

## UNIT – V

### TWO-PORT NETWORKS

#### Introduction

A pair of terminals at which an electrical signal may enter or leave a **network** is called a **port**. The terminals or port is required for connecting input excitation to the **network**. It is also required for connecting some other networks such as load. The terminals are most useful for making measurements. In general, the minimum number of terminals required is two.

A **network** having only one pair of terminals or one port is called **one port network**. The Fig. 7.1 (a) shows a one port **network**.

A **network** consisting two pairs of terminals is called **two port network** as shown in the Fig. 7.1 (b). The terminals are generally named as 1-1' and 2-2'. In general, a port designated 1-1', is connected to the driving energy source while the other port designated 2-2' is connected to the load. A port at which energy source is connected is called **driving point** of the **network** or **input port**. A port at which load is connected is called **output port**.

Fig. 7.1 (c) represents a **network** with n-port called **n-port network**. In such networks, generally one port is connected to energy source, one port is connected to load and other ports may be connected to the different networks.

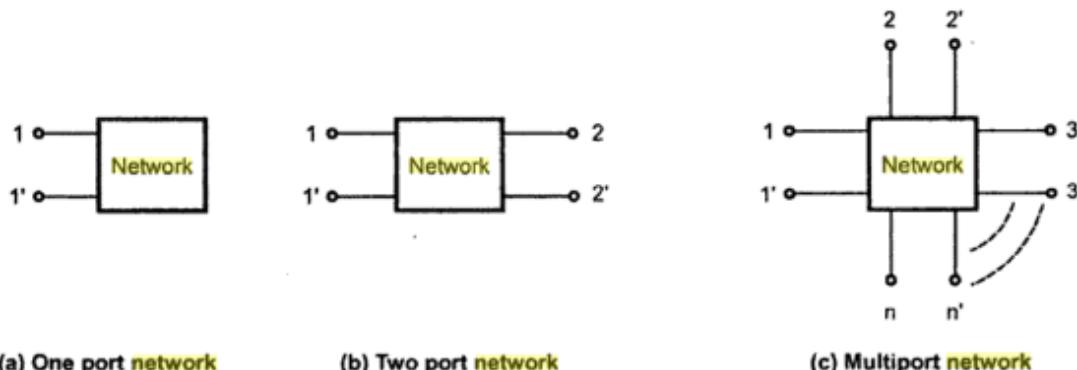


Fig. 7.1

#### Two Port Network Parameters

Consider a two port **network** as shown in the Fig. 7.2. In all there are four variables; two voltages and two currents. In general, any two port **network** has one pair of voltage and current at each port.

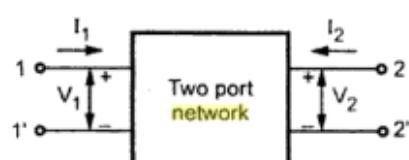


Fig. 7.2 Two port **network**

Assuming the variables at input and output ports as transformed quantities, the voltage and current at input terminals are  $V_1$  and  $I_1$  while at output terminals are  $V_2$  and  $I_2$ , as shown in the Fig. 7.2. The directions of both the currents  $I_1$  and  $I_2$  are assumed to be flowing into the **network**. Such currents entering the **network** are assumed to be positive. Also the voltage have the positive reference polarities. Also the voltages have the positive reference polarities.

Before starting the discussion on the two port **network** parameters, let us make certain assumptions necessary in the **analysis**. The assumptions are as follows.

1. The voltages and currents in the **network** present inside a box are not available for the measurements.
2. The **network** should consists only **linear elements** along with **dependent sources** if any. But **independent** or **active sources** should not be present in the **network** inside box.

3. If the **network** inside box consists energy storing elements such as inductor and capacitor, then **initial conditions** of such elements should be zero.

In order to describe the relationship between port voltages and currents, one requires the linear equations equal to the number of ports. So in two port **network analysis**, we will require two linear equations interms of four above mentioned variables. We can obtain these equations by considering two variables as dependent variables while other as independent variables. As the **network** consists only linear elements, the linear relationship can be obtained by writing two variables interms of other two variables. There are six possible ways of selecting two independent variables out of four variables. Thus there are six different pairs of equations defining their own sets of parameters such as impedance ( $z$ ), admittance ( $y$ ), hybrid ( $h$ ), inverse hybrid ( $g$ ), transmission and inverse transmission parameters. In next sections, we will discuss  $z$ -parameters,  $y$ -parameters, hybrid parameters and transmission parameters in detail.

### **z-Parameters**

These are also called **impedance parameters**. These are obtained by expressing voltages at two ports in terms of currents at two ports. Thus, **currents  $I_1$  and  $I_2$  are independent variables**; while  **$V_1$  and  $V_2$  are dependent variables**.

$$V_1 = f_1(I_1, I_2)$$

$$V_2 = f_2(I_1, I_2)$$

In equation form, above relations can be written as,

$$V_1 = z_{11} I_1 + z_{12} I_2 \quad \dots (1)$$

$$V_2 = z_{21} I_1 + z_{22} I_2 \quad \dots (2)$$

In matrix form, above equations can be written as,

$$\begin{bmatrix} V_1 \\ V_2 \end{bmatrix} = \begin{bmatrix} z_{11} & z_{12} \\ z_{21} & z_{22} \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \end{bmatrix}$$

or  $[V] = [z] [I]$

The individual  $z$ -parameters can be obtained by assigning values of independent variables to be zero. The  $z$ -parameters can be defined as follows.

[A] Let  $I_2 = 0$ ; port - 2 is open circuited.

From equation (1),

$$z_{11} = \left. \frac{V_1}{I_1} \right|_{I_2=0} \Omega$$

The parameter  $z_{11}$  is called **open circuit driving point input impedance**.

From equation (2),

$$z_{21} = \left. \frac{V_2}{I_1} \right|_{I_2=0} \Omega$$

The parameters  $z_{21}$  is called **open circuit forward transfer impedance**.

[B] Let  $I_1 = 0$ ; i.e. port - 1 is open circuited.

Form equation (1),

$$z_{12} = \left. \frac{V_1}{I_2} \right|_{I_1=0} \Omega$$

The parameter  $z_{12}$  is called **open circuit reverse transfer impedance**.

From equation (2),

$$z_{22} = \left. \frac{V_2}{I_2} \right|_{I_1=0} \Omega$$

The parameter  $z_{22}$  is called **open circuit driving point output impedance**.

these parameters are defined only when the current in one of the ports is zero. This corresponds to the conditions that the one of the ports is open circuited. Hence z-parameters are named as **open circuit impedance parameters**.

⇒ **Example 7.1 :** Find the z-parameters for the network shown in following Fig. 7.4.

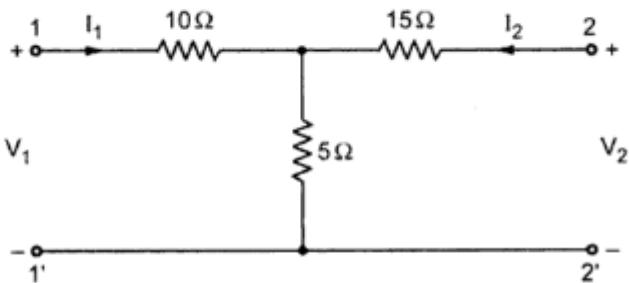
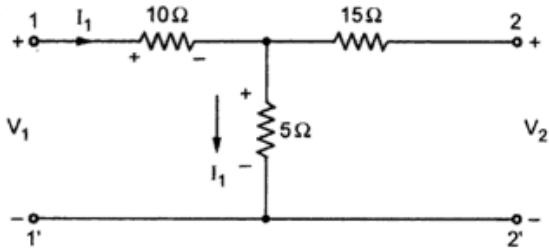


Fig. 7.4

**Solution :** By definition z-parameters are given as,

$$V_1 = z_{11} I_1 + z_{12} I_2$$

$$V_2 = z_{21} I_1 + z_{22} I_2$$



(A) Let  $I_2 = 0$ , i.e. port 2 is open circuited as shown in the Fig. 7.4 (a).

As port 2 is open circuited, the current flowing through 5 Ω is also  $I_1$ . Note that no current will flow through 15 Ω as it is connected to open terminals.

Fig. 7.4 (a)

Applying KVL at input side, we get,

$$-10 I_1 - 15 I_1 + V_1 = 0$$

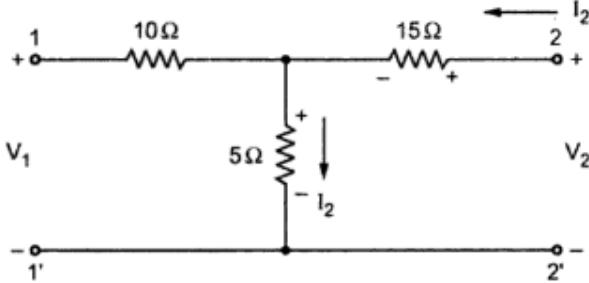
$$\therefore V_1 = 15 I_1$$

$$z_{11} = \left. \frac{V_1}{I_1} \right|_{I_2=0} = 15 \Omega$$

From the Fig. 7.4 (a), we can write,

$$V_2 = 5 I_1$$

$$z_{21} = \left. \frac{V_2}{I_1} \right|_{I_2=0} = 5 \Omega$$



**Fig. 7.4 (b)**

(B) Let  $I_1 = 0$ , i.e. port 1 is open circuited as shown in the Fig. 7.4 (b).

As port 1 is open circuited, the current flowing through  $5 \Omega$  is also  $I_2$ . Note that no current will flow through  $10 \Omega$  as it is connected to open terminals.

Applying KVL at output side, we get,

$$-15 I_2 - 5 I_2 + V_2 = 0$$

$$\therefore V_2 = 20 I_2$$

$$\begin{aligned} \therefore z_{22} &= \left. \frac{V_2}{I_2} \right|_{I_1=0} \\ &= 20 \Omega \end{aligned}$$

From the Fig. 7.4 (b), we can write,

$$V_1 = 5 I_2$$

$$\therefore z_{12} = \left. \frac{V_1}{I_2} \right|_{I_1=0} = 5 \Omega$$

Hence z-parameters of the given network are

$$[z] = \begin{bmatrix} z_{11} & z_{12} \\ z_{21} & z_{22} \end{bmatrix} = \begin{bmatrix} 15 & 5 \\ 5 & 20 \end{bmatrix} \Omega$$

### y-Parameters

These are also called **admittance parameters**. These are obtained by expressing currents at two ports in terms of voltages at two ports. Thus, voltages  $V_1$  and  $V_2$  are independent variables, while  $I_1$  and  $I_2$  are dependent variables. Thus, we have

$$I_1 = f_1(V_1, V_2)$$

$$I_2 = f_2(V_1, V_2)$$

In equation form, above relations can be written as,

$$I_1 = y_{11} V_1 + y_{12} V_2 \quad \dots (1)$$

$$I_2 = y_{21} V_1 + y_{22} V_2 \quad \dots (2)$$

In matrix form, above equations can be written as,

$$\begin{bmatrix} I_1 \\ I_2 \end{bmatrix} = \begin{bmatrix} y_{11} & y_{12} \\ y_{21} & y_{22} \end{bmatrix} \begin{bmatrix} V_1 \\ V_2 \end{bmatrix}$$

$$\text{or} \quad [I] = [y] [V]$$

The individual y-parameters are defined as follows,

[A] Let  $V_2 = 0$  i.e. port-2 is short circuited.

From equation (1)

$$y_{11} = \left. \frac{I_1}{V_1} \right|_{V_2=0}$$

The parameter  $y_{11}$  is called **short circuit driving point input admittance**.

From equation (2)

$$y_{21} = \left. \frac{I_2}{V_1} \right|_{V_2=0} \text{ } \Omega$$

The parameter  $y_{21}$  is called **short circuit forward transfer admittance**.

[B] Let  $V_1 = 0$  i.e. **port-1 is short circuited**.

From equation (1),

$$y_{12} = \left. \frac{I_1}{V_2} \right|_{V_1=0} \text{ } \Omega$$

The parameter  $y_{12}$  is called **short circuit reverse transfer admittance**.

From equation (2),

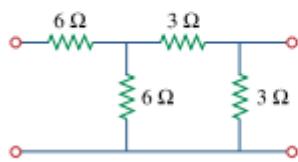
$$y_{22} = \left. \frac{I_2}{V_2} \right|_{V_1=0} \text{ } \Omega$$

The parameter  $y_{22}$  is called **short circuit driving point output admittance**.

These parameters are defined individually only when the voltage in any one of the ports is zero. This corresponds to the condition that one of the ports is short circuited. Hence  $y$ -parameters are also called **short circuit admittance parameters**.

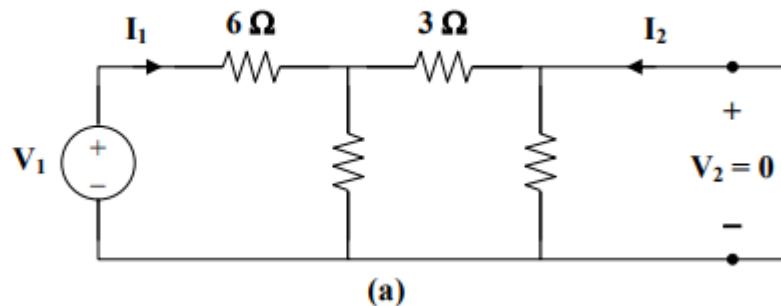
### Problem :

Calculate the  $y$  parameters for the two-port in Fig.



**Figure**

To get  $y_{11}$  and  $y_{21}$ , consider the circuit in Fig.(a).



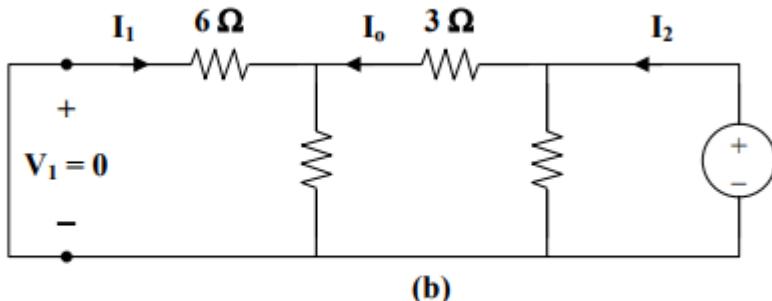
$$V_1 = (6 + 6 \parallel 3) I_1 = 8 I_1$$

$$y_{11} = \frac{I_1}{V_1} = \frac{1}{8}$$

$$I_2 = \frac{-6}{6+3} I_1 = \frac{-2}{3} \frac{V_1}{8} = \frac{-V_1}{12}$$

$$y_{21} = \frac{I_2}{V_1} = \frac{-1}{12}$$

To get  $y_{22}$  and  $y_{12}$ , consider the circuit in Fig.(b).



$$y_{22} = \frac{I_2}{V_2} = \frac{1}{3 \parallel (3 + 6 \parallel 6)} = \frac{1}{3 \parallel 6} = \frac{1}{2}$$

$$I_1 = \frac{-I_0}{2}, \quad I_0 = \frac{3}{3+6} I_2 = \frac{1}{3} I_2$$

$$I_1 = \frac{-I_2}{6} = \left(\frac{-1}{6}\right)\left(\frac{1}{2}V_2\right) = \frac{-V_2}{12}$$

$$y_{12} = \frac{I_1}{V_2} = \frac{-1}{12} = y_{21}$$

Thus,

$$[y] = \begin{bmatrix} 1 & -1 \\ \frac{8}{12} & \frac{1}{12} \\ -\frac{1}{12} & \frac{1}{2} \end{bmatrix} S$$

### **h-Parameters**

These are also called hybrid parameters. These parameters are very useful in constructing models for transistors. The transistor parameters cannot be calculated using by either short circuit admittance parameter or open circuit impedance parameter measurement. These parameters are obtained by expressing voltage at input port and the current at output port in terms of the current at the input port and the voltage at the output port. Thus, the current  $I_1$  and voltage  $V_2$  are independent variables; while current  $I_2$  and voltage  $V_1$  are the dependent variables. Thus, we have,

$$V_1 = f_1(I_1, V_2)$$

$$I_2 = f_2(I_1, V_2)$$

In equation form, above relations can be written as,

$$V_1 = h_{11} I_1 + h_{12} V_2 \quad \dots (1)$$

$$I_2 = h_{21} I_1 + h_{22} V_2 \quad \dots (2)$$

In matrix form, the above equations can be written as,

$$\begin{bmatrix} V_1 \\ I_2 \end{bmatrix} = \begin{bmatrix} h_{11} & h_{12} \\ h_{21} & h_{22} \end{bmatrix} \begin{bmatrix} I_1 \\ V_2 \end{bmatrix}$$

The individual h-parameters can be defined as follows

[A] Let  $V_2 = 0$ ; i.e. port - 2 is short circuited.

From equation (1),

$$h_{11} = \left. \frac{V_1}{I_1} \right|_{V_2=0} \Omega$$

The parameter  $h_{11}$  is called short circuit input impedance.

From equation (2),

$$h_{21} = \left. \frac{I_2}{I_1} \right|_{V_2=0}$$

The parameter  $h_{21}$  is called **short circuit forward current gain**. It is unitless.

[B] Let  $I_1 = 0$ ; i.e. **port - 1 is circuited**.

From equation (1),

$$h_{12} = \left. \frac{V_1}{V_2} \right|_{I_1=0}$$

The parameter  $h_{12}$  is called **open circuit reverse voltage gain**. It is also unitless.

From equations (2),

$$h_{22} = \left. \frac{I_2}{V_2} \right|_{I_1=0} \Omega$$

The parameter  $h_{22}$  is called **open circuit output admittance**.

All above parameters are having different units such as ohm for short circuit impedance and mho for open circuit output admittance, the name of the parameter is **hybrid parameter**.

### Problems-

Determine the h-parameter with the following data:

i. with the output terminals short circuited,  $V_1 = 25 \text{ V}$ ,  $I_1 = 1 \text{ A}$ ,  $I_2 = 2$

A

ii. with the input terminals open circuited,  $V_1 = 10 \text{ V}$ ,  $V_2 = 50 \text{ V}$ ,  $I_2 = 2$

A

### Solution

The h-parameter equations are,

$$V_1 = h_{11}I_1 + h_{12}V_2$$

$$I_2 = h_{21}I_1 + h_{22}V_2$$

a. With output short-circuited,  $V_2 = 0$ , given:  $V_1 = 25 \text{ V}$ ,  $I_1 = 1 \text{ A}$  and  $I_2 = 2 \text{ A}$ .

$$\begin{aligned} \therefore & \quad 25 = h_{11} \times 1 \\ \text{and} & \quad 2 = h_{21} \times 1 \end{aligned} \Rightarrow h_{11} = 25 \Omega, \text{ and } h_{21} = 2$$

b. With input open-circuited,  $I_1 = 0$ , given:  $V_1 = 10 \text{ V}$ ,  $V_2 = 50 \text{ V}$  and  $I_2 = 2 \text{ A}$ .

$$\begin{aligned} \therefore & \quad 10 = h_{12} \times 50 \\ \text{and} & \quad 2 = h_{22} \times 50 \end{aligned} \Rightarrow h_{12} = \frac{1}{5} = 0.2 \text{ and } h_{22} = \frac{1}{25} \Omega = 0.04 \Omega$$

Thus, the h-parameters are:

$$[h] = \begin{bmatrix} 25 \Omega & 0.2 \\ 2 & 0.04 \Omega^{-1} \end{bmatrix}$$

## ABCD Parameters or Transmission Parameters or Chain Parameters

These parameters are known as transmission parameters. These are generally used in the analysis of power transmission in which the input port is referred as the sending end while the output port is referred as receiving end. These are obtained by expressing voltage  $V_1$  and current  $I_1$  at input port interms of voltage  $V_2$  and current  $I_2$  at output port. Thus, voltage  $V_2$  and current  $I_2$  are independent variables while voltage  $V_1$  and current  $I_1$  are dependent variables. Thus, we have,

$$V_1 = f_1(V_2, -I_2)$$

$$I_1 = f_2(V_2, -I_2)$$

Generally we have considered the currents in both the ports are entering the port and both are positive. The negative sign with  $I_2$  indicates that, for the ABCD parameters the current  $I_2$  is leaving the port-2.

In equation form, above relations can be written as,

$$V_1 = A V_2 + B (-I_2) \quad \dots (1)$$

$$I_1 = C V_2 + D (-I_2) \quad \dots (2)$$

In matrix form, above equations can be written as,

$$\begin{bmatrix} V_1 \\ I_1 \end{bmatrix} = \begin{bmatrix} A & B \\ C & D \end{bmatrix} \begin{bmatrix} V_2 \\ -I_2 \end{bmatrix}$$

The individual ABCD parameters can be defined as follows.

[A] Let  $-I_2 = 0$ , i.e. port - 2 is circuited.

From equation (1),

$$A = \left. \frac{V_1}{V_2} \right|_{-I_2 = 0}$$

The parameter A is called open circuit reverse voltage gain. It is unitless.

From equation (2),

$$C = \left. \frac{I_1}{V_2} \right|_{-I_2 = 0} \Omega$$

The parameter C is called open circuit reverse transfer admittance.

[B] Let  $V_2 = 0$ , i.e. port - 2 is short circuited.

From equation (1),

$$B = \left. \frac{V_1}{-I_2} \right|_{V_2 = 0} \Omega$$

The parameter B is called short circuit reverse transfer impedance.

From equation (2),

$$D = \left. \frac{I_1}{-I_2} \right|_{V_2 = 0}$$

The parameter D is called short circuit reverse current gain. It is also unitless.

ABCD or Transmission parameters are also called chain parameters.

These parameter are effectively used for the analysis of power transmission line, so commonly known as transmission parameters.

Example : Find the transmission or general parameters for the circuit shown in Fig. 4.10.

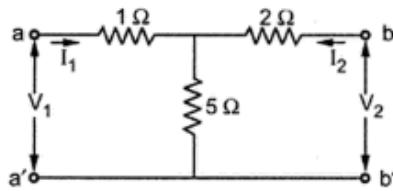


Fig. 4.10

**Solution :** By definition, transmission parameters are given by,

$$V_1 = A V_2 + B(-I_2)$$

$$I_1 = C V_2 + D(-I_2)$$

A) Let  $-I_2 = 0$  i.e. open circuit terminals b - b' (i.e. port-2) as shown in the Fig. 4.10 (a).

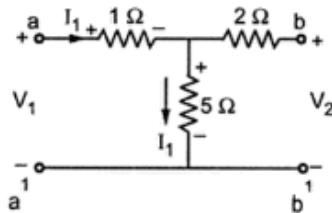


Fig. 4.10 (a)

From circuit drawn above, we can write,

$$V_2 = 5 I_1 \quad \dots (i)$$

$$\therefore C = \left. \frac{I_1}{V_2} \right|_{-I_2=0} = \frac{1}{5} \text{ S}$$

Applying KVL at the input side, we get,

$$-I_1 - 5 I_1 + V_1 = 0 \quad \dots (ii)$$

$$\therefore 6 I_1 = V_1 \quad \dots (ii)$$

From equation (i),

$$I_1 = \frac{V_2}{5} \quad \dots (iii)$$

Substituting value of  $I_1$  in equation (ii), we get,

$$\therefore 6 \left( \frac{V_2}{5} \right) = V_1$$

$$\therefore V_2 = \frac{5}{6} V_1$$

$$\therefore A = \left. \frac{V_1}{V_2} \right|_{-I_2=0} = \frac{6}{5}$$

B) Let  $V_2 = 0$  i.e. short circuit port-2 (terminals b - b') as shown in the Fig. 4.10 (b).

Applying current divider rule, we get,

$$I_2 = -I_1 \left[ \frac{5}{2+5} \right]$$

$$\therefore -I_2 = I_1 \left( \frac{5}{7} \right)$$

$$\therefore D = \left. \frac{I_1}{-I_2} \right|_{V_2=0} = \frac{7}{5}$$

Applying KVL at the input side, we get,

$$-I_1 - 5(I_1 + I_2) + V_1 = 0$$

$$\therefore -6 I_1 - 5 I_2 = -V_1$$

$$\therefore 6 I_1 + 5 I_2 = V_1$$

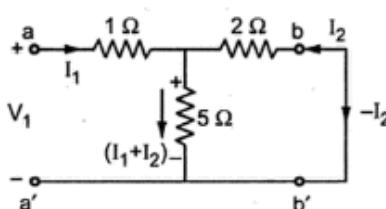


Fig. 4.10 (b)

$$\therefore 6\left(\frac{-7}{5}I_2\right) + 5I_2 = V_1$$

$$\therefore \left(\frac{-42}{5} + 5\right)I_2 = V_1$$

$$\therefore \frac{-17}{5}I_2 = V_1$$

$$\therefore \left(\frac{17}{5}\right)(-I_2) = V_1$$

$$\therefore B = \left. \frac{V_1}{-I_2} \right|_{V_2=0} = +\frac{17}{5} \Omega$$

Hence transmission parameters matrix is given by,

$$\begin{bmatrix} A & B \\ C & D \end{bmatrix} = \begin{bmatrix} \frac{6}{5} & \frac{17}{5} \\ \frac{1}{5} & \frac{7}{5} \end{bmatrix}$$

## INVERSE HYBRID (OR g) PARAMETERS

If  $V_1$  and  $I_2$  are chosen as independent variables, the two-port network equations may be written as

$$I_1 = g_{11}V_1 + g_{12}I_2$$

$$V_2 = g_{21}V_1 + g_{22}I_2.$$

In matrix form, these equations are written as

$$\begin{bmatrix} I_1 \\ V_2 \end{bmatrix} = \begin{bmatrix} g_{11} & g_{12} \\ g_{21} & g_{22} \end{bmatrix} \begin{bmatrix} V_1 \\ I_2 \end{bmatrix}.$$

The constants  $g_{11}$ ,  $g_{12}$ ,  $g_{21}$ , and  $g_{22}$  are known as inverse hybrid parameters or g-parameters. The g-parameters are defined as follows by using Equations

If  $I_2 = 0$  the output port is open circuit.

$$g_{11} = \left. \frac{I_1}{V_1} \right|_{I_2=0} \text{open circuit input admittance.}$$

$$g_{21} = \left. \frac{V_2}{V_1} \right|_{I_2=0} \text{open circuit forward voltage gain.}$$

If

$V_1 = 0$  the input port is short circuit.

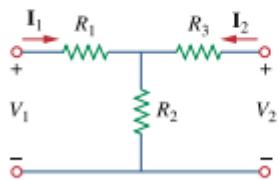
$$g_{12} = \left. \frac{I_1}{I_2} \right|_{V_1=0} \text{short circuit reverse current gain.}$$

$$g_{22} = \left. \frac{V_2}{I_2} \right|_{V_1=0} \text{short circuit output impedance.}$$

From the definitions of the g-parameters, it is seen that  $g_{11}$  has the dimensions of admittance,  $g_{21}$  and  $g_{12}$  are dimensionless, and  $g_{22}$  has the dimensions of impedance.

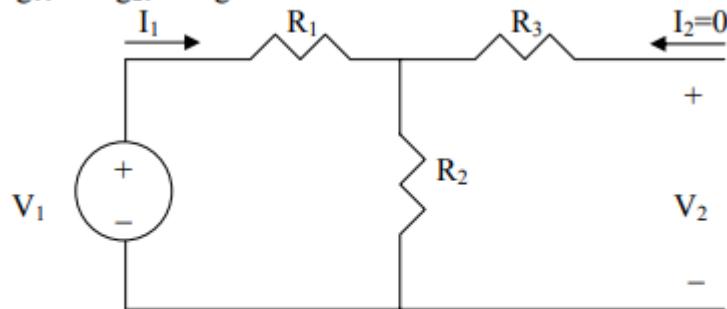
### Problem

Obtain the g parameters for the circuit of Fig.



**Figure**

We obtain  $g_{11}$  and  $g_{21}$  using the circuit below.

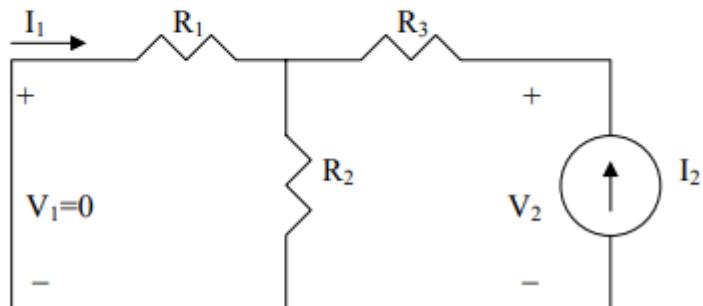


$$I_1 = \frac{V_1}{R_1 + R_2} \longrightarrow g_{11} = \frac{I_1}{V_1} = \underline{\underline{\frac{1}{R_1 + R_2}}}$$

By voltage division,

$$V_2 = \frac{R_2}{R_1 + R_2} V_1 \longrightarrow g_{21} = \frac{V_2}{V_1} = \underline{\underline{\frac{R_2}{R_1 + R_2}}}$$

We obtain  $g_{12}$  and  $g_{22}$  using the circuit below.



By current division,

$$I_1 = -\frac{R_2}{R_1 + R_2} I_2 \longrightarrow g_{12} = \frac{I_1}{I_2} = -\underline{\underline{\frac{R_2}{R_1 + R_2}}}$$

Also,

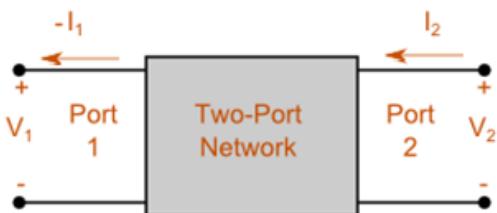
$$V_2 = I_2 (R_3 + R_1 // R_2) = I_2 \left( R_3 + \frac{R_1 R_2}{R_1 + R_2} \right) \quad g_{22} = \frac{V_2}{I_2} = \underline{\underline{R_3 + \frac{R_1 R_2}{R_1 + R_2}}}$$

$$g_{11} = \frac{1}{R_1 + R_2}, \quad g_{12} = -\frac{R_2}{R_1 + R_2}$$

$$g_{21} = \frac{R_2}{R_1 + R_2}, \quad g_{22} = R_3 + \frac{R_1 R_2}{R_1 + R_2}$$

The transmission parameter and the inverse transmission parameter are duals of each other.

If, instead of quantities  $V_1$  and  $I_1$ , quantities  $V_2$  and  $I_2$  are expressed in terms of  $V_1$  and  $I_1$ , the resulting parameter  $(A', B', C', D')$  are called inverse transmission parameter.



The inverse transmission parameters of the two port network in figure having direction of voltages and current as shown, are given by

$$V_2 = A'V_1 + B'(-I_1)$$

$$I_2 = C'V_1 + D'(-I_1)$$

in matrix form,

$$\begin{bmatrix} V_2 \\ I_2 \end{bmatrix} = \begin{pmatrix} A' & B' \\ C' & D' \end{pmatrix} \begin{bmatrix} V_1 \\ -I_1 \end{bmatrix}$$

The inverse transmission parameters can be defined as

$$A' = \frac{V_2}{V_1}; I_1 = 0 \text{ forward voltage ratio with sending end open circuited.}$$

$$C' = \frac{I_2}{V_1}; I_1 = 0 \text{ transfer admittance with sending end open circuited.}$$

$$B' = \frac{V_2}{-I_1}; V_1 = 0 \text{ transfer impedance with sending end short circuited.}$$

$$D' = \frac{I_2}{-I_1}; V_1 = 0 \text{ forward current ratio with sending end short circuited.}$$

Name of the parameters	Port variables		Equations
	Express (Dependent)	in terms of (Independent)	
Open circuit impedance i.e. z-parameters	$V_1, V_2$	$I_1, I_2$	$V_1 = z_{11}I_1 + z_{12}I_2$ $V_2 = z_{21}I_1 + z_{22}I_2$
Short circuit admittance i.e. y-parameters	$I_1, I_2$	$V_1, V_2$	$I_1 = y_{11}V_1 + y_{12}V_2$ $I_2 = y_{21}V_1 + y_{22}V_2$
h-parameters	$V_1, I_2$	$I_1, V_2$	$V_1 = h_{11}I_1 + h_{12}V_2$ $I_2 = h_{21}I_1 + h_{22}V_2$
ABCD or transmission parameters	$V_1, I_1$	$V_2, -I_2$	$V_1 = AV_2 + B(-I_2)$ $I_1 = CV_2 + D(-I_2)$
Inverse hybrid or g-parameters	$I_1, V_2$	$V_1, I_2$	$I_1 = g_{11}V_1 + g_{12}I_2$ $V_2 = g_{21}V_1 + g_{22}I_2$
Inverse transmission parameters	$V_2, I_2$	$V_1, -I_1$	$V_2 = A'V_1 + B'(-I_1)$ $I_2 = C'V_1 + D'(-I_2)$

Table 4.1 Summary of two port network parameters

## Interrelationships between the Parameters

During analysis of a two port network it is often required to obtain more than one set of parameters for a given two port network. Then it becomes time consuming and tedious to obtain different sets of parameter by using their respective basic definitions. Hence to make the analysis easier, any set of parameters is expressed in terms of all the remaining sets of parameters. In other words, parameters are derived from each other. In this section, we will derive such interrelationships between the parameters.

### **z-Parameters in terms of other Parameters**

The equations for z-parameters are as follows,

$$V_1 = z_{11} I_1 + z_{12} I_2 \quad \dots (A)$$

$$V_2 = z_{21} I_1 + z_{22} I_2 \quad \dots (B)$$

#### [A] Terms of y-Parameters

The equations for y-parameters are as follows,

$$I_1 = y_{11} V_1 + y_{12} V_2 \quad \dots (1)$$

$$I_2 = y_{21} V_1 + y_{22} V_2 \quad \dots (2)$$

Writing equations in matrix form, we have,

$$\begin{bmatrix} I_1 \\ I_2 \end{bmatrix} = \begin{bmatrix} y_{11} & y_{12} \\ y_{21} & y_{22} \end{bmatrix} \begin{bmatrix} V_1 \\ V_2 \end{bmatrix}$$

Solving above equations, using Cramer's rule for  $V_1$  and  $V_2$ . We can write,

$$V_1 = \frac{\begin{vmatrix} I_1 & y_{12} \\ I_2 & y_{22} \end{vmatrix}}{\begin{vmatrix} y_{11} & y_{12} \\ y_{21} & y_{22} \end{vmatrix}} = \frac{y_{22} I_1 - y_{12} I_2}{y_{11} y_{22} - y_{12} y_{21}} \quad \dots (3)$$

$$= \frac{y_{22}}{y_{11} y_{22} - y_{12} y_{21}} \cdot I_1 - \frac{y_{12}}{y_{11} y_{22} - y_{12} y_{21}} I_2$$

$$\text{Similarly, } V_2 = \frac{\begin{vmatrix} y_{11} & I_1 \\ y_{21} & I_2 \end{vmatrix}}{\begin{vmatrix} y_{11} & y_{12} \\ y_{21} & y_{22} \end{vmatrix}} = \frac{y_{11} I_2 - y_{21} I_1}{y_{11} y_{22} - y_{12} y_{21}} \quad \dots (4)$$

$$= \frac{-y_{21}}{y_{11} y_{22} - y_{12} y_{21}} I_1 + \frac{y_{11}}{y_{11} y_{22} - y_{12} y_{21}} I_2$$

Let  $y_{11} y_{22} - y_{12} y_{21} = \Delta y$

Rewriting equations (3) and (4), we have,

$$V_1 = \left[ \frac{y_{22}}{\Delta y} \right] I_1 + \left[ \frac{-y_{12}}{\Delta y} \right] I_2 \quad \dots (5)$$

$$\text{and } V_2 = \left[ \frac{-y_{21}}{\Delta y} \right] I_1 + \left[ \frac{y_{11}}{\Delta y} \right] I_2 \quad \dots (6)$$

Comparing equations (5) and (6) to equations (A) and (B), we have,

$z_{11} = \frac{y_{22}}{\Delta y}$	$z_{12} = \frac{-y_{12}}{\Delta y}$
$z_{21} = \frac{-y_{21}}{\Delta y}$	$z_{22} = \frac{y_{11}}{\Delta y}$

### [B] Interms of h-Parameters

The equations for h-parameters are as follows,

$$V_1 = h_{11} I_1 + h_{12} V_2 \quad \dots (1)$$

$$I_2 = h_{21} I_1 + h_{22} V_2 \quad \dots (2)$$

From equation (2) we can write,

$$\begin{aligned} h_{22} V_2 &= -h_{21} I_1 + I_2 \\ \therefore V_2 &= \left[ \frac{-h_{21}}{h_{22}} \right] I_1 + \left[ \frac{1}{h_{22}} \right] I_2 \end{aligned} \quad \dots (3)$$

Substituting value of  $V_2$  in equation (1) we have,

$$\begin{aligned} V_1 &= h_{11} I_1 + h_{12} \left[ \frac{-h_{21}}{h_{22}} I_1 + \frac{1}{h_{22}} I_2 \right] \\ \therefore V_1 &= \left[ h_{11} - \frac{h_{12} h_{21}}{h_{22}} \right] I_1 + \frac{h_{12}}{h_{22}} I_2 \\ \therefore V_1 &= \left[ \frac{h_{11} h_{22} - h_{12} h_{21}}{h_{22}} \right] I_1 + \left[ \frac{h_{12}}{h_{22}} \right] I_2 \end{aligned} \quad \dots (4)$$

Comparing equations (4) and (3) with equations (A) and (B) respectively, we have,

$z_{11} = \frac{h_{11} h_{22} - h_{12} h_{21}}{h_{22}}$	$z_{12} = \frac{h_{12}}{h_{22}}$
$z_{21} = \frac{-h_{21}}{h_{22}}$	$z_{22} = \frac{1}{h_{22}}$

### [C] Interms of Transmission (ABCD) Parameters

The equations for transmission parameters are as follows,

$$V_1 = A V_2 + B (-I_2) \quad \dots (1)$$

$$I_1 = C V_2 + D (-I_2) \quad \dots (2)$$

We can rewrite equation (2) as follows,

$$\begin{aligned} C V_2 &= I_1 + D I_2 \\ \therefore V_2 &= \left[ \frac{1}{C} \right] I_1 + \left[ \frac{D}{C} \right] I_2 \end{aligned} \quad \dots (3)$$

Substituting value of  $V_2$  in equation (1), we have,

$$\begin{aligned} V_1 &= A \left[ \left( \frac{1}{C} \right) I_1 + \left( \frac{D}{C} \right) I_2 \right] - B \cdot I_2 \\ \therefore V_1 &= \left[ \frac{A}{C} \right] I_1 + \left[ \frac{AD}{C} - B \right] I_2 \\ \therefore V_1 &= \left[ \frac{A}{C} \right] I_1 + \left[ \frac{AD - BC}{C} \right] I_2 \end{aligned} \quad \dots (4)$$

Comparing equations (4) and (3) with the equations (A) and (B) respectively, we have,

$z_{11} = \frac{A}{C}$	$z_{12} = \frac{AD - BC}{C}$
$z_{21} = \frac{1}{C}$	$z_{22} = \frac{D}{C}$

## y-Parameters In terms of other Parameters

The equations for y-parameters are as follows,

$$I_1 = y_{11} V_1 + y_{12} V_2 \quad \dots (A)$$

$$I_2 = y_{21} V_1 + y_{22} V_2 \quad \dots (B)$$

### [A] In terms of z-Parameters

The equations for z-parameters are as follows,

$$V_1 = z_{11} I_1 + z_{12} I_2 \quad \dots (1)$$

$$V_2 = z_{21} I_1 + z_{22} I_2 \quad \dots (2)$$

Writing equations in matrix form,

$$\begin{bmatrix} V_1 \\ V_2 \end{bmatrix} = \begin{bmatrix} z_{11} & z_{12} \\ z_{21} & z_{22} \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \end{bmatrix}$$

Solving above equations using Cramer's rule for  $I_1$  and  $I_2$ , we can write,

$$I_1 = \frac{\begin{vmatrix} V_1 & z_{12} \\ V_2 & z_{22} \end{vmatrix}}{\begin{vmatrix} z_{11} & z_{12} \\ z_{21} & z_{22} \end{vmatrix}} = \frac{z_{22} \cdot V_1 - z_{12} \cdot V_2}{z_{11} \cdot z_{22} - z_{12} \cdot z_{21}} \quad \dots (3)$$

$$= \frac{z_{22}}{z_{11} z_{22} - z_{12} z_{21}} \cdot V_1 + \frac{-z_{12}}{z_{11} z_{22} - z_{12} z_{21}} \cdot V_2$$

Similarly,  $I_2 = \frac{\begin{vmatrix} z_{11} & V_1 \\ z_{21} & V_2 \end{vmatrix}}{\begin{vmatrix} z_{11} & z_{12} \\ z_{21} & z_{22} \end{vmatrix}} = \frac{z_{11} \cdot V_2 - z_{21} \cdot V_1}{z_{11} \cdot z_{22} - z_{12} \cdot z_{21}}$

$$= \frac{-z_{21}}{z_{11} \cdot z_{22} - z_{12} z_{21}} \cdot V_1 + \frac{z_{11}}{z_{11} z_{22} - z_{12} z_{21}} \cdot V_2 \quad \dots (4)$$

Let  $\Delta z = z_{11} \cdot z_{22} - z_{12} \cdot z_{21}$

Rewriting equations (3) and (4),

$$I_1 = \left[ \frac{z_{22}}{\Delta z} \right] V_1 + \left[ \frac{-z_{12}}{\Delta z} \right] V_2 \quad \dots (5)$$

and  $I_2 = \left[ \frac{-z_{21}}{\Delta z} \right] V_1 + \left[ \frac{z_{11}}{\Delta z} \right] V_2 \quad \dots (6)$

Comparing equations (5) and (6) with the equations (A) and (B) respectively, we have,

$y_{11} = \frac{z_{22}}{\Delta z}$	$y_{12} = \frac{-z_{12}}{\Delta z}$
$y_{21} = \frac{-z_{21}}{\Delta z}$	$y_{22} = \frac{z_{11}}{\Delta z}$

### [B] In terms of h-Parameters

The equations for h-parameters are as follows,

$$V_1 = h_{11} I_1 + h_{12} V_2 \quad \dots (1)$$

$$I_2 = h_{21} I_1 + h_{22} V_2 \quad \dots (2)$$

We can rewrite equation (1) as follows,

$$h_{11} I_1 = V_1 - h_{12} V_2$$

$$\therefore I_1 = \left[ \frac{1}{h_{11}} \right] V_1 + \left[ \frac{-h_{12}}{h_{11}} \right] V_2 \quad \dots (3)$$

Substituting equation (3) in equation (2), we have,

$$\begin{aligned} I_2 &= h_{21} \left[ \frac{1}{h_{11}} V_1 - \frac{h_{12}}{h_{11}} V_2 \right] + h_{22} V_2 \\ \therefore I_2 &= \left[ \frac{h_{21}}{h_{11}} \right] V_1 + \left[ h_{22} - \frac{h_{12} h_{21}}{h_{11}} \right] V_2 \\ \therefore I_2 &= \left[ \frac{h_{21}}{h_{11}} \right] V_1 + \left[ \frac{h_{11} h_{22} - h_{12} h_{21}}{h_{11}} \right] V_2 \end{aligned} \quad \dots (4)$$

Comparing equations (3) and (4) with the equations (A) and (B) respectively, we have,

$y_{11} = \frac{1}{h_{11}}$	$y_{12} = \frac{-h_{12}}{h_{11}}$
$y_{21} = \frac{h_{21}}{h_{11}}$	$y_{22} = \frac{h_{11} h_{22} - h_{12} h_{21}}{h_{11}}$

### [C] Interms of Transmission (ABCD) Parameters

The equations for transmission parameters are as follows,

$$V_1 = A V_2 + B (-I_2) \quad \dots (1)$$

$$I_1 = C V_2 + D (-I_2) \quad \dots (2)$$

We can rewrite equation (1) as follows,

$$-B \cdot I_2 = V_1 - A V_2$$

$$\therefore I_2 = \left[ -\frac{1}{B} \right] V_1 + \left[ \frac{A}{B} \right] V_2 \quad \dots (3)$$

Substituting value of  $I_2$  in equation (2), we have,

$$\begin{aligned} I_1 &= C \cdot V_2 + D \left[ \frac{+1}{B} V_1 - \frac{A}{B} V_2 \right] \\ \therefore I_1 &= \left[ \frac{D}{B} \right] V_1 + \left[ C - \frac{AD}{B} \right] V_2 \\ \therefore I_1 &= \left[ \frac{D}{B} \right] V_1 + \left[ \frac{BC - AD}{B} \right] V_2 \end{aligned} \quad \dots (4)$$

Comparing equations (4) and (3) with the equations (A) and (B) respectively, we have,

$y_{11} = \frac{D}{B}$	$y_{12} = \frac{BC - AD}{B}$
$y_{21} = \frac{-1}{B}$	$y_{22} = \frac{A}{B}$

### **h-Parameters interms of other Parameters**

The equations for h-parameters are as follows,

$$V_1 = h_{11} I_1 + h_{12} V_2 \quad \dots (A)$$

$$I_2 = h_{21} I_1 + h_{22} V_2 \quad \dots (B)$$

### [A] Interms of z-Parameters

The equations for z-parameter are as follows,

$$V_1 = z_{11} I_1 + z_{12} I_2 \quad \dots (1)$$

$$V_2 = z_{21} I_1 + z_{22} I_2 \quad \dots (2)$$

We can rewrite equations (2) as follows,

$$\begin{aligned} z_{22} I_2 &= -z_{21} I_1 + V_2 \\ I_2 &= \left[ \frac{-z_{21}}{z_{22}} \right] I_1 + \left[ \frac{+1}{z_{22}} \right] V_2 \end{aligned} \quad \dots (3)$$

Substituting value of  $I_2$  in equation (1), we have,

$$\begin{aligned} V_1 &= z_{11} I_1 + z_{12} \left( \left[ \frac{-z_{21}}{z_{22}} \right] I_1 + \left[ \frac{+1}{z_{22}} \right] V_2 \right) \\ \therefore V_1 &= \left[ \frac{z_{11} z_{22} - z_{12} z_{21}}{z_{22}} \right] I_1 + \left[ \frac{z_{12}}{z_{22}} \right] V_2 \end{aligned} \quad \dots (4)$$

Comparing equations (4) and (3) with equations (A) and (B) respectively, we have,

$h_{11} = \frac{z_{11} z_{22} - z_{12} z_{21}}{z_{22}}$	$h_{12} = \frac{z_{12}}{z_{22}}$
$h_{21} = \frac{-z_{21}}{z_{22}}$	$h_{22} = \frac{1}{z_{22}}$

### [B] Terms of y-Parameters

The equations of y-parameters are as follows,

$$I_1 = y_{11} V_1 + y_{12} V_2 \quad \dots (1)$$

$$I_2 = y_{21} V_1 + y_{22} V_2 \quad \dots (2)$$

We can rewrite equation (1) as follows,

$$\begin{aligned} y_{11} V_1 &= I_1 - y_{12} V_2 \\ \therefore V_1 &= \left[ \frac{1}{y_{11}} \right] I_1 + \left[ \frac{-y_{12}}{y_{11}} \right] V_2 \end{aligned} \quad \dots (3)$$

Substituting value of  $V_1$  in equation (2), we have,

$$\begin{aligned} I_2 &= y_{21} \left( \left[ \frac{1}{y_{11}} \right] I_1 + \left[ \frac{-y_{12}}{y_{11}} \right] V_2 \right) + y_{22} V_2 \\ \therefore I_2 &= \left[ \frac{y_{21}}{y_{11}} \right] I_1 + \left[ y_{22} - \frac{y_{12} y_{21}}{y_{11}} \right] V_2 \\ \therefore I_2 &= \left[ \frac{y_{21}}{y_{11}} \right] I_1 + \left[ \frac{y_{11} y_{22} - y_{12} y_{21}}{y_{11}} \right] V_2 \end{aligned} \quad \dots (4)$$

Comparing equations (3) and (4) with equation (A) and (B) respectively, we have,

$h_{11} = \frac{1}{y_{11}}$	$h_{12} = \frac{-y_{12}}{y_{11}}$
$h_{21} = \frac{y_{21}}{y_{11}}$	$h_{22} = \frac{y_{11} y_{22} - y_{12} y_{21}}{y_{11}}$

### [C] Terms of Transmission Parameters

The equations for transmission parameters are as follows,

$$V_1 = A V_2 + B (-I_2) \quad \dots (1)$$

$$I_2 = C V_2 + D (-I_2) \quad \dots (2)$$

We can rewrite equations (2) as follows,

$$\begin{aligned} D(I_2) &= -I_1 + C V_2 \\ \therefore I_2 &= \left[ \frac{-1}{D} \right] I_1 + \left[ \frac{C}{D} \right] V_2 \end{aligned} \quad \dots (3)$$

Substituting value of  $I_2$  in equation (1), we have,

$$\begin{aligned} V_1 &= A V_2 - B \left( \left[ \frac{-1}{D} \right] I_1 + \left[ \frac{C}{D} \right] V_2 \right) \\ \therefore V_1 &= \left[ A - \frac{BC}{D} \right] V_2 + \left[ \frac{B}{D} \right] I_1 \\ \therefore V_1 &= \left[ \frac{B}{D} \right] I_1 + \left[ \frac{AD - BC}{D} \right] V_2 \end{aligned} \quad \dots (4)$$

Comparing equations (4) and (3) with equations (A) and (B) respectively we have,

$h_{11} = \frac{B}{D}$	$h_{12} = \frac{AD - BC}{D}$
$h_{21} = \frac{-1}{D}$	$h_{22} = \frac{C}{D}$

### T-parameters in terms of other parameters

#### (i) T-parameters in terms of Z-parameters:

**Step-I:** We know that T-parameter equations;

$$V_1 = AV_2 - BI_2 \quad \dots (1)$$

$$I_1 = CV_2 - DI_2 \quad \dots (2)$$

and Z-parameters equations;

$$V_1 = Z_{11} I_1 + Z_{12} I_2 \quad \dots (3)$$

$$V_2 = Z_{21} I_1 + Z_{22} I_2 \quad \dots (4)$$

**Step-II:** In order to obtain as equation (1), eliminated  $I_1$  from equations (3) and (4).

From equation (3), we have

$$I_1 = \frac{V_1 - Z_{12} I_2}{Z_{11}}$$

Put this value in equation (4), then we get

$$V_2 = Z_{21} \left( \frac{V_1 - Z_{12} I_2}{Z_{11}} \right) + Z_{22} I_2$$

or  $V_2 = \frac{Z_{21}}{Z_{11}} V_1 + I_2 \left( Z_{22} - \frac{Z_{12} Z_{21}}{Z_{11}} \right)$

or  $\frac{Z_{21}}{Z_{11}} V_1 = V_2 - I_2 \left( \frac{Z_{11} Z_{22} - Z_{12} Z_{21}}{Z_{11}} \right)$

or  $V_1 = \frac{Z_{11}}{Z_{21}} V_2 - \left( \frac{Z_{11} Z_{22} - Z_{12} Z_{21}}{Z_{21}} \right) I_2 \quad \dots (5)$

For obtaining as equations (2), rewrite equation (4), then we get

$$Z_{21} I_1 = V_2 - Z_{22} I_2$$

or  $I_1 = \frac{1}{Z_{21}} V_2 - \frac{Z_{22}}{Z_{21}} I_2 \quad \dots (6)$

**Step-III:** Comparing equations (1) and (5), then we get

$A = \frac{Z_{11}}{Z_{21}}$ and $B = \frac{Z_{11} Z_{22} - Z_{12} Z_{21}}{Z_{21}} = \frac{\Delta Z}{Z_{21}}$
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Similary, comparing equation (2) and (6), then we get

$C = \frac{1}{Z_{21}}$ and $D = \frac{Z_{22}}{Z_{21}}$
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**(ii) T-parameters in terms of Y-parameters:**

**Step-I:** We know that T-parameter equations:

$$V_1 = AV_2 - BI_2 \quad \dots(1)$$

$$I_1 = CV_2 - DI_2 \quad \dots(2)$$

and Y-parameters equations:

$$I_1 = Y_{11} V_1 + Y_{12} V_2 \quad \dots(3)$$

$$I_2 = Y_{21} V_1 + Y_{22} V_2 \quad \dots(4)$$

$V_1 = \frac{I_2 - Y_{22} V_2}{Y_{21}}$ $V_1 = \frac{1}{Y_{21}} I_2 - \frac{Y_{22}}{Y_{21}} V_2$ $V_1 = -\frac{Y_{22}}{Y_{21}} V_2 + \frac{1}{Y_{21}} I_2 \quad \text{---(5)}$ <p>compare equ(5) with (1)</p> <div style="border: 1px solid black; padding: 5px; display: inline-block;"> <math display="block">A = -\frac{Y_{22}}{Y_{21}} \quad ; \quad B = \frac{1}{Y_{21}}</math> </div>	$I_1 = Y_{11} \left[ \frac{I_2 - Y_{22} V_2}{Y_{21}} \right] + Y_{12} V_2$ $I_1 = \frac{Y_{11}}{Y_{21}} I_2 - \frac{Y_{11} Y_{22}}{Y_{21}} V_2 + Y_{12} V_2$ $I_1 = \frac{Y_{11}}{Y_{21}} I_2 - \left[ \frac{Y_{11} Y_{22} - Y_{12} Y_{21}}{Y_{21}} \right] V_2 \quad \text{---(6)}$ <p>I<sub>1</sub> &amp; compare equ(6) with (2)</p> <div style="border: 1px solid black; padding: 5px; display: inline-block;"> <math display="block">D = -\frac{Y_{11}}{Y_{21}} \quad ; \quad C = \frac{-Y_{11} Y_{22} + Y_{12} Y_{21}}{Y_{21}}</math> </div>
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**(iv) T-parameters in terms of h-parameters:**

**Step-I:** We know that T-parameter equations;

$$V_1 = AV_2 - BI_2 \quad \dots(1)$$

$$I_1 = CV_2 - DI_2 \quad \dots(2)$$

and h-parameters equations:

$$V_1 = h_{11} I_1 + h_{12} V_2 \quad \dots(3)$$

$$I_2 = h_{21} I_1 + h_{22} V_2 \quad \dots(4)$$

**Step-II:** In order to obtain as equation (1), eliminated I<sub>1</sub> from equations (3) and (4).

From equation (3), we have

$$I' = \frac{V_1 - h_{12} V_2}{h_{11}}$$

Put this value in equation (4), then we get

$$I_2 = h_{21} \left( \frac{V_1 - h_{12} V_2}{h_{11}} \right) + h_{22} V_2$$

or  $I_2 = \frac{h_{21}}{h_{11}} V_1 + V_2 \left( h_{22} - \frac{h_{12} h_{21}}{h_{11}} \right)$

or  $\frac{h_{21}}{h_{11}} V_1 = \left( \frac{h_{12} h_{21} - h_{11} h_{22}}{h_{11}} \right) V_2 + I_2$

or  $V_1 = \left( \frac{h_{12} h_{21} - h_{11} h_{22}}{h_{21}} \right) V_2 + \frac{h_{11}}{h_{21}} I_2 \quad \dots(5)$

For obtaining as equation (2), rewrite equation (4), then we get

$$h_{21} I_1 = -h_{22} V_2 + I_2$$

or  $I_1 = -\frac{h_{22}}{h_{21}} V_2 + \frac{I_2}{h_{21}} \quad \dots(6)$

**Step-III:** Comparing equations (1) and (5), then we get

$$A = \frac{h_{12} h_{21} - h_{11} h_{22}}{h_{21}} = \frac{-\Delta h}{h_{21}} \text{ and } B = \frac{-h_{11}}{h_{21}}$$

Similarly, comparing equations (2) and (6), then we get

$$C = \frac{-h_{22}}{h_{21}} \text{ and } D = -\frac{1}{h_{21}}$$

## INTERCONNECTIONS OF TWO-PORT NETWORKS

Two-port networks may be interconnected in various configurations, such as series, parallel, cascade, series-parallel, and parallel-series connections. For each configuration a certain set of parameters may be more useful than others to describe the network.

### Series Connection

Figure 10.19 shows a series connection of two-port networks  $N_a$  and  $N_b$ .

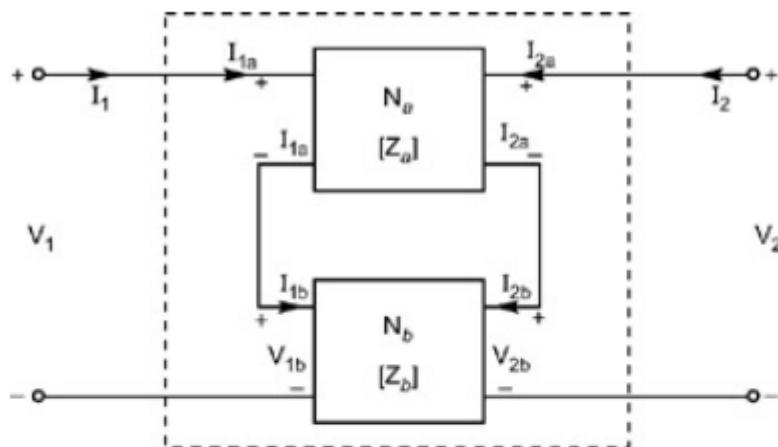


Figure 10.19: Series connection of two two-port networks  
For network  $N_a$ ,

$$\begin{bmatrix} V_{1a} \\ V_{2a} \end{bmatrix} = \begin{bmatrix} Z_{11a} & Z_{12a} \\ Z_{21a} & Z_{22a} \end{bmatrix} \begin{bmatrix} I_{1a} \\ I_{2a} \end{bmatrix}$$

$$(10.62) V_{1a} = Z_{11a} I_{1a} + Z_{12a} I_{2a}$$

$$(10.63) V_{2a} = Z_{21a} I_{1a} + Z_{22a} I_{2a}$$

For network  $N_b$ ,

$$\begin{bmatrix} V_{1b} \\ V_{2b} \end{bmatrix} = \begin{bmatrix} Z_{11b} & Z_{12b} \\ Z_{21b} & Z_{22b} \end{bmatrix} \begin{bmatrix} I_{1b} \\ I_{2b} \end{bmatrix}$$

$$(10.64) V_{1b} = Z_{11b} I_{1b} + Z_{12b} I_{2b}$$

$$(10.65) V_{2b} = Z_{21b} I_{1b} + Z_{22b} I_{2b}$$

The condition for series connection is

$$I_{1a} = I_{1b} = I_1, \text{ and } I_{2a} = I_{2b} = I_2 \text{ (current same)}$$

$$(10.66) V_1 = V_{1a} + V_{1b}$$

$$(10.67) V_2 = V_{2a} + V_{2b}$$

Putting the values of  $V_{1a}$  and  $V_{1b}$  from Equation (10.62) and Equation (10.64),

$$(10.68) \quad \begin{aligned} V_1 &= Z_{11a}I_{1a} + Z_{12a}I_{2a} + Z_{11b}I_{1b} + Z_{12b}I_{2b} \\ &= Z_{11a}I_1 + Z_{12a}I_2 + Z_{11b}I_1 + Z_{12b}I_2 \quad [I_{1a} = I_{1b} = I_1, I_{2a} = I_{2b} = I_2] \\ V_1 &= (Z_{11a} + Z_{11b})I_1 + (Z_{12a} + Z_{12b})I_2. \end{aligned}$$

Putting the values of  $V_{2a}$  and  $V_{2b}$  from Equation (10.63) and Equation (10.65) into Equation (10.67), we get

$$(10.69) V_2 = (Z_{21a} + Z_{21b})I_1 + (Z_{22a} + Z_{22b})I_2$$

The Z-parameters of the series-connected combined network can be written as

$$V_1 = Z_{11}I_1 + Z_{12}I_2$$

$$V_2 = Z_{21}I_1 + Z_{22}I_2,$$

where

$$Z_{11} = Z_{11a} + Z_{11b}$$

$$Z_{12} = Z_{12a} + Z_{12b}$$

$$Z_{21} = Z_{21a} + Z_{21b}$$

$$Z_{22} = Z_{22a} + Z_{22b}$$

or in the matrix form,

$$[Z] = [Z_a] + [Z_b].$$

The overall Z-parameter matrix for series connected two-port networks is simply

the sum of Z-parameter matrices of each individual two-port network connected in series.

## Parallel Connection

Figure 10.20 shows a parallel connection of two two-port networks  $N_a$  and  $N_b$ .

The resultant of two admittances connected in parallel is  $Y_1 + Y_2$ . So in parallel connection, the parameters are Y-parameters.

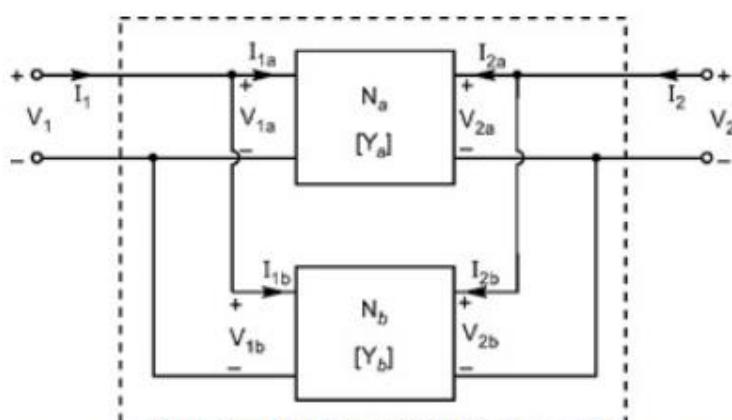


Figure 10.20: Parallel connections for two two-port networks  
For network  $N_a$ ,

$$\begin{bmatrix} I_{1a} \\ I_{2a} \end{bmatrix} = \begin{bmatrix} Y_{11a} & Y_{12a} \\ Y_{21a} & Y_{22a} \end{bmatrix} \begin{bmatrix} V_{1a} \\ V_{2a} \end{bmatrix}$$

or

$$(10.70) I_{1a} = Y_{11a}V_{1a} + Y_{12a}V_{2a}$$

$$(10.71) I_{2a} = Y_{21a}V_{1a} + Y_{22a}V_{2a}$$

For network N<sub>b</sub>,

$$\begin{bmatrix} I_{1b} \\ I_{2b} \end{bmatrix} = \begin{bmatrix} Y_{11b} & Y_{12b} \\ Y_{21b} & Y_{22b} \end{bmatrix} \begin{bmatrix} V_{1b} \\ V_{2b} \end{bmatrix}$$

$$(10.72) I_{1b} = Y_{11b}V_{1b} + Y_{12b}V_{2b}$$

$$(10.73) I_{2b} = Y_{21b}V_{1b} + Y_{22b}V_{2b}$$

Now the condition for parallel,

$$V_{1a} = V_{1b} = V_1, \quad V_{2a} = V_{2b} = V_2 \quad [\text{Same voltage}]$$

and

$$(10.74) I_1 = I_{1a} + I_{1b}$$

$$(10.75) I_2 = I_{2a} + I_{2b}$$

$$I_1 = Y_{11a}V_{1a} + Y_{12a}V_{2a} + Y_{11b}V_{1b} + Y_{12b}V_{2b}$$

$$= Y_{11a}V_1 + Y_{12a}V_2 + Y_{11b}V_1 + Y_{12b}V_2$$

$$(10.76) I_1 = (Y_{11a} + Y_{11b})V_1 + (Y_{12a} + Y_{12b})V_2$$

Similarly,

$$(10.77) I_2 = (Y_{21a} + Y_{21b})V_1 + (Y_{22a} + Y_{22b})V_2$$

The Y-parameters of the parallel connected combined network can be written as

$$I_1 = Y_{11}V_1 + Y_{12}V_2$$

$$I_2 = Y_{21}V_1 + Y_{22}V_2$$

$$Y_{11} = Y_{11a} + Y_{11b}$$

$$Y_{12} = Y_{12a} + Y_{12b}$$

$$Y_{21} = Y_{21a} + Y_{21b}$$

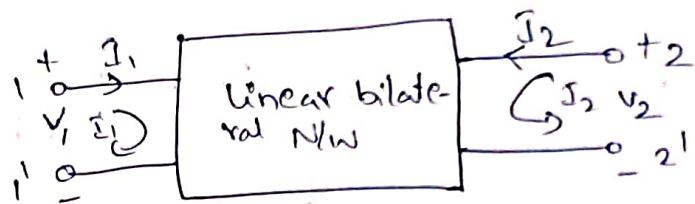
$$Y_{22} = Y_{22a} + Y_{22b}$$



## V - TWO PORT NETWORKS

General Two Port Networks :-

- \* we will consider a general two port network composed of linear, bilateral elements and no independent sources.
- \* Dependent sources are permitted.
- \* It is represented as a black box with two accessible terminals pairs as shown below.



- \* The terminal pair  $1-1'$  represents port 1 and is called input port or sending end.
- \* The terminal pair  $(2-2')$  represent port 2 and is called output port or receiving end.
- \* The voltage and current at port (1) are  $V_1$  and  $I_1$ , and at port (2) are  $V_2$  and  $I_2$ .
- \* The polarities of  $v_1$  and  $v_2$  and the directions of  $i_1$  and  $i_2$  are customarily selected as shown above figure.
- \* Out of the 4 variables  $V_1, I_1, V_2$  and  $I_2$  only two are independent.
- \* The other two are expressed in terms of the independent variables in terms of n/w parameters.

This can be done in number of ways as shown in below Table.

Name of parameters	Expressed (dependent)	In terms of (independent)	Equations
1. open circuit Impedance Parameters (or) (Z - parameters)	$V_1, V_2$	$I_1$ and $I_2$	$\begin{bmatrix} V_1 \\ V_2 \end{bmatrix} = \begin{bmatrix} Z_{11} & Z_{12} \\ Z_{21} & Z_{22} \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \end{bmatrix}$
2. short circuit admittance parameters or Y - parameters	$I_1, I_2$	$V_1$ and $V_2$	$\begin{bmatrix} I_1 \\ I_2 \end{bmatrix} = \begin{bmatrix} Y_{11} & Y_{12} \\ Y_{21} & Y_{22} \end{bmatrix} \begin{bmatrix} V_1 \\ V_2 \end{bmatrix}$
3. Transmission Parameters (ABCD)	$V_1, I_1$	$V_2$ and $I_2$	$\begin{bmatrix} V_1 \\ I_1 \end{bmatrix} = \begin{bmatrix} A & B \\ C & D \end{bmatrix} \begin{bmatrix} V_2 \\ -I_2 \end{bmatrix}$
4. Hybrid parameters (h - parameters)	$V_1, I_2$	$I_1$ and $V_2$	$\begin{bmatrix} V_1 \\ V_2 \end{bmatrix} = \begin{bmatrix} h_{11} & h_{12} \\ h_{21} & h_{22} \end{bmatrix} \begin{bmatrix} I_1 \\ V_2 \end{bmatrix}$

### open circuit Impedance (Z) parameters:-

A general linear two-port network defined in below, which does not contain any independent sources is shown below.



- \* The Z parameters of a two-port in the positive directions of voltages and currents may be defined by expressing the port voltages  $V_1$  and  $V_2$  in terms of the currents  $I_1$  and  $I_2$ .
- \*  $V_1$  and  $V_2$  are dependent variables, and  $I_1$  and  $I_2$  are independent variables.

The voltages at port 1-1' and port 2-2' are

H.F

$$V_1 = Z_{11}I_1 + Z_{12}I_2 \rightarrow ①$$

$$V_2 = Z_{21}I_1 + Z_{22}I_2 \rightarrow ②$$

\* Here  $Z_{11}$ ,  $Z_{12}$ ,  $Z_{21}$  and  $Z_{22}$  are the network functions, and are called Impedance (Z) parameters.

\* These parameters can be represented by matrices

we may write the matrix equation

$$[V] = [Z] [I]$$

where  $V$  is the column matrix =  $\begin{bmatrix} V_1 \\ V_2 \end{bmatrix}$

$Z$  is the square matrix =  $\begin{bmatrix} Z_{11} & Z_{12} \\ Z_{21} & Z_{22} \end{bmatrix}$

and we may write  $I$  in the column matrix =  $\begin{bmatrix} I_1 \\ I_2 \end{bmatrix}$

$$\text{Thus, } \begin{bmatrix} V_1 \\ V_2 \end{bmatrix} = \begin{bmatrix} Z_{11} & Z_{12} \\ Z_{21} & Z_{22} \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \end{bmatrix}$$

\* The individual Z-parameters for a given network can be defined by setting each of the port current equal to zero.

Suppose port 2-2' is left open-circuited, then  $I_2 = 0$

$$\text{Thus } Z_{11} = \frac{V_1}{I_1} \Big|_{I_2=0}$$

where  $Z_{11}$  is the driving-point impedance at port 1-1' with port 2-2' open-circuited. It is called the open-circuit input impedance.

$$\text{Similarly } Z_{21} = \frac{V_2}{I_1} \Big|_{I_2=0}$$

where  $Z_{21}$  is transfer impedance, also called open ckt forward transfer impedance.

Suppose port 1-1' is left open circuited, then  $I_1=0$

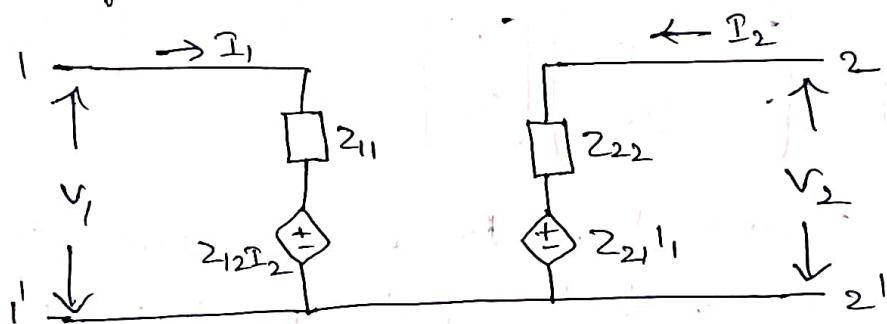
$$\text{Thus } Z_{12} = \frac{V_1}{I_2} \Big|_{I_1=0}$$

where  $Z_{12}$  is the transfer impedance, and it is also called the open circuit reverse transfer impedance.

$$Z_{22} = \frac{V_2}{I_2} \Big|_{I_1=0}$$

where  $Z_{22}$  is the open ckt driving point impedance and it is also called the open ckt output impedance.

The equivalent circuit of the two-port networks governed by eqn ① & ②



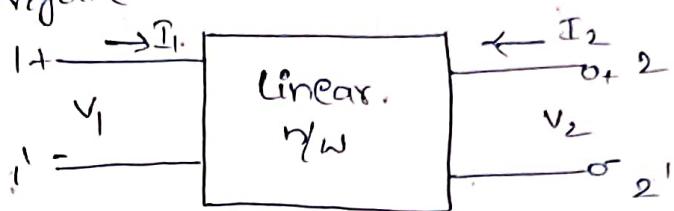
If the network under study is reciprocal or bilateral, then in accordance with the reciprocity principle

$$\frac{V_2}{I_1} \Big|_{I_2=0} = \frac{V_1}{I_2} \Big|_{I_1=0}$$

$$\text{or } Z_{21} = Z_{12}$$

Short circuit Admittance (Y) Parameters:-

A general two-port network which is considered as shown in below figure.



The Y parameters of a two-port network may be defined by expressing the two-port currents  $I_1$  and  $I_2$  in terms of the two port voltages  $V_1$  and  $V_2$ .

$$\text{Thus, } I_1 = Y_{11}V_1 + Y_{12}V_2$$

$$I_2 = Y_{21}V_1 + Y_{22}V_2$$

In matrix form, we can write

$$\begin{bmatrix} I_1 \\ I_2 \end{bmatrix} = \begin{bmatrix} Y_{11} & Y_{12} \\ Y_{21} & Y_{22} \end{bmatrix} \begin{bmatrix} V_1 \\ V_2 \end{bmatrix}$$

$$[I] = [Y][v]$$

The individual Y parameters for a given network can be defined by setting each of the port voltages equal to zero.

Case1:- when the output port is short circuited i.e  $V_2 = 0$

$$Y_{11} = \frac{I_1}{V_1} \Big|_{V_2=0}$$

where  $Y_{11}$  is the driving point admittance with the output port short-circuited. It is also called short circuit input admittance.

$$\text{Similarly, } Y_{21} = \frac{I_2}{V_1} \Big|_{V_2=0}$$

where  $Y_{21}$  is the transfer admittance with the output port short-circuited. It is also called short-circuit forward transfer admittance.

Case 2:- when the  $\text{V}_{\text{pp}}$  port is short-circuited i.e.  $V_1=0$

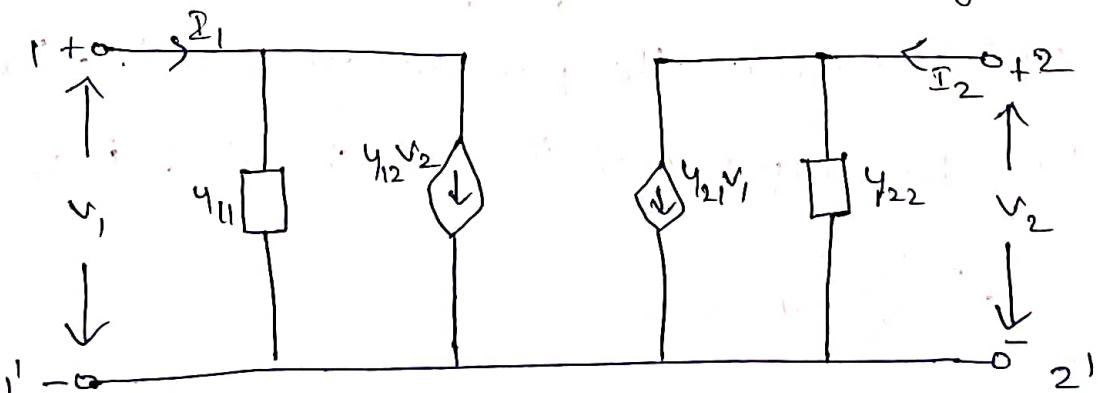
$$Y_{12} = \frac{I_1}{V_2} \Big|_{V_1=0}$$

where  $Y_{12}$  is the transfer admittance with the  $\text{V}_{\text{pp}}$  port short-circuited. It is also called short-circuit reverse transfer admittance.

$$\text{Similarly, } Y_{22} = \frac{I_2}{V_2} \Big|_{V_1=0}$$

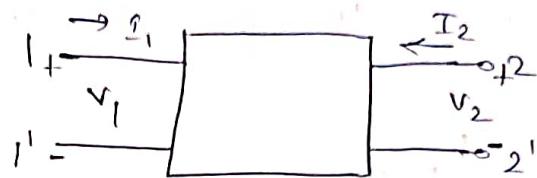
Where  $Y_{22}$  is the short-circuit driving-point admittance with the input port short-circuited. It is also called the short circuit output admittance.

The equivalent circuit of the two-port n/w in terms of  $Y$  parameters as shown below figure.



7

Transmission parameters (ABCD parameters);-



The Transmission parameters or chain parameters or ABCD parameters serve to relate the voltage and current at the input port to voltage and current at the output port.

In equation form,

$$V_1 = A V_2 - B I_2$$

$$I_1 = C V_2 - D I_2$$

\* Here, the negative sign is used with  $I_2$  and not for parameters B and D.

\* The reason is the current  $I_2$  carries a negative sign is that in transmission field, the output current is assumed to be coming out of the output port instead of going into the port.

In Matrix form, we can write

$$\begin{bmatrix} V_1 \\ I_1 \end{bmatrix} = \begin{bmatrix} A & B \\ C & D \end{bmatrix} \begin{bmatrix} V_2 \\ -I_2 \end{bmatrix}$$

Where, Matrix  $\begin{bmatrix} A & B \\ C & D \end{bmatrix}$  is called transmission matrix.

These parameters are determined as follows.

Case:- when the output port is open-circuited i.e.  $I_2 = 0$

$$A = \frac{V_1}{V_2} \Big| I_2=0$$

Where  $A$  is the reverse voltage gain with the output port open-circuited.

$$\text{Similarly, } C = \frac{I_1}{V_2} \Big| I_2=0$$

Where  $C$  is the transfer admittance with the output port open-circuited.

Case 2: - When output port is short circuited, i.e.  $V_2=0$

$$B = -\frac{V_1}{I_2} \Big| V_2=0$$

where  $B$  is the transfer impedance with the output port short-circuited.

$$\text{Similarly } D = -\frac{I_1}{I_2} \Big| V_2=0$$

Where  $D$  is the reverse current gain with the output port short-circuited.

Inverse transmission parameters ( $A' B' C' D'$  parameters).

The inverse transmission parameters serve to relate the voltage and current at the output port to the voltage and current at the input port.

In equation form,

$$V_2 = A' V_1 - B' I_1$$

$$I_2 = C' V_1 - D' I_1$$

In Matrix form, we can write

$$\begin{bmatrix} V_2 \\ I_2 \end{bmatrix} = \begin{bmatrix} A' & B' \\ C' & D' \end{bmatrix} \begin{bmatrix} V_1 \\ -I_1 \end{bmatrix}$$

where matrix  $\begin{bmatrix} A' & B' \\ C' & D' \end{bmatrix}$  is called the inverse transmission matrix.

These parameters are determined as follows

Case 1:- when the input port is open-circuited i.e.  $I_1=0$

$$A' = \frac{V_2}{V_1} \quad | I_1=0$$

where  $A'$  is the forward voltage gain with the input port open-circuited

$$\text{Similarly, } C' = \frac{I_2}{V_1} \quad | I_1=0$$

where  $C'$  is the transfer admittance with the input port open-circuited

Case 2:- when the input port is short-circuited i.e.  $V_1=0$

$$B' = -\frac{V_2}{I_1} \quad | V_1=0$$

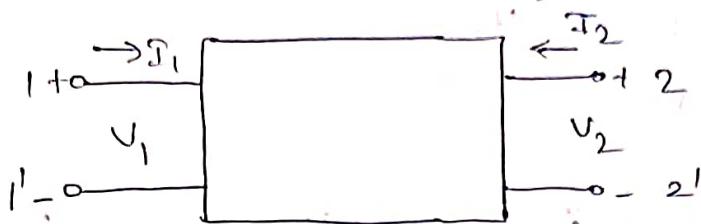
where  $B'$  is the transfer impedance with the input port short-circuited

$$\text{Similarly, } D' = \frac{I_2}{V_1} \quad | V_1=0$$

where  $D'$  is the forward current gain with the input port short-circuited.

10

Hybrid Parameters (h parameters) :-



The hybrid parameters of a two-port network may be defined by expressing the voltages of input port  $v_1$  and current of output port  $i_2$  in terms of current of input port  $i_1$  and voltage of output port  $v_2$ .

In equation form  $v_1 = h_{11} i_1 + h_{12} v_2$

$$i_2 = h_{21} i_1 + h_{22} v_2$$

In matrix form, we can write

$$\begin{bmatrix} v_1 \\ i_2 \end{bmatrix} = \begin{bmatrix} h_{11} & h_{12} \\ h_{21} & h_{22} \end{bmatrix} \begin{bmatrix} i_1 \\ v_2 \end{bmatrix}$$

The individual h parameters can be defined by setting

$$i_1 = 0 \text{ and } v_2 = 0.$$

Case 1:- when the output port is short circuited i.e  $v_2 = 0$

$$h_{11} = \frac{v_1}{i_1} \quad | v_2 = 0$$

where  $h_{11}$  is the short-circuit input impedance.

$$h_{21} = \frac{i_2}{i_1} \quad | v_2 = 0$$

where  $h_{21}$  is the short-circuit forward current gain.

Caseii:- when the input port is open-circuited i.e.  $I_1 = 0$

$$h_{12} = \frac{V_1}{V_2} \Big| I_1 = 0$$

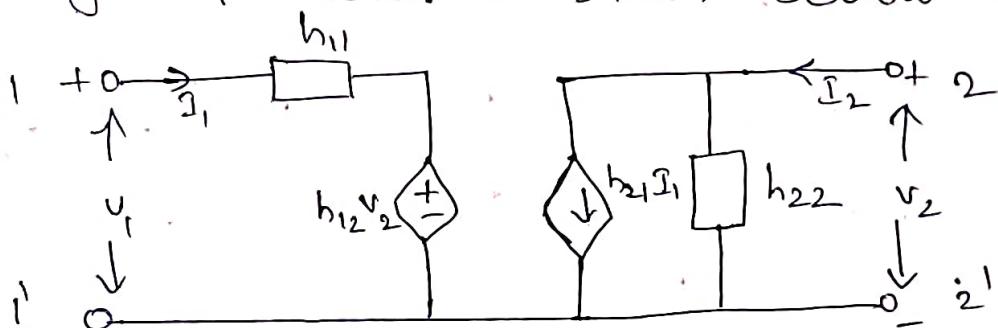
where  $h_{12}$  is the open-circuit reverse voltage gain.

$$h_{22} = \frac{I_2}{V_2} \Big| I_1 = 0$$

where  $h_{22}$  is the open-circuit output admittance.

Since h parameters represent dimensionally an impedance, an admittance, a voltage gain and a current-gain, these are called hybrid parameters.

The equivalent circuit of a two-port network in terms of hybrid parameters as shown below figure or network.



Inverse hybrid parameters (g parameters):-

The inverse hybrid parameters of a net two-port network may be defined by expressing the current of the input port  $I_1$  and voltage of the output port  $V_2$  in terms of the voltage of the input port  $V_1$  and the current of the output port  $I_2$ .

In equation form,

$$I_1 = g_{11}V_1 + g_{12}I_2$$

$$V_2 = g_{21}V_1 + g_{22}I_2$$

In matrix form, we can write

$$\begin{bmatrix} I_1 \\ V_2 \end{bmatrix} = \begin{bmatrix} g_{11} & g_{12} \\ g_{21} & g_{22} \end{bmatrix} \begin{bmatrix} V_1 \\ I_2 \end{bmatrix}$$

The individual  $g$  parameters can be defined by setting

$$V_1 = 0 \text{ and } I_2 = 0$$

Case i:- when the output port is open-circuited i.e.  $I_2 = 0$

$$g_{11} = \frac{I_1}{V_1} \Big|_{I_2=0}$$

where  $g_{11}$  is the open-circuit input admittance.

$$g_{21} = \frac{V_2}{V_1} \Big|_{I_2=0}$$

where  $g_{21}$  is the open-circuit forward voltage gain.

Case 2:- when the input port is short-circuited i.e.  $V_1 = 0$

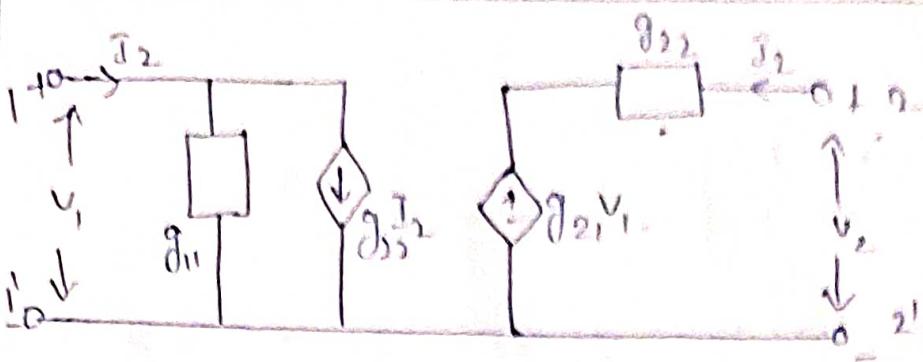
$$g_{12} = \frac{I_1}{I_2} \Big|_{V_1=0}$$

where  $g_{12}$  is the short-circuit reverse current gain.

$$g_{22} = \frac{V_2}{I_2} \Big|_{V_1=0}$$

where  $g_{22}$  is the short-circuit output impedance.

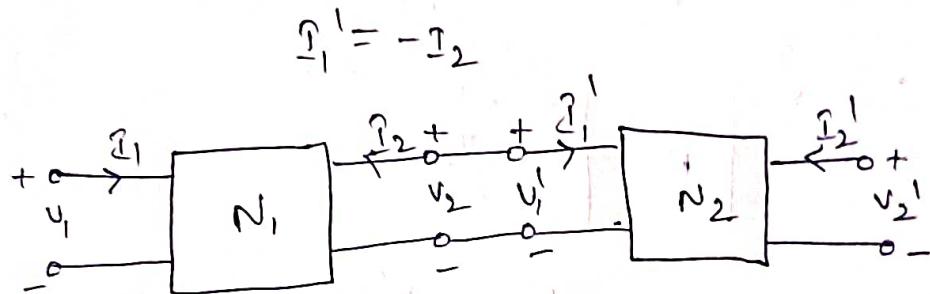
The equivalent circuit of a two-port network in terms of inverse hybrid parameters is shown below N/w.



Interconnection of two-Port networks:-

Cascade connection:-

- 1. Transmission Parameter Representation:-
- \* The below network shows the two-port networks connected in cascade.
- \* In the cascade connection, the output port of the first network becomes the input of the second network.
- \* Since it is assumed that input and output currents are positive when they enter the network, we have



Let  $A_1, B_1, C_1, D_1$  be the transmission parameters of the network  $N_1$ , and  $A_2, B_2, C_2, D_2$  be the transmission parameters of the network  $N_2$ .

For the network  $N_1$ ,

$$\begin{bmatrix} V_1 \\ I_1 \end{bmatrix} = \begin{bmatrix} A_1 & B_1 \\ C_1 & D_1 \end{bmatrix} \begin{bmatrix} V_1 \\ -I_2 \end{bmatrix} \rightarrow ①$$

For the network  $N_2$ ,

$$\begin{bmatrix} v'_1 \\ i'_1 \end{bmatrix} = \begin{bmatrix} A_2 & B_2 \\ C_2 & D_2 \end{bmatrix} \begin{bmatrix} v'_2 \\ -i'_2 \end{bmatrix} \rightarrow ③$$

Since  $v'_1 = v_2$  and  $i'_1 = -i_2$ , Then we can write

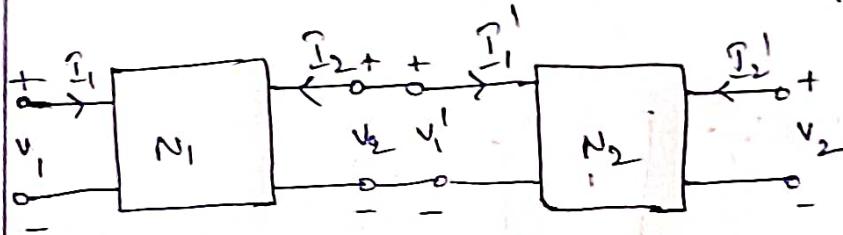
$$\begin{bmatrix} v_2 \\ -i_2 \end{bmatrix} = \begin{bmatrix} A_2 & B_2 \\ C_2 & D_2 \end{bmatrix} \begin{bmatrix} v'_2 \\ -i'_2 \end{bmatrix} \rightarrow ④$$

Combining equation ③ and ④

$$\begin{bmatrix} v_1 \\ i_1 \end{bmatrix} = \begin{bmatrix} A_1 & B_1 \\ C_1 & D_1 \end{bmatrix} \begin{bmatrix} A_2 & B_2 \\ C_2 & D_2 \end{bmatrix} \begin{bmatrix} v'_2 \\ -i'_2 \end{bmatrix} = \begin{bmatrix} A & B \\ C & D \end{bmatrix} \begin{bmatrix} v'_2 \\ -i'_2 \end{bmatrix}$$

$$\text{where } \begin{bmatrix} A & B \\ C & D \end{bmatrix} = \begin{bmatrix} A_1 & B_1 \\ C_1 & D_1 \end{bmatrix} \begin{bmatrix} A_2 & B_2 \\ C_2 & D_2 \end{bmatrix}$$

Inverse Transmission Parameter Representation:-



The above network shows two-port networks connected in cascade. Since it is assumed that input and output currents are positive when they enter the network,

$$\text{we have } -i'_1 = i_2$$

Let  $A'_1, B'_1, C'_1, D'_1$  be the transmission parameters of the nw  $N_1$  and  $A'_2, B'_2, C'_2, D'_2$  be the transmission parameters of the nw  $N_2$ .

$$\begin{bmatrix} V_2 \\ I_2 \end{bmatrix} = \begin{bmatrix} A'_1 & B'_1 \\ C'_1 & D'_1 \end{bmatrix} \begin{bmatrix} V_1 \\ -I_1 \end{bmatrix} \quad \text{for the network } N_1, \rightarrow ①$$

For the networks  $N_2$ ,

$$\begin{bmatrix} V'_2 \\ I'_2 \end{bmatrix} = \begin{bmatrix} A'_2 & B'_2 \\ C'_2 & D'_2 \end{bmatrix} \begin{bmatrix} V'_1 \\ -I'_1 \end{bmatrix} \rightarrow ②$$

Since  $V'_1 = V_2$  and  $-I'_1 = I_2$ , we can write

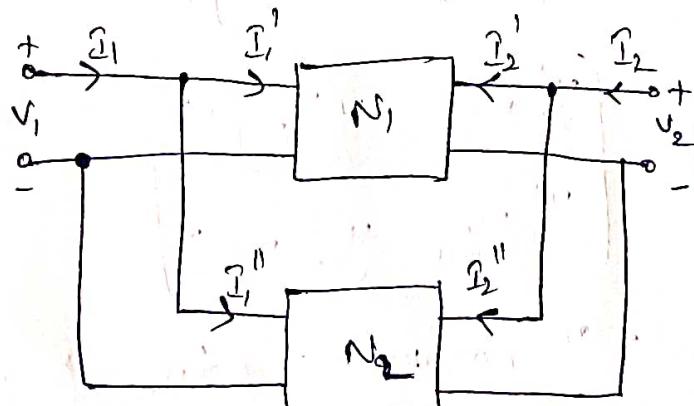
$$\begin{bmatrix} V'_2 \\ I'_2 \end{bmatrix} = \begin{bmatrix} A'_2 & B'_2 \\ C'_2 & D'_2 \end{bmatrix} \begin{bmatrix} V_2 \\ I_2 \end{bmatrix} \rightarrow ③$$

Combining equations ① and ③

$$\begin{bmatrix} V'_2 \\ I'_2 \end{bmatrix} = \begin{bmatrix} A'_2 & B'_2 \\ C'_2 & D'_2 \end{bmatrix} \begin{bmatrix} A'_1 & B'_1 \\ C'_1 & D'_1 \end{bmatrix} \begin{bmatrix} V_1 \\ -I_1 \end{bmatrix} = \begin{bmatrix} A' & B' \\ C' & D' \end{bmatrix} \begin{bmatrix} V_1 \\ -I_1 \end{bmatrix}$$

where  $\begin{bmatrix} A' & B' \\ C' & D' \end{bmatrix} = \begin{bmatrix} A'_2 & B'_2 \\ C'_2 & D'_2 \end{bmatrix} \begin{bmatrix} A'_1 & B'_1 \\ C'_1 & D'_1 \end{bmatrix}$

Parallel connection! —



Above figure shows two-port networks connected in parallel.

In the parallel connection, the two networks have the same input voltages and the same output voltages.

Let  $y_{11}', y_{12}', y_{21}', y_{22}'$  be the  $\gamma$ -parameters of the network  $N_1$  and  $y_{11}'', y_{12}'', y_{21}'', y_{22}''$  be the  $\gamma$ -parameters of the network  $N_2$ .

For the network  $N_1$ ,

$$\begin{bmatrix} \bar{I}_1' \\ \bar{I}_2' \end{bmatrix} = \begin{bmatrix} y_{11}' & y_{12}' \\ y_{21}' & y_{22}' \end{bmatrix} \begin{bmatrix} v_1 \\ v_2 \end{bmatrix}$$

For the network  $N_2$ ,

$$\begin{bmatrix} \bar{I}_1'' \\ \bar{I}_2'' \end{bmatrix} = \begin{bmatrix} y_{11}'' & y_{12}'' \\ y_{21}'' & y_{22}'' \end{bmatrix} \begin{bmatrix} v_1 \\ v_2 \end{bmatrix}$$

For the combined network,  $\bar{I}_1 = \bar{I}_1' + \bar{I}_1''$  and

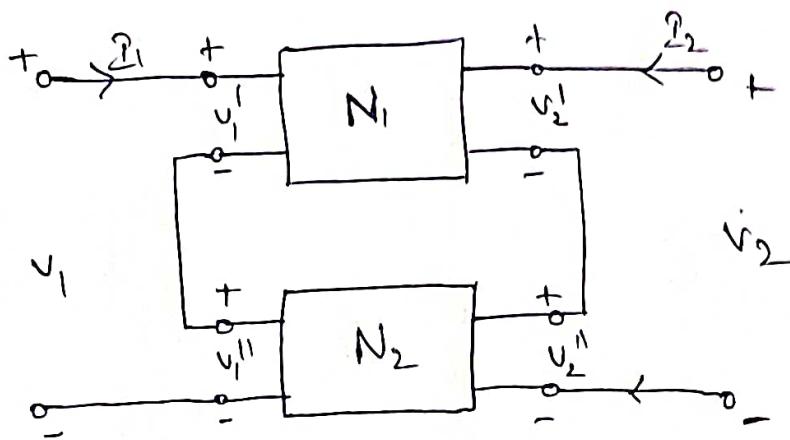
$$\bar{I}_2 = \bar{I}_2' + \bar{I}_2''$$

Hence,  $\begin{bmatrix} \bar{I}_1 \\ \bar{I}_2 \end{bmatrix} = \begin{bmatrix} \bar{I}_1' + \bar{I}_1'' \\ \bar{I}_2' + \bar{I}_2'' \end{bmatrix} = \begin{bmatrix} y_{11}' + y_{11}'' & y_{12}' + y_{12}'' \\ y_{21}' + y_{21}'' & y_{22}' + y_{22}'' \end{bmatrix} \begin{bmatrix} v_1 \\ v_2 \end{bmatrix} = \begin{bmatrix} y_{11} & y_{12} \\ y_{21} & y_{22} \end{bmatrix} \begin{bmatrix} v_1 \\ v_2 \end{bmatrix}$

where  $\begin{bmatrix} y_{11} & y_{12} \\ y_{21} & y_{22} \end{bmatrix} = \begin{bmatrix} y_{11}' + y_{11}'' & y_{12}' + y_{12}'' \\ y_{21}' + y_{21}'' & y_{22}' + y_{22}'' \end{bmatrix}$

Thus, the resultant  $\gamma$ -parameter matrix for parallel connected networks is the sum of  $\gamma$ -matrices of each individual two-port networks.

Series connection:-



The above figure shows two-port networks connected in series. In a series connection, both the networks carry the same input current. Their output currents are also equal.

Let  $\underline{Z}_{11}^1, \underline{Z}_{12}^1, \underline{Z}_{21}^1, \underline{Z}_{22}^1$  be the Z-parameters of the network  $N_1$  and

$\underline{Z}_{11}^{\prime\prime}, \underline{Z}_{12}^{\prime\prime}, \underline{Z}_{21}^{\prime\prime}, \underline{Z}_{22}^{\prime\prime}$  be the Z-parameters of the network  $N_2$ ,

For the network  $N_1$ ,

$$\begin{bmatrix} v_1^+ \\ v_2^+ \end{bmatrix} = \begin{bmatrix} \underline{Z}_{11}^1 & \underline{Z}_{12}^1 \\ \underline{Z}_{21}^1 & \underline{Z}_{22}^1 \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \end{bmatrix}$$

For the network  $N_2$ ,

$$\begin{bmatrix} v_1''^+ \\ v_2''^+ \end{bmatrix} = \begin{bmatrix} \underline{Z}_{11}^{\prime\prime} & \underline{Z}_{12}^{\prime\prime} \\ \underline{Z}_{21}^{\prime\prime} & \underline{Z}_{22}^{\prime\prime} \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \end{bmatrix}$$

for the combined network

$$v_1 = v_1^+ + v_1''^+ \text{ and } v_2 = v_2^+ + v_2''^+$$

$$\text{Hence, } \begin{bmatrix} v_1 \\ v_2 \end{bmatrix} = \begin{bmatrix} v_1^+ + v_1''^+ \\ v_2^+ + v_2''^+ \end{bmatrix} = \begin{bmatrix} \underline{Z}_{11}^1 + \underline{Z}_{11}^{\prime\prime} & \underline{Z}_{12}^1 + \underline{Z}_{12}^{\prime\prime} \\ \underline{Z}_{21}^1 + \underline{Z}_{21}^{\prime\prime} & \underline{Z}_{22}^1 + \underline{Z}_{22}^{\prime\prime} \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \end{bmatrix} = \begin{bmatrix} \underline{Z}_{11} & \underline{Z}_{12} \\ \underline{Z}_{21} & \underline{Z}_{22} \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \end{bmatrix}$$

$$\text{where } \begin{bmatrix} \underline{Z}_{11} & \underline{Z}_{12} \\ \underline{Z}_{21} & \underline{Z}_{22} \end{bmatrix} = \begin{bmatrix} \underline{Z}_{11}^1 + \underline{Z}_{11}^{\prime\prime} & \underline{Z}_{12}^1 + \underline{Z}_{12}^{\prime\prime} \\ \underline{Z}_{21}^1 + \underline{Z}_{21}^{\prime\prime} & \underline{Z}_{22}^1 + \underline{Z}_{22}^{\prime\prime} \end{bmatrix}$$

### 13.52 Network Analysis and Synthesis

$$g_{21} = -\frac{h_{21}}{\Delta h}$$

$$g_{22} = \frac{h_{11}}{\Delta h}$$

**Table 13.3 Inter-relationship between parameters**

$$\Delta X = X_{11}X_{22} - X_{12}X_{21}$$

		In terms of					
[Z]		[Y]	[T]	[T']	[h]	[g]	
[Z]	$Z_{11} Z_{12}$	$\frac{Y_{22}}{\Delta Y} - \frac{Y_{12}}{\Delta Y}$	$\frac{A}{C} \frac{\Delta T}{C}$	$\frac{D'}{C'} \frac{1}{C'}$	$\frac{\Delta h}{h_{22}} \frac{h_{12}}{h_{22}}$	$\frac{1}{g_{11}} - \frac{g_{21}}{g_{11}}$	
	$Z_{21} Z_{22}$	$-\frac{Y_{21}}{\Delta Y} \frac{Y_{11}}{\Delta Y}$	$\frac{1}{C} \frac{D}{C}$	$\frac{\Delta T'}{C'} \frac{A'}{C'}$	$-\frac{h_{21}}{h_{22}} \frac{1}{h_{22}}$	$\frac{g_{11}}{g_{11}} \frac{\Delta g}{g_{11}}$	
[Y]	$\frac{Z_{22}}{\Delta Z} - \frac{Z_{12}}{\Delta Z}$	$Y_{11} Y_{12}$	$\frac{D}{B} - \frac{\Delta T}{B}$	$\frac{A'}{B'} - \frac{1}{B'}$	$\frac{1}{h_{11}} - \frac{h_{12}}{h_{11}}$	$\frac{\Delta g}{g_{22}} \frac{g_{12}}{g_{22}}$	
	$-\frac{Z_{21}}{\Delta Z} \frac{Z_{11}}{\Delta Z}$	$Y_{21} Y_{22}$	$-\frac{1}{B} \frac{A}{B}$	$\frac{\Delta T'}{B'} \frac{D'}{B'}$	$\frac{h_{21}}{h_{11}} \frac{\Delta h}{h_{11}}$	$-\frac{g_{21}}{g_{22}} \frac{1}{g_{22}}$	
[T]	$\frac{Z_{11}}{Z_{21}} \frac{\Delta Z}{Z_{21}}$	$-\frac{Y_{22}}{Y_{21}} - \frac{1}{Y_{21}}$	$A B$	$\frac{D'}{\Delta T'} \frac{B'}{\Delta T'}$	$-\frac{\Delta h}{h_{21}} - \frac{h_{11}}{h_{21}}$	$\frac{1}{g_{21}} \frac{g_{11}}{g_{21}}$	
	$\frac{1}{Z_{21}} \frac{Z_{22}}{Z_{21}}$	$-\frac{\Delta Y}{Y_{21}} - \frac{Y_{11}}{Y_{21}}$	$C D$	$\frac{C'}{\Delta T'} \frac{A'}{\Delta T'}$	$-\frac{h_{22}}{h_{21}} - \frac{1}{h_{21}}$	$\frac{g_{11}}{g_{21}} \frac{\Delta g}{g_{21}}$	
[T']	$\frac{Z_{22}}{Z_{12}} \frac{\Delta Z}{Z_{12}}$	$-\frac{Y_{11}}{Y_{12}} - \frac{1}{Y_{12}}$	$\frac{D}{\Delta T} \frac{B}{\Delta T}$	$A' B'$	$\frac{1}{h_{12}} \frac{h_{11}}{h_{12}}$	$-\frac{\Delta g}{g_{12}} - \frac{g_{11}}{g_{12}}$	
	$\frac{1}{Z_{12}} \frac{Z_{11}}{Z_{12}}$	$-\frac{\Delta Y}{Y_{12}} - \frac{Y_{22}}{Y_{12}}$	$\frac{C}{\Delta T} \frac{A}{\Delta T}$	$C' D'$	$\frac{h_{22}}{h_{12}} \frac{\Delta h}{h_{12}}$	$-\frac{g_{11}}{g_{12}} - \frac{1}{g_{12}}$	
[h]	$\frac{\Delta Z}{Z_{22}} \frac{Z_{12}}{Z_{22}}$	$\frac{1}{Y_{11}} - \frac{Y_{12}}{Y_{11}}$	$\frac{B}{D} \frac{\Delta T}{D}$	$\frac{B'}{A'} \frac{1}{A'}$	$h_{11} h_{12}$	$\frac{g_{11}}{\Delta g} - \frac{g_{11}}{\Delta g}$	
	$-\frac{Z_{21}}{Z_{22}} \frac{1}{Z_{22}}$	$\frac{Y_{21}}{Y_{11}} \frac{\Delta Y}{Y_{11}}$	$-\frac{1}{D} \frac{C}{D}$	$-\frac{\Delta T'}{A'} \frac{C'}{A'}$	$h_{21} h_{22}$	$-\frac{g_{21}}{\Delta g} \frac{g_{11}}{\Delta g}$	
[g]	$\frac{1}{Z_{11}} - \frac{Z_{12}}{Z_{11}}$	$\frac{\Delta Y}{Y_{22}} \frac{Y_{12}}{Y_{22}}$	$\frac{C}{A} - \frac{\Delta T}{A}$	$\frac{C'}{D'} - \frac{1}{D'}$	$\frac{h_{22}}{\Delta h} - \frac{h_{12}}{\Delta h}$	$g_{11} g_{12}$	
	$\frac{Z_{21}}{Z_{11}} \frac{\Delta Z}{Z_{11}}$	$-\frac{Y_{21}}{Y_{22}} \frac{1}{Y_{22}}$	$\frac{1}{A} \frac{B}{A}$	$\frac{\Delta T'}{D'} \frac{B'}{D'}$	$-\frac{h_{21}}{\Delta h} \frac{h_{11}}{\Delta h}$	$g_{21} g_{22}$	

**Example 13.26** The Z parameters of a two-port network are  $Z_{11} = 20 \Omega$ ,  $Z_{22} = 30 \Omega$ ,  $Z_{12} = 10 \Omega$ . Find Y and ABCD parameters.