

$$= h_{ie} - h_{ie} R_L \frac{I_c}{I_b}$$

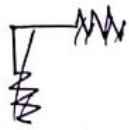
$$R_i = h_{ie} + h_{ie} A_I R_L$$

$$3) A_v = \frac{V_2}{V_1} = \frac{V_c}{V_b} = \frac{I_c R_L}{V_b} = \frac{-I_c R_L}{V_b}$$

$$A_v = \frac{-I_c}{I_b} \cdot \frac{I_b}{V_b} \cdot R_L$$

$$A_v = \frac{A_I R_L}{R_i}$$

$$4) R_o = \frac{1}{Y_o}, \quad V_s = 0, \quad R_L = \infty$$



$$Y_o = \frac{I_c}{V_c} = \frac{h_{fe} I_b + h_{oe} V_c}{V_c}$$

$$Y_o = h_{fe} \frac{I_b}{V_c} + h_{oe}$$

The relation b/w I_b & V_c can be obtained by applying KVL at input loop.

$$(R_s + h_{ie}) I_b + h_{ie} V_c = 0$$

$$(R_s + h_{ie}) I_b = -h_{ie} V_c$$

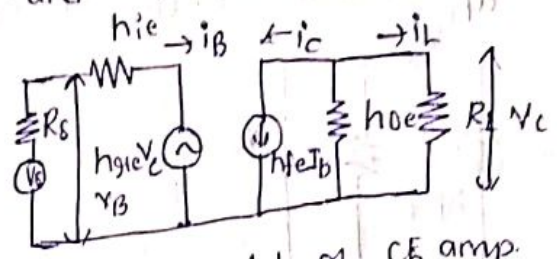
$$\frac{I_b}{V_c} = \frac{-h_{ie}}{R_s + h_{ie}}$$

$$Y_o = h_{fe} \left(\frac{-h_{ie}}{R_s + h_{ie}} \right) + h_{oe}$$

$$Y_o = \frac{-h_{fe} h_{ie}}{R_s + h_{ie}} + h_{oe}$$

$$R_o = \frac{1}{Y_o} = \frac{1}{h_{oe} - \frac{h_{fe} h_{ie}}{R_s + h_{ie}}}$$

Approximate analysis of CE amplifier :-
 To draw the approximate model, first we consider exact model and it is shown in fig (a)



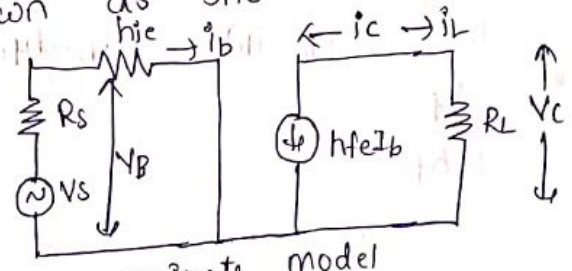
a) Exact model of CE amp.

Step 1: As $\frac{1}{h_{oe}} \gg R_L$, h_{oe} is neglected from the op circuit

$$i_c = h_{fe} I_B$$

Step 2: $h_{ie} V_C = h_{ie} (i_L R_L) = h_{ie} (-i_C R_L) = -h_{ie} (h_{fe} I_B) R_L$

where the product of $h_{ie} h_{fe} \approx 0.01$ and then voltage source $h_{ie} V_C$ is neglected from input equivalent circuit with these two conditions, the remaining circuit is drawn as shown in fig (b).



b) approximate model

Analysis :-

1. Current gain $(A_I) = \frac{I_2}{I_1} = \frac{i_L}{i_B} = \frac{-i_C}{i_B} = \frac{-h_{fe} i_B}{i_B} = -h_{fe}$

$$A_I = -h_{fe}$$

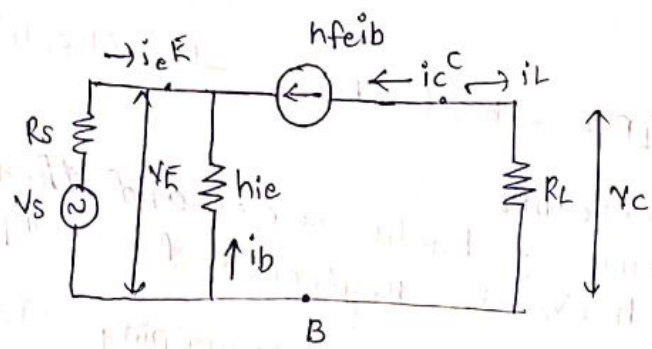
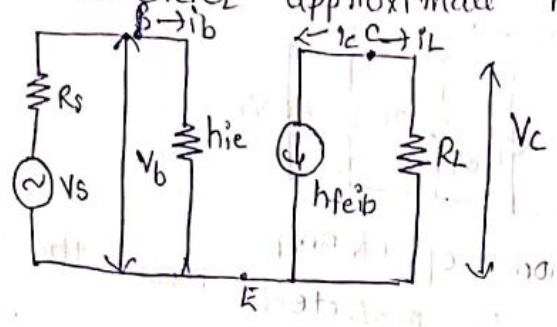
2. $R_I = \frac{V_1}{I_1} = \frac{V_B}{I_B} = \frac{h_{ie} i_B}{i_B} = h_{ie}$

3. $A_V = \frac{V_2}{V_1} = \frac{V_C}{V_B} = \frac{i_L R_L}{V_B} = \frac{-i_C R_L}{V_B} = \frac{-h_{fe} I_B R_L}{h_{ie} i_B}$

$$A_V = \frac{-h_{fe} R_L}{h_{ie}}$$

4. $R_O = \infty$

Approximate analysis of CB amplifier:
 To draw the approximate model of CB amplifier, first we consider approximate model of CE amplifier.



a) app. model of CB amp.

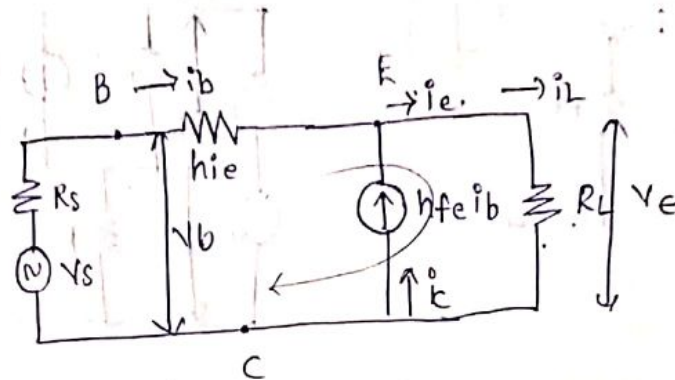
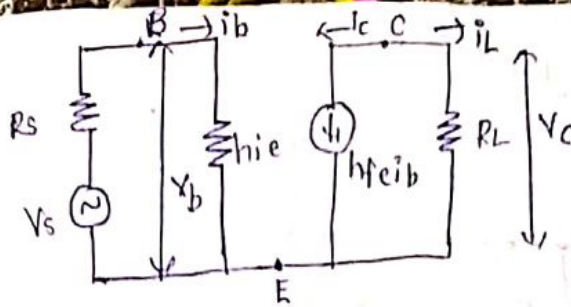
$$\begin{aligned}
 1) \quad A_I &= \frac{i_L}{i_e} = hfe \cdot i_b \cdot \frac{-i_c}{i_e} = \frac{hfe \cdot i_b}{-(i_c + i_b)} = \frac{hfe \cdot i_b}{i_b + hfe \cdot i_b} \\
 &= \frac{hfe \cdot i_b}{i_b(1 + hfe)} = \frac{hfe}{1 + hfe} \\
 A_I &= \frac{hfe}{1 + hfe}
 \end{aligned}$$

$$2) \quad R_I = \frac{V_E}{I_E} = \frac{hfe \cdot i_b}{-(i_b + i_c)} = \frac{hfe \cdot i_b}{i_b + hfe \cdot i_b} = \frac{hfe}{1 + hfe}$$

$$3) \quad A_V = \frac{V_c}{V_E} = \frac{i_L R_L}{V_E} = \frac{hfe \cdot i_b R_L}{hfe \cdot i_b} = \frac{hfe R_L}{hfe}$$

$$4) \quad R_o = \infty$$

Approximate analysis of CE amplifier:
 To draw the approximate analysis of CE amplifier, first we obtain approximate model of CE amplifier.



$$1) A_I = \frac{i_L}{i_b} = \frac{h_{fe} i_b}{i_b} = h_{fe}$$

$$= \frac{i_e}{i_b} = \frac{i_b + i_c}{i_b} = \frac{i_b + h_{fe} i_b}{i_b} = 1 + h_{fe}$$

$$2) R_I = \frac{V_b}{I_b} = \frac{h_{ie} i_b + h_{fe} i_b R_L}{i_b} = h_{ie} + h_{fe} R_L$$

$$= \frac{h_{ie} i_b + (i_b + h_{fe} i_b) R_L}{i_b} = h_{ie} + (1 + h_{fe}) R_L$$

$$3) A_v = \frac{V_e}{V_b} = \frac{(1 + h_{fe}) i_b R_L}{h_{ie} i_b + (1 + h_{fe}) i_b R_L} = \frac{A_I R_L}{R_I}$$

$$A_v = \frac{(1 + h_{fe}) R_L}{h_{ie} + (1 + h_{fe}) R_L}$$

$$4) R_o = \frac{V_e}{I_e} = i_e$$

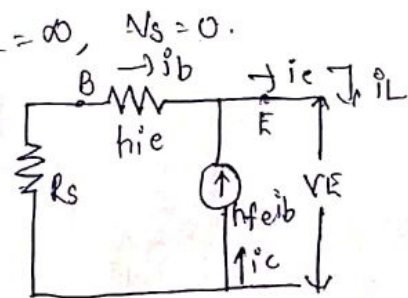
R_o :-

remove the

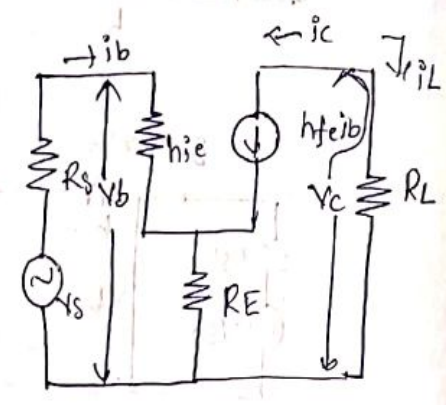
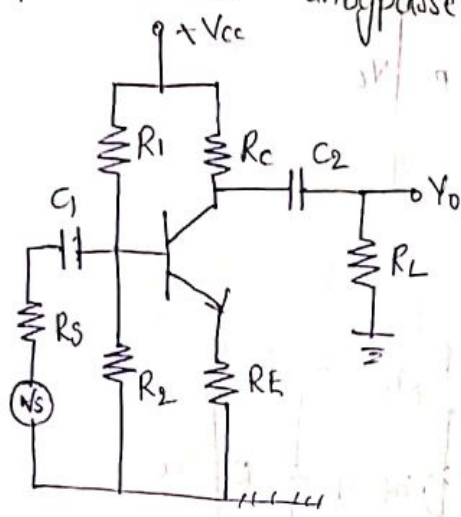
$$R_o = \frac{V_e}{I_e} = \frac{h_{ie} i_b + R_s i_b}{i_b + i_c}$$

$$= \frac{(h_{ie} + R_s) i_b}{i_b + h_{fe} i_b}$$

$$R_o = \frac{h_{ie} + R_s}{1 + h_{fe}}$$



CE amplifier with unbypassed RE :-

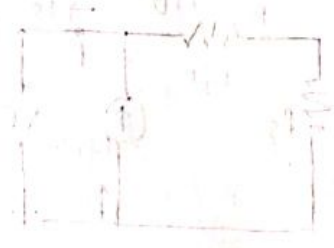


i) $A_I = \frac{I_c}{I_b} = \frac{-I_c}{I_b} = \frac{-h_{fe} i_b}{i_b} = -h_{fe}$

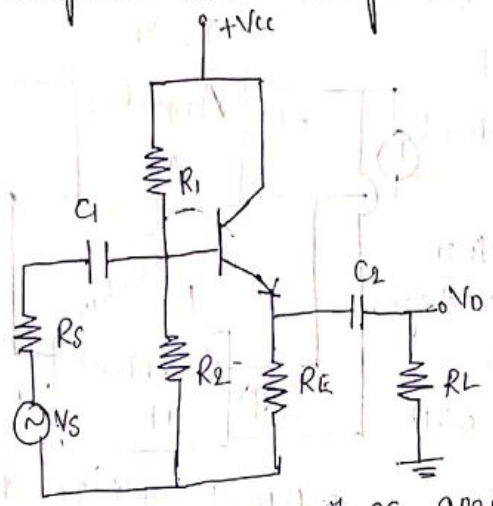
ii) $R_I = \frac{V_B}{I_b} = \frac{h_{ie} i_b + R_E (i_b + h_{fe} i_b)}{i_b}$
 $= h_{ie} + (1 + h_{fe}) R_E$

iii) $A_v = \frac{V_c}{V_b} = \frac{A_I R_L}{R_I} = \frac{-h_{fe} (h_{ie} + (1 + h_{fe}) R_E)}{h_{ie} + (1 + h_{fe}) R_E}$
 $= \frac{-h_{fe} R_L}{h_{ie} + (1 + h_{fe}) R_E}$

iv) $R_L = \infty, V_s = 0$.
 Output resistance, $R_o = \infty$.



practical ckt diagram for voltage series feed back amplifier.

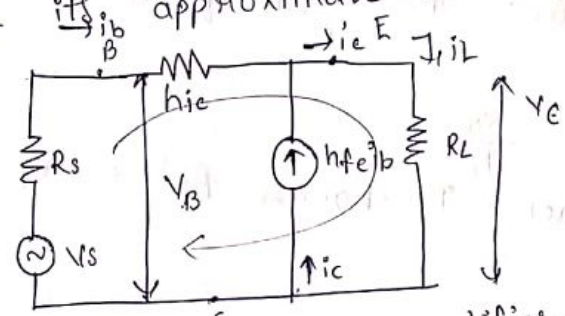


a) ckt diagram of CE amplifier

Fig (a) shows the ckt diagram of CE amplifier. The voltage which is developed across resistor RE back to the i/p in series opposition. Thus, it is the best example of voltage series feed back amplifier.

Analysis:-

To derive the characteristics circuit diagram is replaced by its approximate model.



(b) app. model of CE amplifier.

$$i) A_I = \frac{I_L}{I_B} = \frac{i_e}{i_b} = \frac{i_b + i_c}{i_b} = \frac{i_b + h_{fe} i_b}{i_b} = 1 + h_{fe}$$

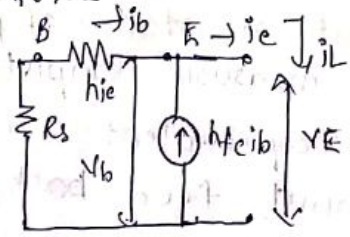
$$ii) R_I = \frac{V_b}{i_b} = \frac{i_b h_{ie} + i_c R_L}{i_b} = \frac{i_b h_{ie} + (1 + h_{fe}) i_b R_L}{i_b}$$

$$= \frac{h_{ie} + (1 + h_{fe}) R_L}{1}$$

$$iii) A_v = \frac{V_E}{V_b} = \frac{A_I R_I}{R_I} = \frac{(1 + h_{fe}) R_L}{h_{ie} + (1 + h_{fe}) R_L}$$

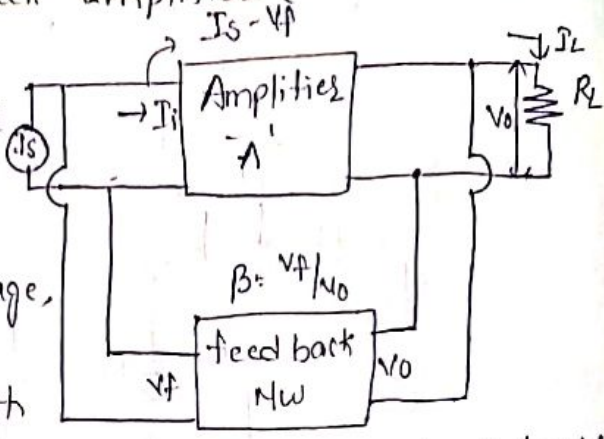
$$iv) R_o = \frac{V_E}{(1 + h_{fe}) i_b} = \frac{h_{ie} i_b + R_s i_b}{(1 + h_{fe}) i_b}$$

$$= \frac{h_{ie} + R_s}{1 + h_{fe}}$$



Voltage shunt feed back amplifier (Trans resistance amplifier)

Fig (a) shows the block diagram of voltage shunt feed back amplifier.



As shown in fig (a), for sampling the output voltage, the feed back Mw is connected in shunt with the o/p terminals.

Fig (a) B-D of voltage shunt feed back amplifier.

~~This feed back~~

This sampled signal back to the i/p in shunt opposition through a feed back network. The shunt sampling decreases its output resistance and the shunt mixing decreases its input resistance by a factor of $(1+A\beta)$.

Trans resistance gain with feed back (A_f)? without feed back the gain of an amplifier is given by

$$A = \frac{V_o}{I_s} = \frac{V_o}{I_i}$$

By -ve feed back, I_i changes to $I_s - V_f$

$$A = \frac{V_o}{I_s - V_f}$$

$$A I_s - A V_f = V_o$$

$$A I_s - A \beta V_o = V_o$$

$$A I_s = (1 + A \beta) V_o$$

$$\frac{V_o}{I_s} = \frac{A}{1 + A \beta}$$

$$A_f = \frac{A}{1 + A \beta}$$

Input resistance (R_{if}):

To measure input resistance, fig (a) is replaced by its equivalent circuit without feed back, input resistance is given as

$$R_i = \frac{V_i}{I_i}$$

with feed back,

$$I_i \leftarrow I_s - V_f$$

$$R_i \cdot \frac{V_i}{I_s - V_f}$$

$$R_i I_s - R_i V_f = V_i$$

$$R_i I_s - R_i \beta V_o = V_i$$

$$R_i I_s - R_i \beta A I_i = V_i$$

$$R_i I_s - R_i \beta A \left(\frac{V_i}{R_i} \right) = V_i$$

$$R_i I_s - A \beta V_i = V_i$$

$$R_i I_s = (1 + A \beta) V_i$$

$$\frac{V_i}{I_s} = \frac{R_i}{1 + A \beta}$$

$$R_{i-f} = \frac{R_i}{1 + A \beta}$$

Output Resistance (R_{o-f}):-

To measure the o/p resistance, $I_s = 0$ and $R_L = \infty$.

Apply KVL to o/p loop

$$V_o = R_o I_o + A I_i$$

$$V_o = R_o I_o + A (I_s - V_f)$$

$$V_o = R_o I_o + A (-V_f)$$

$$V_o + A V_f = R_o I_o$$

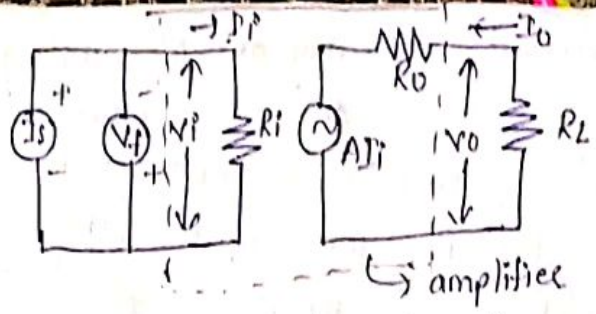
$$V_o + A \beta V_o = R_o I_o$$

$$(1 + A \beta) V_o = R_o I_o$$

$$\frac{V_o}{R_o} = \frac{I_o}{1 + A \beta}$$

$$\frac{V_o}{I_o} = \frac{R_o}{1 + A \beta}$$

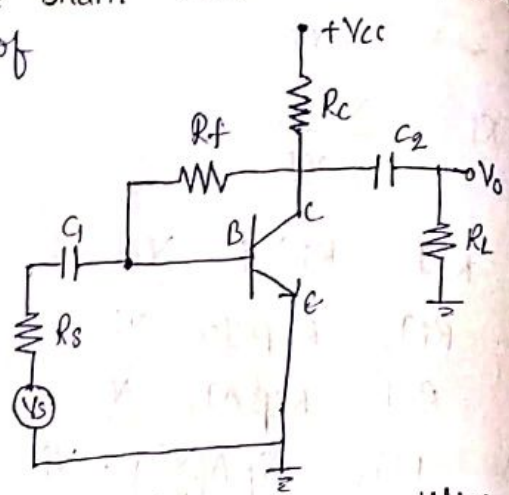
$$R_{o-f} = \frac{R_o}{1 + A \beta}$$



b) Equivalent circuit

Practical ckt diagram of voltage shunt feed back amplifier

Fig (a) shows the ckt diagram of CE amplifier with feed back resistor R_f . The output voltage which is developed across resistor R_L , back to the input through a feed back resistor R_f in shunt opposition.

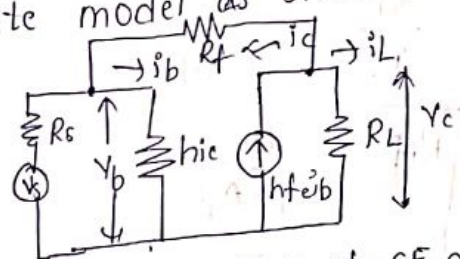


Thus, it is the best example for voltage shunt feedback amplifier.

Analysis:-

To determine the characteristics, the amplifier is replaced by its approximate model when an amplifier with feed back resistor R_f , it is difficult to do analysis.

a) ckt diagram of CE amplifier with feed back resistor R_f . The circuit diagram is shown in fig (b).



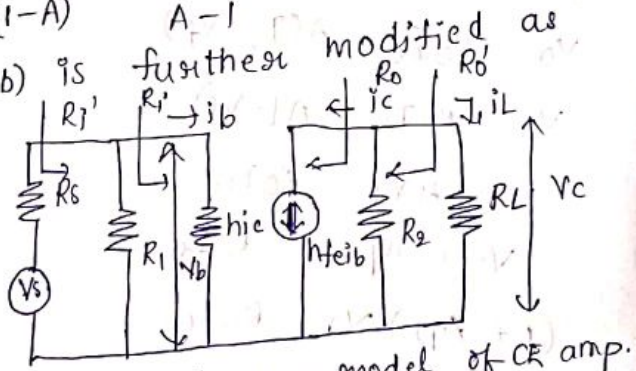
b) app. model of CE amplifier with resistor R_f .

Thus, by using Miller's theorem R_f can be separated into two resistances R_1 and R_2 .

From Miller's theorem,

$$R_1 = \frac{R_f}{1-A} \quad , \quad R_2 = \frac{R_f \cdot A}{-(1-A)} = \frac{R_f \cdot A}{A-1}$$

with these two values, fig (b) shown in fig (c).



c) modified app. model of CE amp. with R_f .

$$1) A_I = \frac{i_L}{i_b} = \frac{-i_c}{i_b} = \frac{-h_f e i_b}{i_b}$$

$$A_I = -h_f e$$

$$2) R_I = \frac{V_b}{i_b} = \frac{h_i e i_b}{i_b} = h_i e$$

$$R_i' = R_i \parallel R_1$$

$$3) A_V = \frac{A_I R_L}{R_I} = \frac{-h_f e R_L}{h_i e}$$

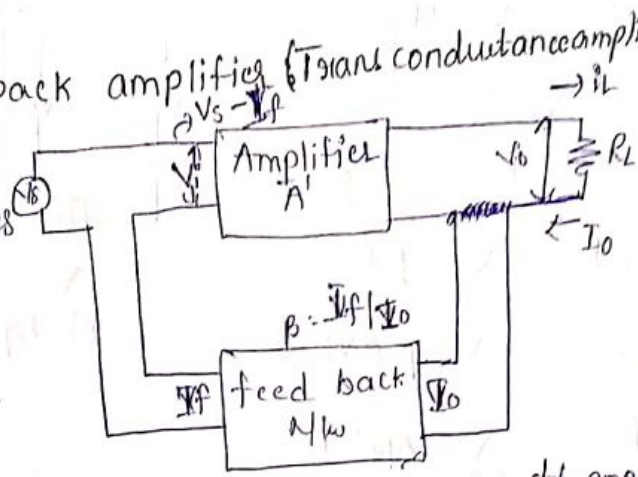
4) $R_o = \infty$

5) $R_o' = R_o || R_2 = R_2$

Consistent series feed back amplifier (Trans conductance amplifier)

Fig (a) shows the block diagram of consistent series feed back amplifier.

As shown in fig (a), for sampling the output current, the feedback N/w is connected in series with the output terminals. This feedback signal back to the input in series opposition.



a) B.C of current series fb amp.

By series sampling, the output resistance increases by a factor of $(1 + A\beta)$ and by series mixing, the input resistance increases by a factor of $(1 + A\beta)$.

Trans conductance gain with feed back (A_f):

without feed back, V_i changed to $V_s - I_f$

$$A = \frac{I_o}{V_s} = \frac{I_o}{V_i}$$

with feed back, V_i changed to $V_s - I_f$

$$A = \frac{I_o}{V_s - I_f}$$

$$\beta = \frac{I_f}{I_o}$$

$$AV_s - AI_f = I_o$$

$$AV_s - A\beta I_o = I_o$$

$$AV_s = (1 + A\beta) I_o$$

$$\frac{V_s}{I_o} = A \frac{I_o}{V_s} = \frac{A}{1 + A\beta}$$

$$A_f = \frac{A}{1 + A\beta}$$

Input resistance (R_{if}):

To measure the input resistance, fig (a) is replaced by its equivalent ckt without feed back, the input resistance is given by

$$R_i = \frac{V_f}{I_f}$$

with feed back,

$$V_i \leftrightarrow V_s - I_i R_s$$

$$R_i = \frac{V_s - I_i R_s}{I_i}$$

$$R_i I_i = V_s - \beta I_o$$

$$R_i I_i = V_s - \beta A V_i$$

$$R_i I_i = V_s - \beta A I_i R_i$$

$$I_i R_i (1 + \beta A) = V_s$$

$$\frac{V_s}{I_i} = R_i (1 + \beta A)$$

$$R_{if} = R_i (1 + \beta A)$$

Output resistance (R_{of}):

To measure the o/p resistance, $R_L = \infty$, $V_s = 0$.

Apply KCL to o/p loop.

$$I_o = A V_i + \frac{V_o}{R_o}$$

$$I_o = A (V_s - I_i) + \frac{V_o}{R_o}$$

$$I_o = -A I_i + \frac{V_o}{R_o}$$

$$I_o = -\beta I_o + \frac{V_o}{R_o}$$

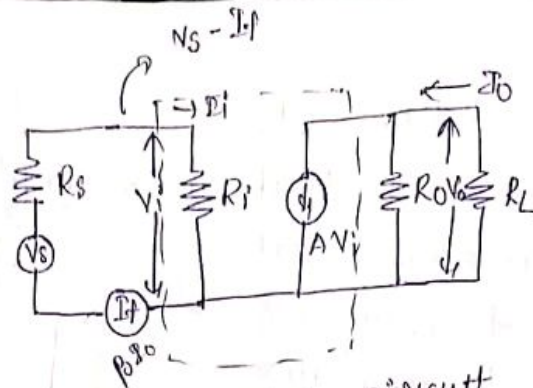
$$I_o (1 + \beta A) = \frac{V_o}{R_o}$$

$$\frac{V_o}{I_o} = R_o (1 + \beta A)$$

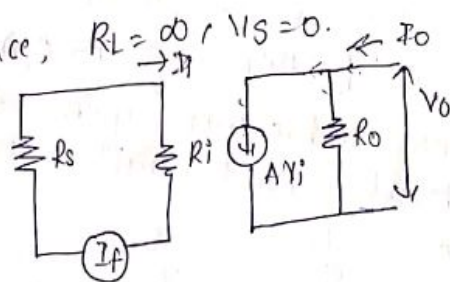
$$R_{of} = R_o (1 + \beta A)$$

Practical ckt diagram for current series feed back amplifier:

Fig(a) shows the ckt diagram of CE amplifier with unbypassed R_E . The current flowing through R_E back to the i/p in series opposition. Thus, it is the best example for current series feed back amplifier.



b) Equivalent circuit



Analysis:

To determine the characteristics, ckt diagram is replaced by its approximate model in fig (b).

$$A_v = \frac{i_L}{i_b} = \frac{-i_c}{i_b} = \frac{-h_{fe} i_b}{i_b}$$

$$A_v = -h_{fe}$$

$$R_i = \frac{V_b}{i_b}$$

$$= \frac{h_{ie} i_b + R_e (i_b + i_c)}{i_b}$$

$$= \frac{h_{ie} i_b + R_e (i_b + h_{fe} i_b)}{i_b}$$

$$R_i = h_{ie} + R_e (1 + h_{fe})$$

$$A_v = \frac{A_i R_L}{R_i} = \frac{-h_{fe} R_L}{-h_{ie} + R_e (1 + h_{fe})}$$

R_o $V_s = 0, R_L = \infty$.

$R_o = \infty$.

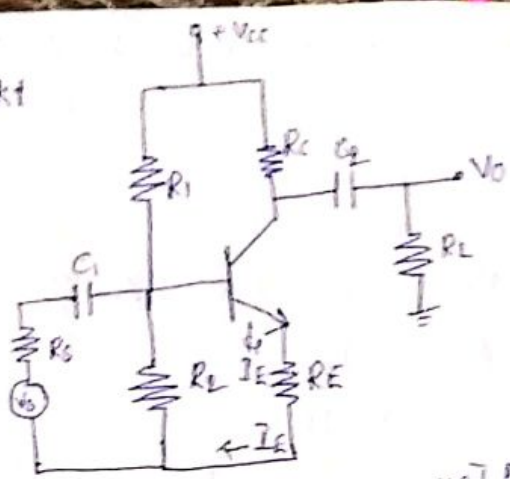
Current shunt feedback amplifiers (current amplification):

Fig (a) shows the block diagram of current shunt feedback amplifier.

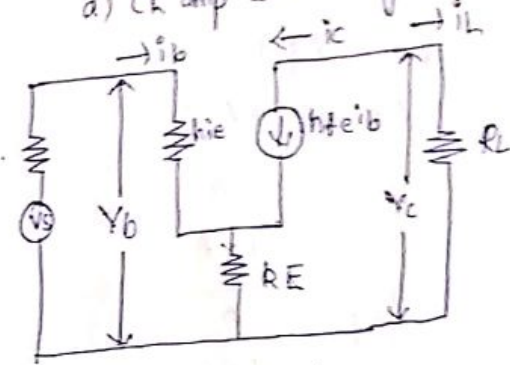
As shown in fig (a), for sampling the o/p current the feedback N/w is connected in series with the o/p terminals.

This feedback signal is connected in shunt in the input terminals.

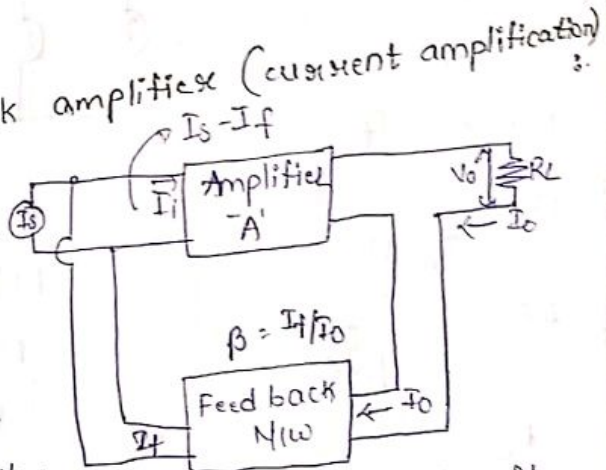
By series mixing, the output resistance increases and by shunt mixing, the input resistance decreases.



a) CE amp with unbypassed RE.



b) app. model.



a) B.d of current shunt fb

Gain with feed back (A) :-

The gain of an amplifier without feedback is given by

$$A = \frac{I_o}{I_i}$$

with feed back, $I_i \leftrightarrow I_s - I_f$

$$A = \frac{I_o}{I_s - I_f}$$

$$\beta = I_f / I_o$$

$$A I_s - A I_f = I_o$$

$$A I_s - A \beta I_o = I_o$$

$$A I_s = (1 + A \beta) I_o$$

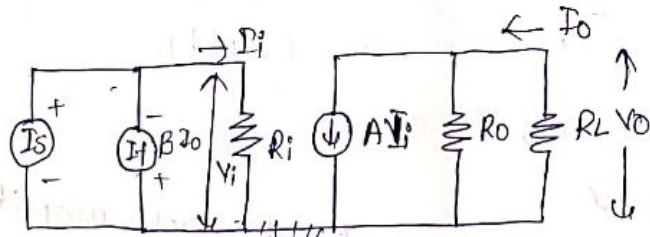
$$\frac{I_o}{I_s} = \frac{A}{1 + A \beta}$$

$$A_f = \frac{A}{1 + A \beta}$$

Input resistance (R_{if}) :-

To measure the i/p resistance, fig (a) is replaced by its equivalent ckt.

$$R_i = \frac{V_i}{I_i}$$



$$I_i \leftrightarrow I_s - I_f$$

$$R_i = \frac{V_i}{I_s - I_f}$$

$$R_i = \frac{V_i}{I_s - \beta I_o}$$

$$R_i = \frac{V_i}{I_s - \beta A I_i}$$

$$\Rightarrow R_i I_s - R_i I_f = V_i$$

$$R_i I_s - R_i \beta I_o = V_i$$

$$R_i I_s - R_i \beta A I_i = V_i$$

$$R_i I_s - R_i A \beta \left(\frac{V_i}{R_i} \right) = V_i$$

$$R_i I_s = (1 + A \beta) V_i$$

$$R_i I_s - R_i A \beta I_i = V_i$$

$$R_i I_s - R_i A \beta I_i = R_i I_i$$

$$R_i I_s = (1 + A \beta) R_i$$

$$\frac{V_i}{I_s} = \frac{R_i}{1 + A \beta}$$

$$R_{if} = \frac{R_i}{1 + A \beta}$$

Output Resistance (R_{of}) :-

To measure the o/p resistance, $V_s = 0, R_L = \infty$.

Apply KCL to the o/p loop.

$$A I_i = I_o + \frac{V_o}{R_L}$$

$$A I_i = I_o + \frac{V_o}{R_o}$$

$$I_o = A(I_s - I_i) + \frac{V_o}{R_o}$$

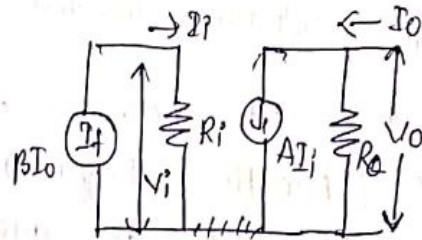
$$I_o = -A I_i + \frac{V_o}{R_o}$$

$$I_o = -A \beta I_o + \frac{V_o}{R_o}$$

$$(1 + A \beta) I_o = \frac{V_o}{R_o}$$

$$\frac{V_o}{I_o} = R_o (1 + A \beta)$$

$$R_{of} = R_o (1 + A \beta)$$



Performance comparison of -ve feed back amplifiers.

Parameter	Type of feed back amplifier			
	Voltage-Series	Voltage-Shunt	Current-Series	Current-Shunt
1) Gain	Decreases	Decreases	Decreases	Decreases
2) Band width	Increases	Increases	Increases	Increases
3) distortion & noise	decreases	decreases	decreases	decreases
4) input resistance	increases	decreases	increases	decreases
5) Output resistance	decreases	decreases	increases	increases

- 1) what is effect of -ve feed back on amplifier gain?
- 2) Derive the expression for o/p resistance of a voltage-sampled circuit
- 3) Analyse CE amplifier with resistor R_E using feed back concept?

- 4) S.T the bandwidth improved with -ve feedback?
- 5) S.T i/p resistance increases with series mixing? (Voltage series)
- 6) Draw the ckt diagram of current series feedback amplifier? (Practical diagram), R_{if} , R_{of} , A_f .
- 7) Explain the steps in analysis of -ve feedback amplifier? (Feedback topology)
- 8) Explain the concept of feedback with block diagram and also advantages and disadvantages of -ve fb.
- 9) What is fb and its types?
- 10) with a neat sketch obtain -ve fb amplifier and derive expression for closed loop gain?
- 11) How does +ve fb reduce distortion of an amplifier?
- 12) Compare -ve fb and +ve fb amplifiers.
- 13) Difference b/w amplifier and oscillator?
- 14) An amplifier with gain 200 is provided with -ve feedback of fb ratio 0.005, find new gain?

Given, $A = 200$

$$\beta = 0.005$$

$$A_f = \frac{A}{1 + A\beta} = \frac{200}{1 + 200(0.005)} = 100$$

- 15) An amplifier with $2.5k\Omega$ i/p resistance and $50k\Omega$ o/p resistance has a voltage gain of 100. The amplifier is now modified to provide 5% of -ve feedback. In series with the i/p. Calculate the voltage gain, i/p resistance, and the o/p resistance with fb.

Given, $R_i = 2.5k\Omega$

$$R_o = 50k\Omega$$

$$A = 100$$

$$\beta = 5\% = 0.05$$

$$A_f = \frac{A}{1 + A\beta} = \frac{100}{1 + 100(0.05)} = 16.66$$

$$R_{if} = R_i(1 + A\beta) = 2.5 \times 10^3 (1 + 100(0.05)) = 15k\Omega$$

$$R_{of} = R_o(1 + A\beta) = 50 \times 10^3 (6) = 300 \times 10^3 = 300k\Omega$$

16) Feed back factor $\beta = 0.1$ is introduced into an amplifier with a gain of 20 and band width of 0.6 MHz, obtain the resultant band width of the fb amplifier.

Given,

$$\beta = 0.1$$

$$A = 20$$

$$BW = 0.6 \times 10^6 \text{ Hz}$$

$$BW_f = BW (1 + A\beta)$$

$$= 0.6 \times 10^6 (1 + 20(0.1))$$

$$BW_f = 1.8 \text{ MHz.}$$

17) An amplifier has a gain of 50 with -ve fb. For a specified o/p voltage, if the i/p required is 0.1V without fb and 0.8V with fb, compute β and open loop gain?

Given,

$$A_f = 50$$

$$\beta = ?$$

$$A = ?$$

$$V_i = 0.1 \text{ V}$$

$$V_{if} = 0.8 \text{ V}$$

$$A = \frac{V_o}{V_i} = \frac{V_o}{0.1} \quad \text{--- (1)}$$

$$A_f = \frac{V_o}{V_{if}} = \frac{V_o}{0.8} \quad \text{--- (2)}$$

$$V_o = A(0.1) \quad V_o = A_f(0.8)$$

$$A(0.1) = A_f(0.8)$$

$$A(0.1) = 50 \times 0.8$$

$$A = \frac{50 \times 0.8}{0.1} = 400$$

$$\text{wkt } A_f = \frac{A}{1 + A\beta}$$

$$50 = \frac{400}{1 + 400\beta}$$

$$1 + 400\beta = 8$$

$$1 + 400\beta = 8$$

$$400\beta = 7$$

$$\beta = 0.0175$$

Draw the topologies of $-ve$ feed back

feed back connection - topology

Feed back topology:-

The type of feed back connection in $-ve$ feed back amplifiers is called feed back topology:

There are 4 basic ways of connecting the feedback signal

Both voltage and current feedback ^{can be} ~~at the~~ ^{into the} input either in series or parallel.

There are 4 types of feedback topologies.

Voltage - series ; Current - series
voltage - shunt ; Current - shunt

Explain the method of identifying the feedback topology?

Identify topology:-

1) To find the type of sampling network.

i) By shortening the output terminals, if feedback signal becomes zero then it is called voltage sampling.

ii) By opening the output loop, if feedback signal becomes zero then it is called current sampling.

2) To find the type of mixing network.

i) If the feedback signal is subtracted from the externally applied as a voltage in the input loop, it is called series mixing.

ii) If the feedback signal is subtracted from the externally applied signal as a current in the input loop, it is called shunt mixing.

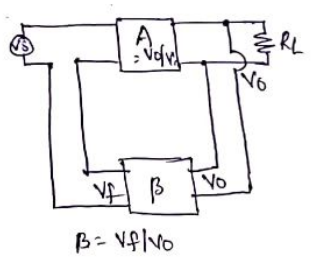
Thus by finding the type of sampling network and mixing network, type of feedback amplifiers can be determined.

For example: If an amplifier uses a voltage sampling and series mixing, then it is called a voltage-series feedback amplifier.

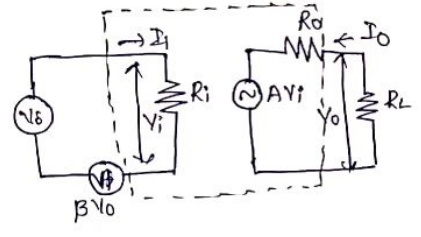
Type of topology

Voltage - series feedback amplifier
(Voltage amplifiers)

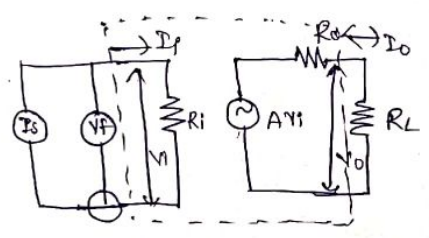
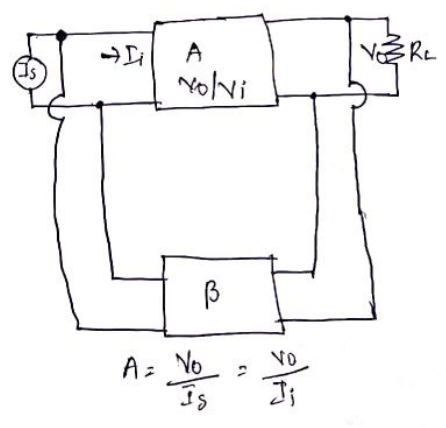
Block diagram



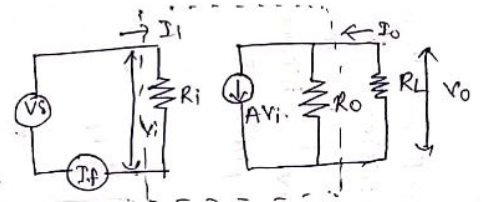
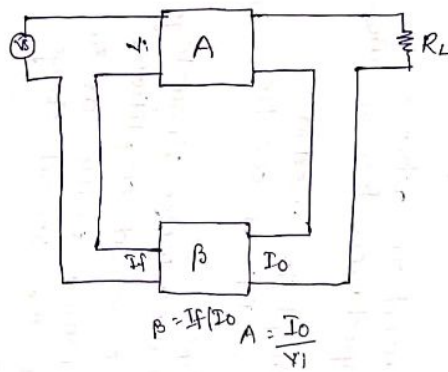
Equivalent Circuit



Voltage - shunt feedback amplifier
(Trans resistance amplifiers)



Current-series
 feedback amplifier
 (Trans conductance
 amplifier)



Current-shunt
 feedback amplifier
 (Current amplifier)

